Multi-Channel Integrated Power Management IC

General Description

The MAX77752 is a highly-integrated power management solution including three step-down converters, a low-dropout linear regulator, two external regulator enable outputs, two dedicated load switch controllers, and an inrush-current limiter which can be configured as a third load switch controller using OTP. The MAX77752 provides a combination of high-performance power management components, high-accuracy monitoring, and a customized top level controller that results in an efficient, size optimized solution.

The 40-pin, 5mm x 5mm x 0.8mm, 0.4mm pitch TQFN package is ideal for space constrained applications.

Numerous factory programmable options allow the device to be tailored for many variations of the end application.

Applications

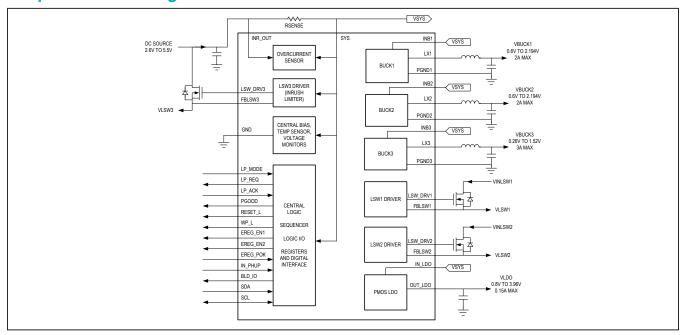
- Solid-State Drive Systems
- Handheld Devices
- Gaming Consoles
- Drones
- Automation Systems
- Cameras

Benefits and Features

- Highly Integrated
 - · Three Buck Regulators
 - Integrated High-Accuracy Brownout Comparators
 - · One Low-Dropout Linear Regulator
 - Low-Input Voltage
 - · Two Dedicated Load Switch Controllers
 - One Inrush-Current Limiter, Configurable to be Load Switch 3 Controller Using OTP
 - Two External Regulator Enable Outputs
 - Voltage Monitor for Backup Power Control
- Highly Flexible and Configurable
 - I²C-Compatible Interface
 - · Factory OTP Options Available
 - Flexible Power Sequencer
 - · Configurable Sleep State Control
- Small Size
 - 40-Pin, 5mm x 5mm x 0.8mm, 0.4mm Pitch TQFN
 - 70mm² Total Solution Size

Ordering Information appears at end of data sheet.

Simplified Block Diagram





Absolute Maximum Ratings

Тор	
IN_DRV to GND	0.3V to +16.0V
IN_SNS to GND (Note 1)	
INR_OUT to GND	0.3V to +6.0V
SYS to GND	0.3V to +6.0V
IN_PHUP to GND	0.3V to +6.0V
RESET_L to GND	0.3V to V _{SYS} +0.3V
LP_REQ to GND	0.3V to V _{SYS} +0.3V
LP_ACK to GND	0.3V to V _{SYS} +0.3V
LP_MODE to GND	0.3V to V _{SYS} +0.3V
WP_L to GND (Note 2)	
PGOOD to GND (Note 2)	
EREG_EN1 to GND (Note 2)	0.3V to V _{H_INT}
EREG_EN2 to GND	0.3V to 6.0V
EREG_POK to GND	
BLD_IO to GND (Note 2)	
WP_L Sink Current	35mA
RESET_L Sink Current	
PGOOD Sink Current	
EREG_EN1 Sink Current	
EREG_EN2 Sink Current	
LP_REQ Sink Current	35mA
DGND to GND	0.3V to +0.3V
LDO	
IN_LDO to GND	0.3V to +6.0V
OUT_LDO to GND	0.3V to V _{IN_LDO} +0.3V

Buck
INB1, INB2, INB3 to SYS0.3V to +0.3V
INB1 to PGND10.3V to +6.0V
INB2 to PGND20.3V to +6.0V
INB3 to PGND30.3V to +6.0V
LX1 to PGND1 (Note 3)0.3V to V _{INB1} +0.3V
LX2 to PGND2 (Note 3)0.3V to V _{INB2} +0.3V
LX3 to PGND3 (Note 3)0.3V to V _{INB3} +0.3V
LX1, LX2 RMS Current per pin (T _J = +110°C)
(RMS current per pin (T _J = +110°C))1.7A
LX3 RMS Current per pin ($T_J = +110^{\circ}C$)
(RMS current per pin (T _J = +110°C))
FBB1, FBB2, FBB3 to GND0.3V to V _{SYS} +0.3V
PGND1, PGND2, PGND3 to GND0.3V to +0.3V
I ² C
SDA, SCL to GND0.3V to V _{IN_VIO_I2C} +0.3V
SDA Sink Current35mA
Load Switch
LSW_DRV1 to GND0.3V to +16.0V
LSW_DRV2 to GND0.3V to +16.0V
FBLSW1 to GND0.3V to V _{SYS} +0.3V
FBLSW2 to GND0.3V to V _{SYS} +0.3V
Continuous Power Dissipation (Multilayer Board)
$T_A = +70$ °C, derate 35.70mW/°C
above +70°C mW to 2857.1mW
Operating Temperature Range40°C to +85°C
Junction Temperature+150°C
Storage Temperature Range40°C to +150°C
Soldering Temperature (reflow)+260°C

