2A, 1.2MHz Synchronous Step-Down Converter

General Description

The RT2652 is a high efficiency synchronous, step-down DC/DC converter. The available input voltage range is from 2.7V to 5.5V the regulated output voltage is adjustable from 0.6V to VIN while delivering up to 2A of output current. The internal synchronous low on-resistance power switches increase efficiency and eliminate the need for an external Schottky diode. The switching frequency is fixed internally at 1.2MHz. The 100% duty cycle provides low dropout operation extending battery life in portable systems. Current mode operation with internal compensation allows the transient response to be optimized. The RT2652 is available in the WDFN-10L 3x3 package.

Ordering Information

RT2652□□

Package Type
QW : WDFN-10L 3x3 (W-Type)

Lead Plating System
G : Green (Halogen Free and Pb Free)

Note :

Richtek products are :

- RoHS compliant and compatible with the current requirements of IPC/JEDEC J-STD-020.
- Suitable for use in SnPb or Pb-free soldering processes.

Features

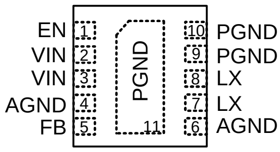
- **High Efficiency : Up to 95%**
- **Fixed Frequency : 1.2MHz**
- **No Schottky Diode Required**
- **Internal Compensation**
- **0.6V Reference Allows Low Output Voltage**
- **100% Duty Cycle for Low Dropout Operation**
- **OCP, UVP, OTP**
- **RoHS Compliant and Halogen Free**

Applications

- Enterprise Servers
- Ethernet Switches & Routers
- Global Storage
- Telecom & Industrial
- Cell Phones & DSC's

Pin Configurations

(TOP VIEW)



WDFN-10L 3x3

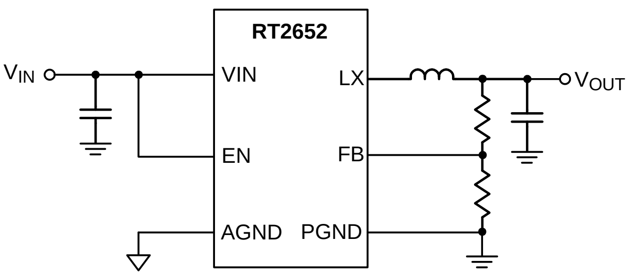
Marking Information

0K=YM
DNN

0K= : Product Code

YMDNN : Date Code

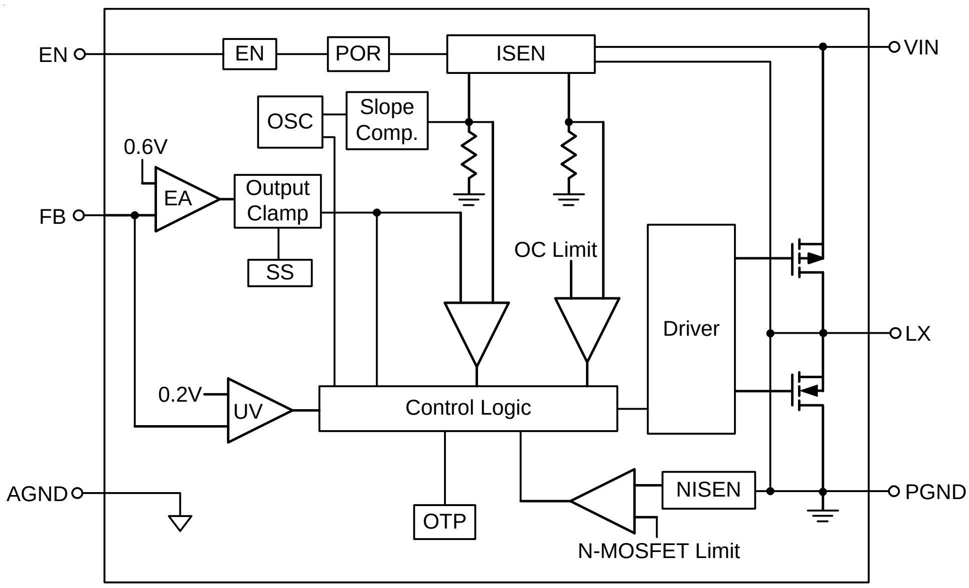
Simplified Application Circuit



Functional Pin Description

Pin No.	Pin Name	Pin Function
1	EN	Enable Control Input. Pull high to turn on. Do not float.
2, 3	VIN	Power Input. Decouple this pin to GND with a 22μF ceramic capacitor at least.
4, 6	AGND	Analog Ground.
5	FB	Feedback Voltage Input. This pin receives the feedback voltage from an external resistive divider connected across the output.
7, 8	LX	Power MOSFET Switch Node. Connect this pin to the inductor.
9, 10, 11 (Exposed Pad)	PGND	Power Ground. The exposed pad must be soldered to a large PCB and connected to GND for maximum power dissipation.