- Note 1: IN_SNS voltage ramp rates greater than 2.8V/µs trigger the internal ESD device and should be avoided. The ESD device recovers if exposed to an excessive ramp rate.
- Note 2: $V_{H\ INT}$ is the maximum voltage of V_{SYS} and $V_{IN\ PHUP}$.
- Note 3: The specified voltage limitation is for steady state conditions. Dead times of a few nano seconds exist during the dynamic BUCK regulator transitions from inductor charging to inductor discharging and vice versa. These dead times allow internal clamping diodes to PGNDx and INBx to forward bias (Vf~1V). When the LXx waveform is observed on a high-bandwidth oscilloscope (≥100MHz), the LXx transition edges are commonly seen with 1.5V spikes. These spikes are due to (1) the internal clamping diode forward voltage and (2) the high rate of current change through the current loop's inductance (V = L x di/dt). Designs must follow the recommended printed circuit board (PCB) layout in order to minimize this current loop's inductance.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Package Information

TQFN

PACKAGE CODE	T4055+1C
Outline Number	21-0140
Land Pattern Number	90-0016
Thermal Resistance, Single-Layer Board:	
Junction to Ambient (θ _{JA})	45°C/W
Junction to Case (θ _{JC})	2°C/W
Thermal Resistance, Four-Layer Board:	
Junction to Ambient (θ _{JA})	28°C/W
Junction to Case (θ _{JC})	2°C/W

For the latest package outline information and land patterns (footprints), go to <u>www.maximintegrated.com/packages</u>. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to www.maximintegrated.com/thermal-tutorial.

Electrical Characteristics—Global Resources

 $(V_{SYS} = 3.6V, V_{IO} = 1.8V, T_A = -40^{\circ}C$ to +85°C, limits are 100% tested at $T_A = +25^{\circ}C$. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	
SUPPLY CURRENT							
OFF State Quiescent Current	lqsys_off	Vsysuvlo < Vsys < Vsys_Reset (rising), OTP_INT_PU = 1, all regulators are disabled. This includes any central bias currents disabled (EREG_EN1 pulled to Vsys)		86	135	μА	
DEVSLP State Quiescent Current		V _{SYS} = 3.3V, V _{SYS} > V _{SYS_RESET} , OTP_INT_PU = 0, PMIC in DEVSLP State, Buck2, Buck3, LDO enabled in low-power mode. No load on all regulators. All other regulators disabled		70	125		
	QSYS_DEVSLP	V _{SYS} = 5V, V _{SYS} > V _{SYS} RESET, OTP_INT_PU = 0, PMIC in DEVSLP state, Buck2, Buck3, LDO enabled in low-power mode. No load on all regulators. All other regulators disabled		90	155	μΑ	
Buck Quiescent Supply Current	I _{QSYS_BUCK}	V _{SYS} = 5V, V _{SYS} > V _{SYS_RESET} , all bucks enabled in normal-power mode and skip mode		233	420	μA	

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PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
BIAS AND REFERENCE CU	RRENT GENERA	ATOR				
Operating Voltage Range	V _{SYS}		2.6		5.5	V
Quiescent Supply Current	I _{QCBRG}	V _{SYS} > V _{SYSUVLO} (rising)		25		μA
Shutdown Supply Current		V _{SYS} < V _{SYSUVLO} (falling)		0.1		μA
Bias Enable time	^t BIASOK			100		μs
POR COMPARATOR (INTER	RNAL)					
Quiescent Supply Current	I _{QSYS_POR}			1		μA
POR Undervoltage-Lockout Threshold	V _{POR}	V _{SYS} falling		1.33		V
POR Threshold Hysteresis	V _{HYS_POR}	V _{SYS} rising		160		mV
Response Time		100mV overdrive		300		μs
DOD to UVI O Dolov		V _{SYS} rising across POR (1V to 2V)		100		
POR to UVLO Delay	^t PORUVLO	V _{SYS} falling across POR		50		μs
SYS UNDERVOLTAGE-LOC	KOUT COMPAR	ATOR				
Quiescent Supply Current	IQSYS_UVLO			1		μA
SYS Undervoltage-Lockout Threshold	V _{SYSUVLO}	V _{SYS} falling	2.00	2.10	2.25	V
SYS Undervoltage-Lockout Hysteresis	V _{INUVLO_HYS}			400		mV
SYS Undervoltage-Lockout Response Time	tsysuvlo	100mV overdrive, falling edge		150		μs
SYS RESET COMPARATOR						
Quiescent Supply Current	I _{QSYS_RESET}			3		μA
Reset Falling Threshold Range	V _{SYS_RESET}	Programmed by SYSRST[3:0]	2650		4150	mV
Reset Threshold Step Size				100		mV
Reset Threshold Hysteresis Range	VSYSRESET_ HYS	Programmed by SYSRSTHYS[1:0]	150		300	mV
Reset Threshold Hysteresis Step Size				50		mV
Reset Comparator Response Time	^t sysreset			5		μs
Reset Comparator Accuracy		SYSRSTTH[3:0] = 0x0, 0x1, 0x5, 0xA, 0xF	-2.5		+2.5	%