Function Block Diagram



Operation

The RT2652 is a monolithic, constant-frequency, current mode step-down DC/DC converter. During normal operation, the internal high side MOSFET is turned on at the beginning of each cycle. Current in the inductor increases until the peak inductor current reaches the value defined by the internal error amplifier. The error amplifier adjusts the voltage of its output by comparing the feedback signal from a resistor divider on the FB pin with an internal 0.6V reference. When the load current increases, it causes a reduction in the feedback voltage relative to the reference. The error amplifier raises its output voltage until the average inductor current matches the new load current. When the high side MOSFET turns off, the synchronous power switch (N-MOSFET) turns on until either the bottom current limit is reached or the beginning of the next cycle. The operating frequency is set by the internal oscillator at 1.2MHz. In VIN larger than 6V condition, the high side MOSFET is turned off and the low side MOSFET is switched on until either the VIN over voltage condition is cleared or the low side MOSFET's current limit is reached.

Absolute Maximum Ratings (Note 1)

- Supply Voltage, VIN -- –0.3V to 6.5V
- Switch Node Voltage, LX --- –0.3V to (V_{IN} + 0.3V)
- Other Pins --- –0.3V to 6.5V
- Power Dissipation, P_D @ T_A = 25°C
  WDFN-10L 3x3 --- 1.429W
- Package Thermal Resistance (Note 2)
  WDFN-10L 3x3, θ_{JA} --- 70°C/W
  WDFN-10L 3x3, θ_{JC} --- 8.2°C/W
- Junction Temperature -- 150°C
- Lead Temperature (Soldering, 10 sec.) ------------------------------ 260°C
- Storage Temperature Range -- –65°C to 150°C
- ESD Susceptibility (Note 3)
  HBM (Human Body Model) --- 2kV

Recommended Operating Conditions (Note 4)

- Supply Voltage, VIN -- 2.7V to 5.5V
- Junction Temperature Range -- –40°C to 125°C
- Ambient Temperature Range --- –40°C to 85°C

Electrical Characteristics

(V_{IN} = 3.3V, T_A = 25°C unless otherwise specified)

Parameter		Symbol	Test Conditions	Min	Typ	Max	Unit
Feedback Reference Voltage		V_{REF}		0.594	0.6	0.606	V
Feedback Leakage Current		I_{FB}		--	0.1	0.4	μA
DC Bias Current			Active, V_{FB} =0.7V, not switching	--	110	220	μA
			Shutdown	--	--	1	μA
Output Voltage Line Regulation			V_{IN} = 2.7V to 5.5V	--	0.04	--	%/V
Output Voltage Load Regulation			I_{OUT} = 10mA to 2000mA	--	0.2	--	%/A
Switch Leakage Current			EN = 0V	--	--	1	μA
Switching Frequency				0.96	1.2	1.44	MHz
Switch On-Resistance	High-Side	$R_{DS(ON)\_H}$		--	110	130	mΩ
	Low-Side	$R_{DS(ON)\_L}$		--	70	90	
P-MOSFET Current Limit		I_{LIM}		2.5	3.5	--	A
Under Voltage Lockout Threshold			VDD Rising	--	2.4	--	V
			VDD Falling	--	2.2	--	V
EN Input Voltage	Logic-High	V_{IH}		1.5	--	--	V
	Logic-Low	V_{IL}		--	--	0.3	
EN Pull Low Resistance				--	500	--	kΩ

Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit
Over Temperature Protection (latch-off)			--	150	--	°C
Soft-Start Time			1.3	--	--	ms
V_{OUT} Discharge Resistance			--	100	150	Ω
V_{OUT} UVP (latch-off)			--	33	--	%

Note 1. Stresses beyond those listed "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions may affect device reliability.

Note 2. θ_{JA} is measured at T_A = 25°C on a high effective thermal conductivity four-layer test board per JEDEC 51-7. θ_{JC} is measured at the exposed pad of the package.

Note 3. Devices are ESD sensitive. Handling precaution is recommended.

Note 4. The device is not guaranteed to function outside its operating conditions..

Typical Application Circuit

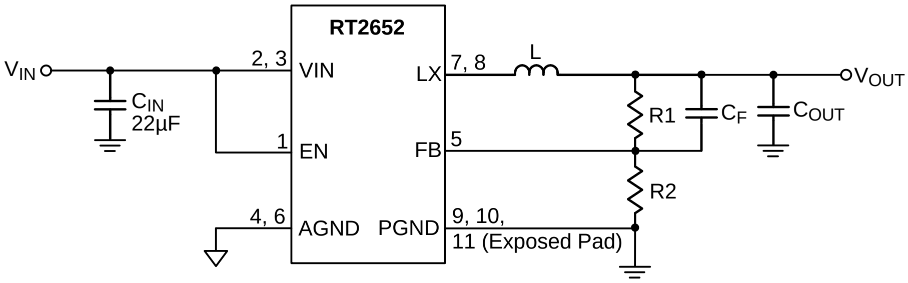
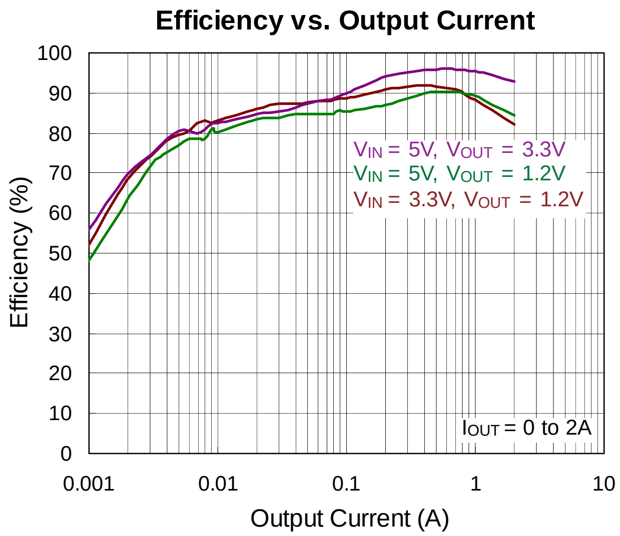
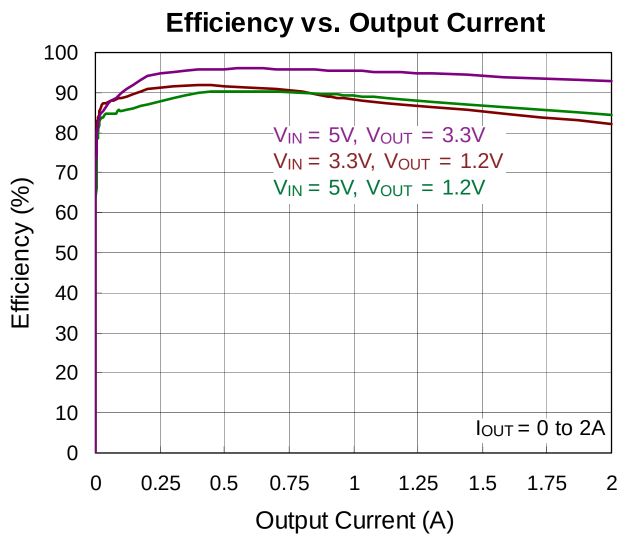


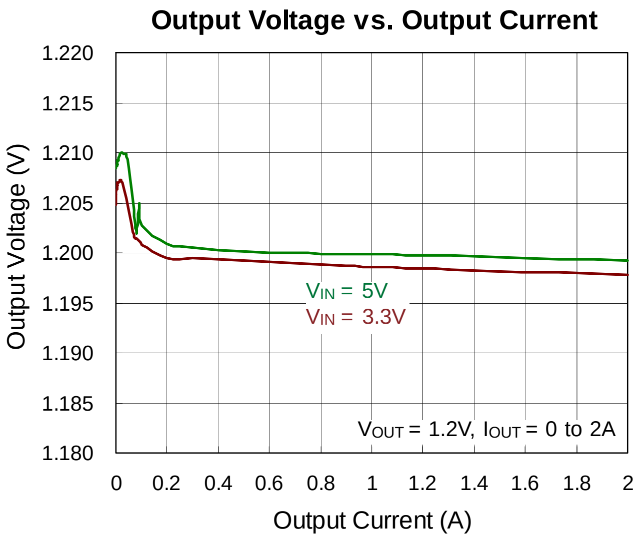
Table 1. Recommended Component Selection

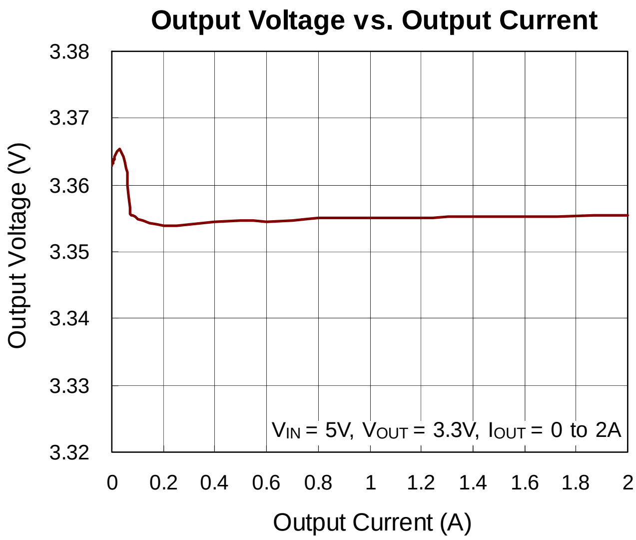
V_{OUT} (V)	R1 (kΩ)	R2 (kΩ)	C_F (pF)	L (μH)	C_{OUT} (μF)
3.3	37	8.2	200	2	22
2.5	26	8.2	200	2	22
1.8	16.5	8.2	200	1.5	22
1.5	12.3	8.2	200	1.5	22
1.2	8.2	8.2	200	1.5	22
1	5.6	8.2	200	1.5	22