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PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
SYS BROWNOUT COMPAR	ATOR					
Brownout Falling Threshold Range	V _{SYS_BO}	Programmed by SYSBOTH[3:0]	2800		4300	mV
Brownout Threshold Step Size				100		mV
Brownout Threshold Hysteresis Range	V _{SYS_BO_HYS}	Programmed by SYSBOHYS[1:0]	150		300	mV
Brownout Threshold Hysteresis Step Size				50		mV
		SYS_BO_PR[1:0] = 0b00 (fast), PMIC not in DEVSLP state, 100mV under-drive with falling slew rate of 150mV/µs		1.04		
Brownout Comparator Response Time		SYS_BO_PR[1:0] = 0b01 (med-fast), PMIC not in DEVSLP state, 100mV under-drive with falling slew rate of 150mV/µs		1.14		
	^t sysbo	SYS_BO_PR[1:0] = 0b10 (med-slow), PMIC not in DEVSLP state, 100mV under-drive with falling slew rate of 150mV/µs		1.30		- µs
		SYS_BO_PR[1:0] = 0b11 (slow), PMIC not in DEVSLP state, 100mV under-drive with falling slew rate of 150mV/µs		1.68		
Brownout Comparator Response Time (DEVSLP)	t _{SYSBO}	PMIC in DEVSLP state, 100mV under-drive with falling slew rate of 150mV/µs		3.53		μs
		SYS_BO_PR[1:0] = 0b00 (fast), PMIC not in DEVSLP state		13.4		
Quiescent Supply Current	lanua an	SYS_BO_PR[1:0] = 0b01 (med-fast), PMIC not in DEVSLP state		10.4		- μΑ
Quiescent Supply Surrent	IQSYS_BO	SYS_BO_PR[1:0] = 0b10 (med-slow), PMIC not in DEVSLP state		7.4		μΑ.
		SYS_BO_PR[1:0] = 0b11 (slow), PMIC not in DEVSLP state		4.4		
Quiescent Supply Current (DEVSLP)	I _{QSYS_BO}	PMIC in DEVSLP state		1.3		μA
Brownout Comparator Accuracy		SYSBO[3:0] = 0x0, 0x1, 0x5, 0xA, 0xF, PMIC is not in DEVSLP state	-2.5		+2.5	%
Brownout Comparator Accuracy (DEVSLP)		SYSBO[3:0] = 0x0, 0x1, 0x5, 0xA, 0xF, PMIC is in DEVSLP state	-2.5		+2.5	%
Brownout Timer Period	t _{BO}	T_BO_EN = 1		100		ms

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PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
OSCILLATOR						
Clock Frequency	CLK32K	V _{SYS} = 5V		31.5		kHz
0 11 1 7 1		V _{SYS} = 3.3V	-10		+10	0,
Oscillator Tolerance		V _{SYS} = 5V	-10		+10	- %
WP_L OUTPUT (OPEN DRA	IN)					
WP_L Output-Voltage Low	V _{OL}	I _{SINK} = 2mA			0.4	V
WP_L Open Leakage		V _{SYS} = V _{WP_L} = 5.5V, T _A = +25°C, OTP_INT_PU[0] = 0b0		0.001	1	
Current		V _{SYS} = V _{WP_L} = 5.5V, T _A = +85°C, OTP_INT_PU[0] = 0b0		0.01		- μΑ
WP_L Falling Edge Time		$C_{WP_L} = 25pF, V_{WP_L} = 1.8V \ge 0$		25		ns
		WP_L_DLY[1:0] = 0b00 (based on an internal 31.5kHz clock)		0		
WP_L Output Deassert	t	WP_L_DLY[1:0] = 0b01 (based on an internal 31.5kHz clock)		254		
Delay Time	^t WPDLY	WP_L_DLY[1:0] = 0b10 (based on an internal 31.5kHz clock)		508		- µs
		WP_L_DLY[1:0] = 0b11 (based on an internal 31.5kHz clock)		1016		
WP_L Output Assert Delay Time				0		μs
WP_L Pullup Resistance	R _{PU_WP_L}	Pulled up to V _{IN_VIO} , OTP_INT_PU[0] = 0b1	50	100	170	kΩ
RESET_L OUTPUT (OPEN I	DRAIN)					
RESET_L Output-Voltage Low	V _{OL}	I _{SINK} = 2mA			0.4	V
RESET_L Open Leakage		V _{SYS} = V _{RESET_L} = 5.5V, T _A = +25°C, OTP_INT_PU[0] = 0b0		0.001	1	,
Current		V _{SYS} = V _{RESET_L} = 5.5V, T _A = +85°C, OTP_INT_PU[0] = 0b0		0.01		- μΑ
RESET_L Falling Edge Time		$C_{RESET_L} = 25pF, V_{RESET_L}$ falling from 1.8V ≥ 0		25		ns

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PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
RESET L Output Deassert		RST_L_DLY[1:0] = 0b00 (based on an internal 31.5kHz clock)		0		
	toorny	RST_L_DLY[1:0] = 0b01 (based on an internal 31.5kHz clock)		254		116
Delay Time	^t RSTDLY	RST_L_DLY[1:0] = 0b10 (based on an internal 31.5kHz clock)		508		μs
		RST_L_DLY[1:0] = 0b11 (based on an internal 31.5kHz clock)		1016		
RESET_L Output Assert Delay Time				0		μs
RESET_L Pullup Resistance	R _{PU_RESET_L}	Pulled up to V _{IN_VIO} , OTP_INT_PU[0] = 0b1	50	100	170	kΩ
PGOOD OUTPUT (OPEN DI	RAIN)			-		
PGOOD Output-Voltage Low	V _{OL}	I _{SINK} = 2mA			0.4	V
PGOOD Open Leakage		V _{SYS} = V _{PGOOD} = 5.5V, T _A = +25°C, OTP_INT_PU[0] = 0b0		0.001	1	
Current		V _{SYS} = V _{PGOOD} = 5.5V, T _A = +85°C, OTP_INT_PU[0] = 0b0		0.01		μA
PGOOD Falling Edge Time		C _{PGOOD} = 25pF, V _{PGOOD} = 1.8V ≥ 0		25		ns
		PG_DLY[1:0] = 0b00 (based on an internal 31.5kHz clock)		31.5		
PGOOD Output Assert	t	PG_DLY[1:0] = 0b01 (based on an internal 31.5kHz clock)		254		110
Delay Time	^t PGOODDLY	PG_DLY[1:0] = 0b10 (based on an internal 31.5kHz clock)		508		μs
		PG_DLY[1:0] = 0b11 (based on an internal 31.5kHz clock)		1016		
PGOOD Output Deassert Delay Time				0		μs
PGOOD Pullup Resistance	R _{PU_PGOOD}	Pulled up to V _{IN_VIO} , OTP_INT_PU[0] = 0b1	50	100	170	kΩ

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PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
LP_MODE INPUT						
LP_MODE I/O Pad Operating Voltage	V _{SYS}		2.6		5.5	V
LP_MODE Input-Low Voltage	V _{IL}				0.4	V
LP_MODE Input-High Voltage	V _{IH}		1.4			V
LP_MODE Input Hysteresis	V _{HYS}			50		mV
LP_MODE Input Leakage		V _{SYS} = V _{IN_VIO} = 5.5V, V _{LP_MODE} = 0V and 5.5V, T _A = +25°C		0.001	1	μA
Current		V _{SYS} = V _{IN_VIO} = 5.5V, V _{LP_MODE} = 0V and 5.5V, T _A = +85°C		0.01		μΑ
LP_MODE Debounce	t _{LPMD_DBNC}	Debounce applies to rising and falling edge. Does not account for oscillator tolerance (Note 4)		95	127	μs
LP_MODE I/O Pad Undervoltage Lockout	V _{SYSUVLO}	V _{SYS} falling		2.1		V
LP_MODE Mask Deassertion Timer	t _{LPMD_MSK}		16	20	25	ms
LP_ACK INPUT						
I/O Pad Operating Voltage	V _{SYS}		2.6		5.5	V
Input Low Voltage	V _{IL}				0.4	V
Input High Voltage	V _{IH}		1.4			V
Input Hysteresis	V _{HYS}			50		mV
Input Leakage Current		$V_{SYS} = 5.5V$, $V_{LP_ACK} = 0V$ and 5.5V, $T_A = +25^{\circ}C$, $OTP_INT_PU[0] = 0b0$		0.001	1	
		V _{SYS} = 5.5V, V _{LP_ACK} = 0V and 5.5V, T _A = +85°C, OTP_INT_PU[0] = 0b0		0.01		- μΑ
LP_ACK Pullup Resistance	R _{PU_LP_ACK}	Pulled up to V _{IN_VIO} , OTP_INT_PU[0] = 0b1	50	100	170	kΩ

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PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
LP_REQ OUTPUT (OPEN D	RAIN)			-		
LP_REQ Output Voltage Low	V _{OL}	I _{SINK} = 2mA			0.4	V
LP_REQ Open Leakage		V _{SYS} = V _{LP_REQ} = 5.5V, T _A = +25°C, OTP_INT_PU[0] = 0b0		0.001	1	
Current		V _{SYS} = V _{LP_REQ} = 5.5V, T _A = +85°C, OTP_INT_PU[0] = 0b0		0.01		- μΑ
LP_REQ Falling Edge Time		$C_{LP_REQ} = 25pF, V_{LP_REQ} = 1.8V \ge 0$		25		ns
12.250.2		LP_REQ_T_EN = 0, PMIC in master mode (OTP_SLP_MSTRSLV = 0), applies during DevSlp exit sequence		31.75		μs
LP_REQ Delay	tLPREQ_LOW	LP_REQ_T_EN = 1, PMIC in master mode (OTP_SLP_MSTRSLV = 0), applies during DevSlp exit sequence		20		ms
LP_REQ Pullup Resistance	R _{PU_LP_REQ}	Pulled up to V _{IN_VIO} , OTP_INT_PU[0] = 0b1	50	100	170	kΩ
EREG_ENx OUTPUT (OPEN	DRAIN)					
EREG_EN1 Output-Voltage Low	V _{OL}	I _{SINK} = 2mA			0.4	V
EREG_EN2 Output-Voltage Low	V _{OL}	I _{SINK} = 10mA			0.4	V
EREG_ENx Open		V _{SYS} = V _{EREG_ENx} = 5.5V, T _A = +25°C, OTP_INT_PU[0] = 0b0		0.001	1	
Leakage Current		V _{SYS} = V _{EREG_ENx} = 5.5V, T _A = +85°C, OTP_INT_PU[0] = 0b0		0.01		- μΑ
EREG_ENx Falling Edge Time		C _{EREG_ENx} = 25pF, V _{EREG_ENx} = 1.8V ≥ 0		25		ns
EREG_EN1 Pullup Resistance	R _{PU_EREG_} ENx	Pulled up to V _{H_INT} , OTP_INT_PU[0] = 0b1	50	100	170	kΩ
EREG_EN2 Pullup Resistance	R _{PU_EREG_} ENx	Pulled up to V _{IN_VIO,} OTP_INT_PU[0] = 0b1	50	100	170	kΩ
EREG_POK INPUT	l					
I/O Pad Operating Voltage	V _{SYS}		2.6		5.5	V
Input Low Voltage	V _{IL}				0.4	V
Input High Voltage	V _{IH}		1.4			V
Input Hysteresis	V _{HYS}			50		mV