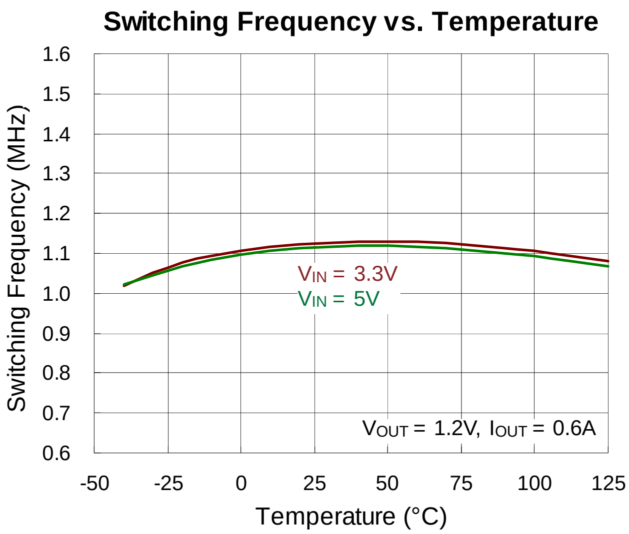
Typical Operating Characteristics

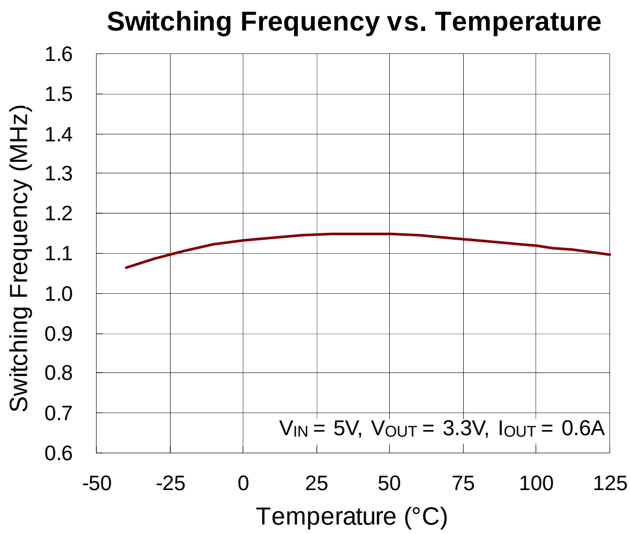




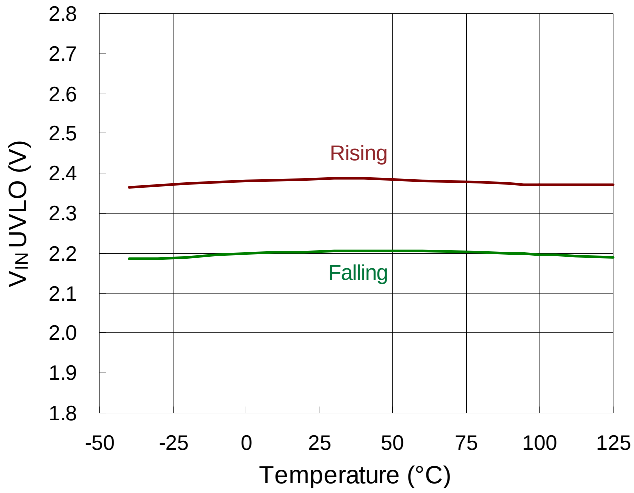




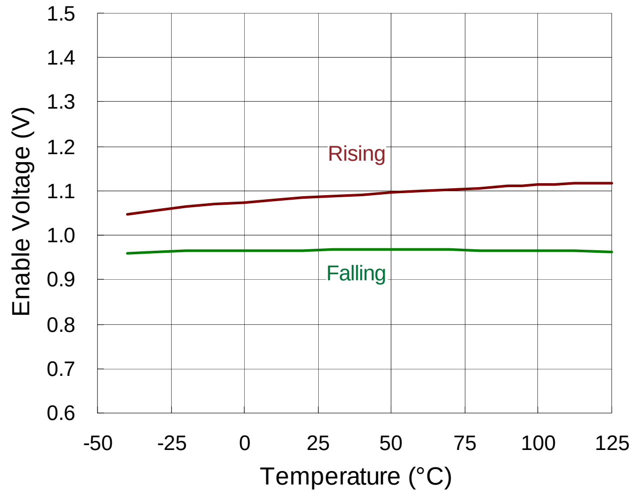




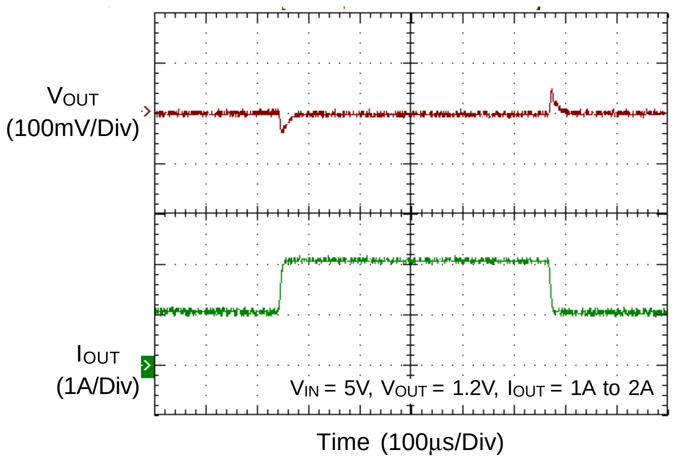
V_{IN} UVLO vs. Temperature



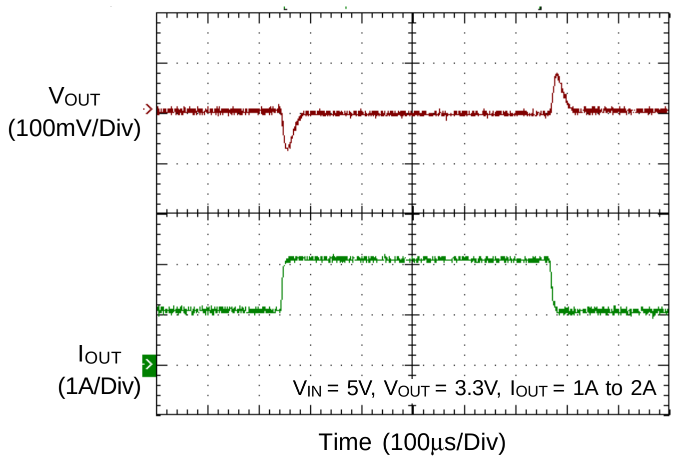
Enable Voltage vs. Temperature



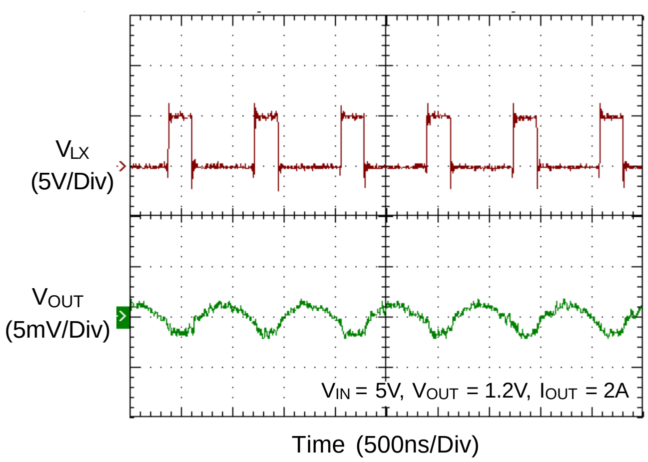
Load Transient Response



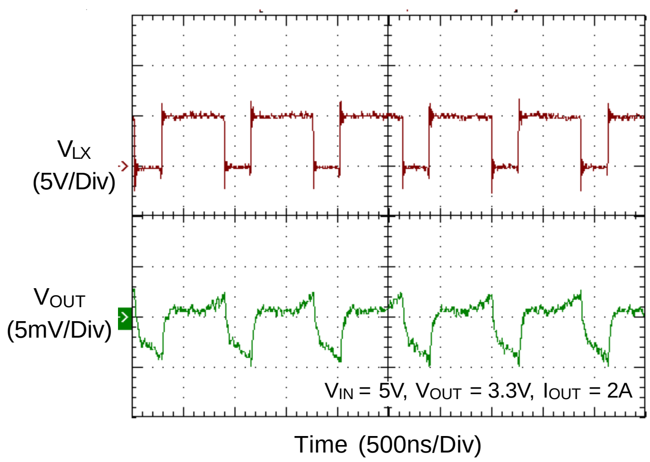
Load Transient Response



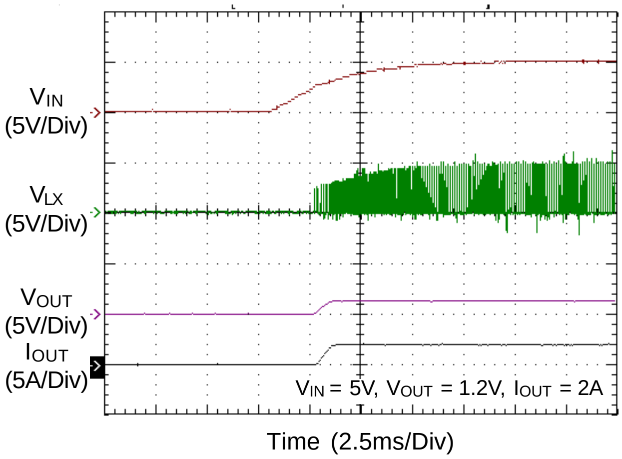
Output Ripple Voltage



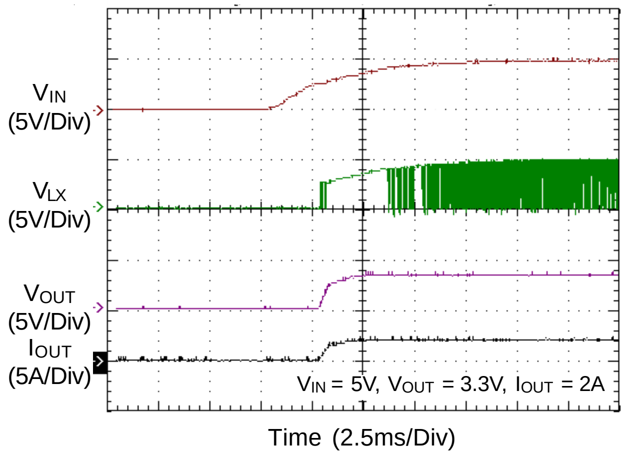
Output Ripple Voltage



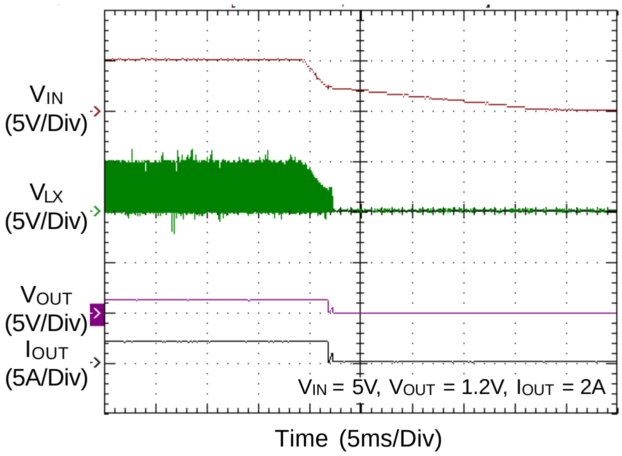
Power On from VIN



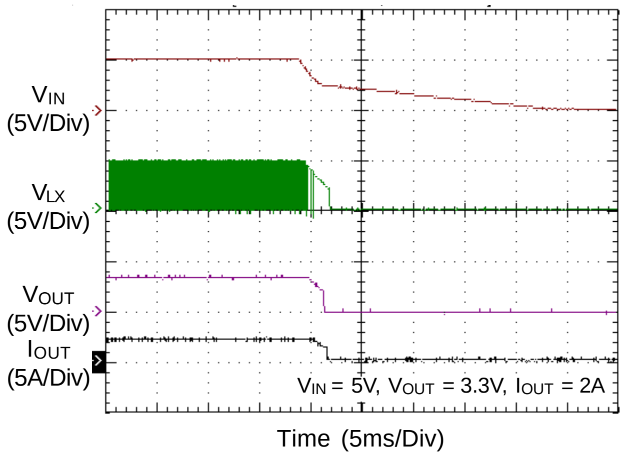
Power On from VIN



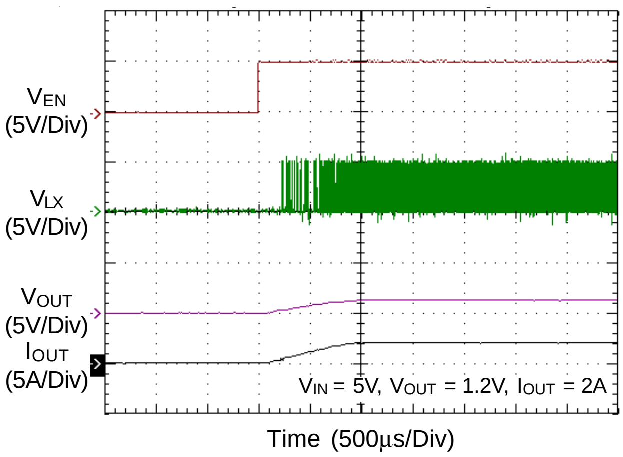
Power Off from VIN



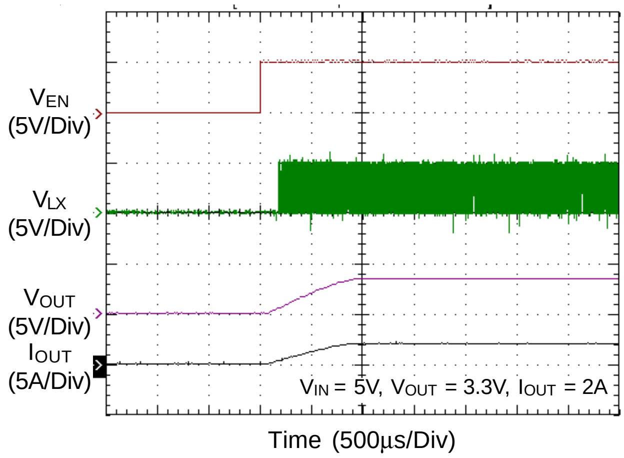
Power Off from VIN



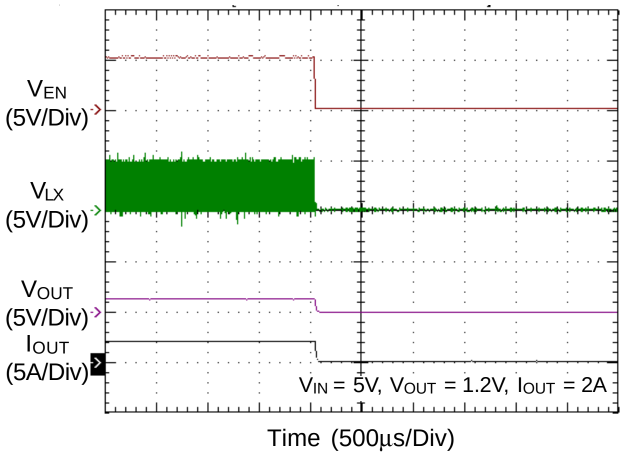
Power On from EN



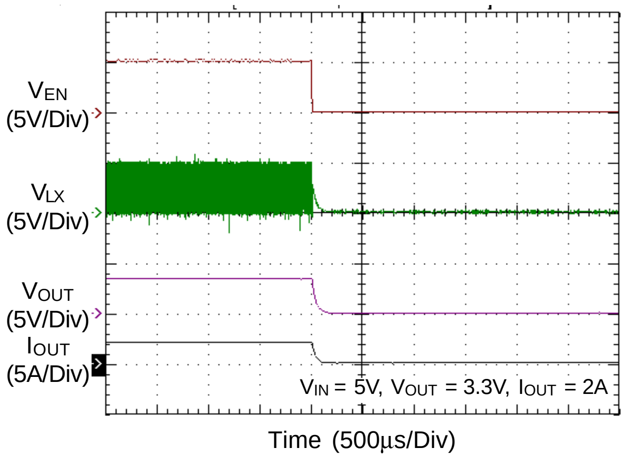
Power On from EN



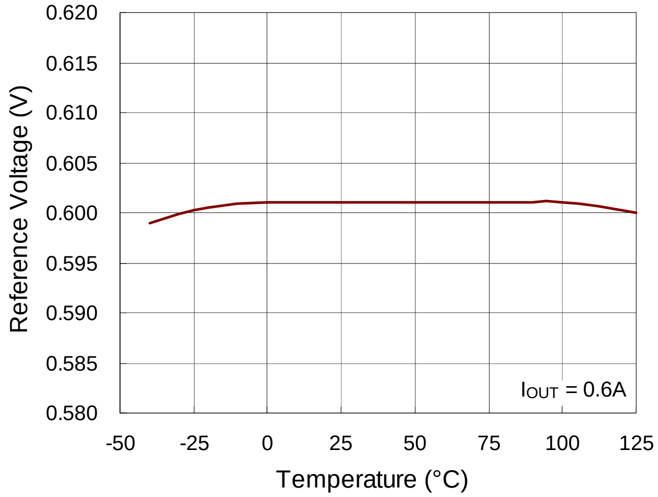
Power Off from EN



Power Off from EN



Reference Voltage vs. Temperature



Application Information