 $(V_{SYS} = 3.6V, V_{IO} = 1.8V, T_A = -40^{\circ}C$ to +85°C, limits are 100% tested at $T_A = +25^{\circ}C$. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Input Lankage Current		$V_{SYS} = 5.5V$, $V_{EREG_POK} = 0V$ and 5.5V, $T_A = +25^{\circ}C$, $OTP_INT_PU[0] = 0b0$		0.001	1	
Input Leakage Current		$V_{SYS} = 5.5V$, $V_{EREG_POK} = 0V$ and 5.5V, $T_A = +85^{\circ}C$, $OTP_INT_PU[0] = 0b0$		0.01		- μΑ
EREG_POK Pullup Resistance	R _{PU_EREG_} POK	Pulled up to V _{IN_VIO} , OTP_INT_PU[0] = 0b1	50	100	170	kΩ
THERMAL MONITORS						
Quiescent Supply Current	I _{QTM}			1.5		μA
Shutdown Supply Current				0.1		μA
Thermal Overload	T _{JOVLD}	T _J rising, 15°C hysteresis		165		°C
Response Time		5°C overdrive		10		μs
FLEXIBLE POWER SEQUE	NCER					
Power-Up Sequence Enable Delay	t _{FPSDON}	Measured from internal FPSxEN = 1 to start of sequence (based on a 31.5kHz clock)		63.492		μs
Power-Down Sequence Enable Delay	^t FPSDOFF	Measured from internal FPSxEN = 0 to start of sequence (based on a 31.5kHz clock)		95.240		μs
		MSTRxUPF[2:0] = MSTRxDNF[2:0] = 0b000		31		
		MSTRxUPF[2:0] = MSTRxDNF[2:0] = 0b001		63		
		MSTRxUPF[2:0] = MSTRxDNF[2:0] = 0b010		127		
Flexible Power Sequencer	t _{FPS PU,}	MSTRxUPF[2:0] = MSTRxDNF[2:0] = 0b011		253		
Event Period	t _{FPS_PD}	MSTRxUPF[2:0] = MSTRxDNF[2:0] = 0b100		508		μs
		MSTRxUPF[2:0] = MSTRxDNF[2:0] = 0b101		984		
		MSTRxUPF[2:0] = MSTRxDNF[2:0] = 0b110		1936		
		MSTRxUPF[2:0] = MSTRxDNF[2:0] = 0b111		3904		

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PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
		PD_DLY[1:0] = 0b00		0		
Power-Down Sequence		PD_DLY[1:0] = 0b01		1.0		
Delay	t _{PD_DLY}	PD_DLY[1:0] = 0b10		1.5		ms
		PD_DLY[1:0] = 0b11		2.0		
BLD_IO						
Maximum Bleed Time	t _{BLEED_MAX}			20	22	ms
Minimum Bleed Time	t _{BLEED_MIN}			31.5		μs
Bleed Threshold		BLD_IO falling		90	100	mV
Bleed Resistance	R _{BLEED}	BLD_IO = 0.3V		20	27	Ω
BLD_IO Input Leakage		$V_{SYS} = 5.5V$, $V_{BLD_IO} = 0V$ and 5.5V, $T_A = +85$ °C		0.01		
Current		V_{SYS} = 5.5V, V_{BLD_IO} = 0V and 5.5V, T_A = +25°C		0.001	1	μA
ON/OFF CONTROLLER						
Hiccup Counter Limit	HICCUP_ CNT_LIM			7		counts
IN_PHUP						•
Operating Voltage Range	V _{IN_PHUP}		2.4		5.5	V
IN_PHUP Supply Current	I _{IN_PHUP}	V _{SYS} = V _{IN_PHUP} = 5.5V, T _A = +25°C		5.0		μΑ

Note 4: The LP_MODE debounce period has a variation due to the variability associated with quantizing an asynchronous input signal. Additionally, while measuring the period from a valid LP_MODE edge to a subsequent event, such as LP_REQ assertion, there is one more clock cycle (CLK32K) of delay observed in a real system.

Electrical Characteristics—Inrush Control

 $(V_{IN_SNS} = 5.0V, limits are 100\% tested at T_A = +25^{\circ}C.$ Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization.)