The RT2652 is a single-phase buck PWM converter. It provides single feedback loop, current mode control with fast transient response. An internal 0.6V reference allows the output voltage to be precisely regulated for low output voltage applications. A fixed switching frequency (1.2MHz) oscillator and internal compensation are integrated to minimize external component count.

Output Voltage Setting

The output voltage is set by an external resistive voltage divider according to the following equation :

$$V_{OUT} = V_{REF}\left(1+\frac{R1}{R2}\right)$$

Where V_{REF} is equals 0.6V (typ.).

The resistive voltage divider allows the FB pin to sense a fraction of the output voltage as shown in Figure 1.

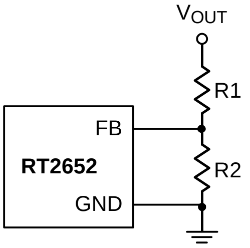


Figure 1. Setting the Output Voltage

Inductor Selection

For a given input and output voltage, the inductor value and operating frequency determine the ripple current. The ripple current ΔI_L increases with higher V_{IN} and decreases with higher inductance.

$$\Delta I_L = \left[\frac{V_{OUT}}{f \times L}\right] \times \left[1 - \frac{V_{OUT}}{V_{IN}}\right]$$

Having a lower ripple current reduces not only the ESR losses in the output capacitors but also the output voltage ripple. Highest efficiency operation is achieved by reducing ripple current at low frequency, but a large inductor is required to attain this goal. For ripple current selection, the value of $\Delta I_L = 0.4(I_{MAX})$ is a reasonable starting point. The largest ripple current occurs at the highest V_{IN}. To guarantee that the ripple current stays below a specified maximum value, the inductor should be chosen according to the following equation :

$$L = \left[\frac{V_{OUT}}{f \times \Delta I_{L(MAX)}}\right] \times \left[1 - \frac{V_{OUT}}{V_{IN(MAX)}}\right]$$

Slope Compensation and Inductor Peak Current

Slope compensation provides stability in constant frequency architectures by preventing sub-harmonic oscillations at duty cycles greater than 50%. It is accomplished internally by adding a compensating ramp to the inductor current signal. Normally, the maximum inductor peak current is reduced when slope compensation is added. In this IC, however, separated inductor current signal is used to monitor over current condition and this keeps the maximum output current relatively constant regardless of duty cycle.

Low Dropout Operation

The RT2652 is designed to operate down to an input supply voltage of 2.7V. One important consideration at low input supply voltage is that the $R_{DS(ON)}$ of the P-Channel and N-Channel power switches increases. The user should calculate the power dissipation when the RT2652 is used at 100% duty cycle with low input voltages to ensure that thermal limits are not exceeded. Slope compensation and inductor peak current slope compensation provides stability in constant frequency architectures by preventing sub-harmonic oscillations at duty cycles greater than 50%.It is accomplished internally by adding a compensating ramp to the inductor current signal. Normally, the maximum inductor peak current is reduced when slope compensation is added. In the RT2652, however, separated inductor current signals are used to monitor over current condition. This keeps the maximum output current relatively constant regardless of duty cycle.

Short Circuit Protection

When the output is shorted to ground, the inductor current decays very slowly during a single switching cycle. A current runaway detector is used to monitor inductor current. As current increases beyond the control of current loop, switching cycles will be skipped to prevent current runaway from occurring.

Under Voltage Lockout Threshold

The IC includes an input Under Voltage Lockout Protection (UVLO). If the input voltage exceeds the UVLO rising threshold voltage, the converter resets and prepares the PWM for operation. If the input voltage falls below the UVLO falling threshold voltage during normal operation, the device stops switching. The UVLO rising and falling threshold voltage includes a hysteresis to prevent noise caused reset.

Thermal Shutdown

The device implements an internal thermal shutdown function when the junction temperature exceeds 150°C. The thermal shutdown disables the device until the junction temperature drops below the hysteresis (20°C typ.). Then, the device is re-enabled and automatically reinstates the power up sequence.

Thermal Considerations

For continuous operation, do not exceed absolute maximum junction temperature. The maximum power dissipation depends on the thermal resistance of the IC package, PCB layout, rate of surrounding airflow, and difference between junction and ambient temperature. The maximum power dissipation can be calculated by the following formula :

$P_{D(MAX)} = (T_{J(MAX)} - T_A) / \theta_{JA}$

where $T_{J(MAX)}$ is the maximum junction temperature, T_A is the ambient temperature, and θ_{JA} is the junction to ambient thermal resistance.

For recommended operating condition specifications, the maximum junction temperature is 125°C. The junction to ambient thermal resistance, θ_{JA}, is layout dependent. For WDFN-10L 3x3 packages, the thermal resistance, θ_{JA}, is 70°C/W on a standard JEDEC 51-7 four-layer thermal test board. The maximum power dissipation at T_A = 25°C can be calculated by the following formulas :

$P_{D(MAX)}$ = (125°C − 25°C) / (70°C/W) = 1.429W for WDFN-10L 3x3 package

The maximum power dissipation depends on the operating ambient temperature for fixed $T_{J(MAX)}$ and thermal resistance, θ_{JA}. The derating curves in Figure 2 allow the designer to see the effect of rising ambient temperature on the maximum power dissipation.

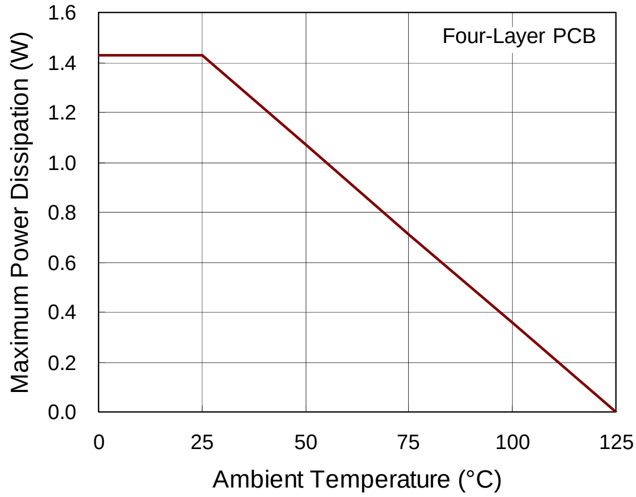


Figure 2. Derating Curve of Maximum Power Dissipation

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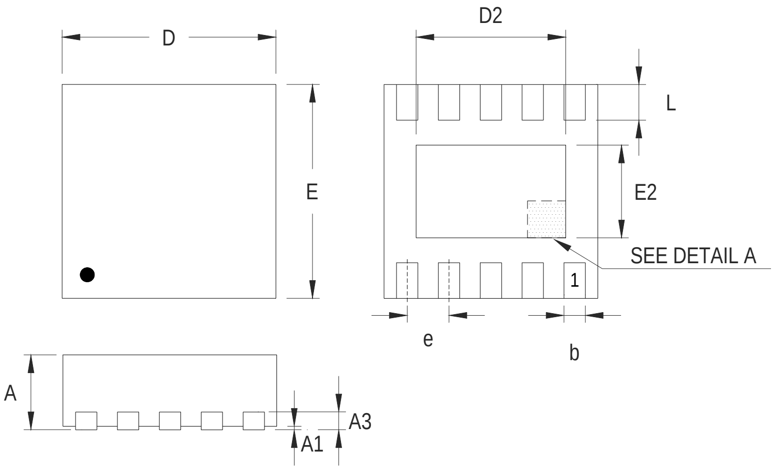
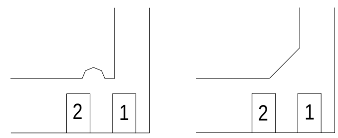
Outline Dimension





DETAIL A

Pin #1 ID and Tie Bar Mark Options

Note : The configuration of the Pin #1 identifier is optional, but must be located within the zone indicated.

Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	0.700	0.800	0.028	0.031
A1	0.000	0.050	0.000	0.002
A3	0.175	0.250	0.007	0.010
b	0.180	0.300	0.007	0.012
D	2.950	3.050	0.116	0.120
D2	2.300	2.650	0.091	0.104
E	2.950	3.050	0.116	0.120
E2	1.500	1.750	0.059	0.069
e	0.500		0.020	
L	0.350	0.450	0.014	0.018

W-Type 10L DFN 3x3 Package

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