PARAMETER	SYMBOL		CONDITIONS	MIN	TYP	MAX	UNITS
POWER SUPPLY	•						
Supply Voltage Range	V _{IN}			2.1		5.5	V
IN Undervoltage-Lockout Threshold	V _{INUVLO}	V _{IN} rising			2.3	2.55	V
IN Undervoltage-Lockout Hysteresis	VINUVLO_HYS				200		mV
IN Undervoltage-Lockout Response Time	t _{INUVLO}	V _{IN} rising	(V _{IN} = V _{INUVLO} + 100mV)		39		μs
IN Overvoltage-Lockout Threshold	V _{INOVLO}	V _{IN} rising		5.70	5.87	6.10	V
IN Overvoltage-Lockout Hysteresis	V _{INOVLO_HYS}				80		mV
IN Overvoltage-Lockout Response Time	t _{INOVLO}	V _{IN} rising	$(V_{IN} = V_{INOVLO} + 50mV)$		8		μs
Leakage	luce van pro	$V_{IN} = 5.5V$ $T_A = +25^{\circ}C$	7, V _{IN_DRV} = 0V and 11V,		0.001	1	μA
	ILKG_VIN_DRV	$V_{IN} = 5.5V,$ $T_A = +85^{\circ}C$, V _{IN_DRV} = 0V and 11V ,		0.01		
	^I Q_IN_SS	start state)	/INR_OUT < V _{IN_SNS} (soft- , OTP_GDRV_FREQ =)kHz), V _{IN_SNS} = 3.3V		85		
Supply Current (Soft-Start)		(soft-start	/INR_OUT < V _{IN_SNS} state), OTP_GDRV_FREQ = DkHz), V _{IN_SNS} = 5V		138		μА
Supply Current (Steady-		(steady sta	/INR_OUT = V _{IN_SNS} ate), t _{SS_DONE} expired, 2.5kHz, V _{IN_SNS} = 3.3V		26		
State)	I _{IN}	(steady sta	VINR_OUT = V _{IN_SNS} ate), t _{SS_DONE} expired, 2.5kHz, V _{IN_SNS} = 5V		37		μА
NMOS SWITCH DRIVER							
Gate Drive ON Voltage	V _{IN_DRV_ON}	V _{IN} = 5V	Voltage with respect to ground when external MOSFET is being driven to it's fully ON state	8.5		11	V
Gate Drive Current	I _{GDRV_INRUSH}	V _{IN} = 3.3V setting	, 1X gate drive frequency	1.8	3.0	4.2	μΑ
4x Gate Drive Oscillator Frequency	fGDRV_4X	OTP_INR_ V _{IN} = 3.3V	FREQ[2:0] = 0b111, ', V _{IN} = 5V		720		kHz

Electrical Characteristics—Inrush Control (continued)

 $(V_{IN_SNS} = 5.0V, limits are 100\% tested at T_A = +25^{\circ}C.$ Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
2x Gate Drive Oscillator Frequency	fGDRV_2X	OTP_INR_FREQ[2:0] = 0b110, V _{IN} = 3.3V, V _{IN} = 5V		360		kHz
1x Gate Drive Oscillator Frequency	f _{GDRV_1X}	OTP_INR_FREQ[2:0] = 0b101 (nominal gate drive strength), V _{IN} = 3.3V, V _{IN} = 5V	120	180	240	kHz
0.5x Gate Drive Oscillator Frequency	fGDRV_0.5X	OTP_INR_FREQ[2:0] = 0b100, V _{IN} = 3.3V, V _{IN} = 5V		90		kHz
0.25x Gate Drive Oscillator Frequency	f _{GDRV_0.25} X	OTP_INR_FREQ[2:0] = 0b011, V _{IN} = 3.3V, V _{IN} = 5V		45		kHz
0.125x Gate Drive Oscillator Frequency	fGDRV_0.125X	OTP_INR_FREQ[2:0] = 0b010, V _{IN} = 3.3V, V _{IN} = 5V	15	23	32	kHz
0.0625x Gate Drive Oscillator Frequency	f _{GDRV_0.0625} X	OTP_INR_FREQ[2:0] = 0b001, V _{IN} = 3.3V, V _{IN} = 5V		11.25		kHz
0.03125x Gate Drive Oscillator Frequency	fGDRV_0.03125X	OTP_INR_FREQ[2:0] = 0b000, V _{IN} = 3.3V, V _{IN} = 5V		5.625		kHz
Gate Drive Discharge Resistance		Resistance from INR_DRV to INR_OUT, V _{INR_DRV-INR_OUT} = 4V		74		
	R _{GDRV_DIS}	Resistance from INR_DRV to INR_OUT, V _{INR_DRV-INR_OUT} = 3.3V		100		Ω
TIMING	1					
Start-Up Delay	t _{EN_INRUSH}	Time from V _{IN} rising above V _{INUVLO} to the internal charge pump being enabled. Duration is based on the gate drive oscillator frequency (f _{GDRV}) selected by OTP_INR_FREQ[2:0]		128		cycles of fGDRV
Soft-Start Done Time	^t ss_1	Duration from MOSFET drive circuit being enabled (subsequent to startup delay) to the point when the IN_SS_DONE (internal signal) is asserted allowing a power-up sequence to occur. Based on default gate drive frequency (f _{GDRV}) selected by OTP_INR_FREQ[2:0]		512		cycles of fGDRV
Gate Drive Idle Time	^t ss_done	Duration from MOSFET drive circuit being enabled (subsequent to the startup delay) to the point when the gate drive oscillator frequency folds back to the 12.5kHz setting (idle gate drive). Based on default gate drive frequency (f _{GDRV}) selected by OTP_INR_FREQ[2:0]		1024		cycles of fGDRV

Electrical Characteristics—Current Sense Amplifier

 $(V_{SYS} = 3.3V, C_{LOAD} = 10pF, T_A = -40^{\circ}C$ to +85°C, limits are 100% tested at $T_A = +25^{\circ}C$. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS				
INPUT OVERCURRENT										
Input Overcurrent Threshold		2.25A setting, V _{SYS} = 3.3V	-6.5		+6.5	%				
CSA Debounce Timer		OTP_CSA_DBNC = 0		100						
		OTP_CSA_DBNC = 1		50		μs				
Overcurrent-Sense Comparator Threshold 1	V _{OC_THR}	Overcurrent limit, CSTH_OPT[1:0] = 0b00		30		mV				
Overcurrent-Sense Comparator Threshold 2	V _{OC_THR}	Overcurrent limit, CSTH_OPT[1:0] = 0b01		35		mV				
Overcurrent-Sense Comparator Threshold 3	V _{OC_THR}	Overcurrent-limit, CSTH_OPT[1:0] = 0b10		40		mV				
Overcurrent-Sense Comparator Threshold 4	V _{OC_THR}	Overcurrent limit, CSTH_OPT[1:0] = 0b11		45		mV				

Electrical Characteristics—Buck Regulators (BUCK1/2 - 2A Output)

 $(V_{SYS} = 3.6V, T_A = -40^{\circ}C \text{ to } +85^{\circ}C, \text{ limits are } 100\% \text{ tested at } T_A = +25^{\circ}C. \text{ Limits over the operating temperature range are guaranteed by design and characterization.)}$

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
SUPPLY VOLTAGE AND CU	IRRENT					
Input Voltage Range	V _{INBx}		2.6		5.5	V
Shutdown Supply Current	IQSHDN_BUCKx	(Note 5)		0.1		μΑ
	IQ SKIP NM	No switching, no load, (Note 6), V _{SYS} = 3.3V		19	30	
	BUCKx	No switching, no load, (Note 6), V _{SYS} = 5V		19	30	μA
Supply Ouisseent Current	I _{Q_FPWM} _ BUCKx	FPWM mode (switching at fixed frequency), no load, V _{SYS} = 3.3V		10		Λ
Supply Quiescent Current		FPWM mode (switching at fixed frequency), no load, V _{SYS} = 5V		10		· mA
	IQ_SKIP_LPM_ BUCKx	Low-power mode (no switching), no load, (Note 6), V _{SYS} = 3.3V		5	9	^
		Low-power mode (no switching), no load, (Note 6), V _{SYS} = 5V		5	9	- μΑ

 $(V_{SYS} = 3.6V, T_A = -40^{\circ}C \text{ to } +85^{\circ}C, \text{ limits are } 100\% \text{ tested at } T_A = +25^{\circ}C. \text{ Limits over the operating temperature range are guaranteed by design and characterization.)}$

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
OUTPUT VOLTAGE						
Output Voltage Range	V _{OUT_BUCK} x	Programmable in 6.25mV steps with BUCK1VOUT[7:0] and BUCK2VOUT[7:0]	0.600		2.194	V
	VOUT_ACC_NM_ BUCKx	FPWM mode, normal mode, no load, V _{OUT_BUCK1} = 1.800V	-2		+2	
Output Voltage Accuracy	VOUT_ACC_ LPM_BUCKx	Low-power mode, no load, VOUT_BUCK1 = 1.800V	-4		+4	%
Output Voltage Accuracy	VOUT_ACC_NM_ BUCKx	FPWM mode, normal mode, no load, V _{OUT_BUCK2} = 1.200V	-2		+2	70
	V _{OUT_ACC_} LPM_BUCKx	Low-power mode, no load, V _{OUT_BUCK2} = 1.200V	-4		+4	
OUTPUT CURRENT						
Maximum Output Current	IOUT_MAX_ NM_BUCKx	RMS, normal mode, L = 1µH	2000			mA
	IOUT_MAX_ LPM_BUCKx	RMS, low-power mode, L = 1µH		10		
PMOS Peak Current Limit	1	V _{SYS} = 3.6V	2300	2875	4200	mA
	I _{LIMP}	V _{SYS} = 5V	2300	2875	4200	MA
NMOS Valley Current Limit	1	V _{SYS} = 3.6V		2125		mA
NWOS valley Current Limit	I _{LIMV}	V _{SYS} = 5V		2125		ША
NMOS Negative Current	1	V _{SYS} = 3.6V		800		mΛ
Limit	I _{LIMN}	V _{SYS} = 5V		800		mA
PERFORMANCE PARAMET	TERS					
Line Regulation		$V_{SYS} = V_{INBx} = 2.6V \text{ to } 5.5V$		0.2		%/V
Load Regulation		Load = 0 to 1A, FPWM mode		0.125		%/A
Load Transient Response		FPWM mode, V_{OUT_BUCKX} = default, L = 1µH, C_{OUT} = 12µF effective Δ I _{OUT} = 0.2A–2A, Δ t = 3µs		88		mV
Load Hansiell Nesponse		Skip mode, V_{OUT_BUCKX} = default, L = 1µH, C_{OUT} = $\overline{12}$ µF effective Δ I _{OUT} = 10mA to 0.7A, Δ t = 3µs,		90		IIIV
Switching Frequency	f _{SW}	V _{SYS} = 3.3V	1.8	2	2.2	MHz
Dead Time	t _{DEAD}	V _{SYS} = 3.3V		2.0		ns
Switching Frequency	f _{SW}	V _{SYS} = 5V	1.8	2	2.2	MHz

 $(V_{SYS} = 3.6V, T_A = -40^{\circ}C \text{ to } +85^{\circ}C, \text{ limits are } 100\% \text{ tested at } T_A = +25^{\circ}C. \text{ Limits over the operating temperature range are guaranteed by design and characterization.)}$

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Dead Time	t _{DEAD}	V _{SYS} = 5V		2.0		ns
Soft-Start Slew Rate		Fixed for buck 1		6.5		m\//uo
Soil-Start Siew Nate		Fixed for buck 2		17		mV/μs
Output Voltage Ramp-Up Slew Rate		Fixed for buckx (Notes 5, 8, 9), C _{OUT} = 22µF		40		mV/μs
Output Voltage Ramp-Down Slew Rate		Fixed for buck 1, 2 (Notes 5, 8), C _{OUT} = 22µF, BUCKxFPWMEN = 1 (x = 1, 2), no load		18		mV/μs
PMOS ON Resistance	В	V _{SYS} = V _{INBUCKx} = 3.6V, I _{OUT} = 150mA		100	150	~ 0
	R _{ON_PCH}	$V_{SYS} = V_{INBUCKx} = 5V,$ $I_{OUT} = 150mA$		100	150	mΩ
NMOS ON Resistance	D	$V_{SYS} = V_{INBUCKx} = 3.6V,$ $I_{OUT} = 150 \text{mA}$		60	100	mΩ
	R _{ON_NCH}	$V_{SYS} = V_{INBUCKx} = 5V,$ $I_{OUT} = 150mA$		60	100	11152
NMOS Zero-Crossing Threshold	I _{ZX}	Threshold to determine transition from PWM to SKIP mode		20		mA
Output Voltage Ripple in Skip Mode		V _{OUT_BUCKx} = 1.0V, L = 1μH, C _{OUT} = 12μF effective, no load (Note 5)		40		mV _{P-P}
Output Voltage Ripple in PWM Mode		V _{OUT_BUCKx} = 1.0V, L = 1µH, C _{OUT} = 12µF effective, I _{LOAD} = 0.5 x I _{OUT_MAX_BUCKx} (Note 5)		5		mV _{P-P}
	I _{L_LX_25C}	V _{LXx} = 5.5V or 0V, T _A = +25°C		0.1	1	
LX Leakage	IL_LX_85C	V _{LXx} = 5.5V or 0V, T _A = +85°C (Note 5)		1		μA
Output Active Discharge Resistance	R _{DISCHG} _ BUCKx	Resistance from FBBx to PGNDx, output disabled, (Note 7)		100		Ω
Nominal Output Inductance	L _{NOM}			1.0		μH
Minimum Effective Output Capacitance	C _{OUT_EFF_MIN}	0mA < I _{OUT} < 2000mA	18			μF
Light Load Efficiency	Eff _{LIGHT}	Low-power mode, I_{OUT} = 0.5mA, V_{OUT_BUCKx} = 1.0V, L = 1 μ H, DCR _L = 50m Ω , C _{OUT} = 22 μ F (Note 5)		75		%

 $(V_{SYS} = 3.6V, T_A = -40^{\circ}C \text{ to } +85^{\circ}C, \text{ limits are } 100\% \text{ tested at } T_A = +25^{\circ}C. \text{ Limits over the operating temperature range are guaranteed by design and characterization.)}$

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Typical Load Efficiency	Eff _{IOUT_TYP}	I_{OUT} = 0.25 x $I_{OUT_MAX_BUCKx}$, V_{OUT_BUCKx} = 1.0V, L = 1 μ H, DCR _L = 50m Ω , C _{OUT} = 22 μ F (Note 5)		85		%
Maximum Load Efficiency	EFF _{IOUT} MAX	$I_{OUT} = I_{OUT_MAX_BUCKx}$, $V_{OUT} = 1.0V$, $L = 1\mu H$, DCR _L = 50mΩ, C _{OUT} = 22 μ F (Note 5)		70		%
Turn-On Delay Time	tON_DLY_BUCKx	EN signal to LX switching with bias ON		30		μs
Maximum Duty Cycle		VOUT_BUCKx / VIN_BUCKx expressed as %		90		%
BROWNOUT COMPARATOR	R					
Output Brownout		Normal-power mode, falling threshold, BUCKx_BO_THR[1:0] = 0b00		75		
		Normal-power mode, falling threshold, BUCKx_BO_THR[1:0] = 0b01		80		0,
Threshold	V _{BO_BUCKx}	Normal-power mode, falling threshold, BUCKx_BO_THR[1:0] = 0b10		85		%
		Normal-power mode, falling threshold, BUCKx_BO_THR[1:0] = 0b11		90.7		
Output Brownout Accuracy		Normal-power mode. V _{OUT_BUCKx} = 1.0V (VOUT_BUCKx[7:0] = 0 x 40)	-4.0		+4.0	%
Output Brownout Threshold (Low-Power Mode)	V _{BO_BUCKx}	Falling threshold, low-power mode		86.0		%
Output Brownout Accuracy		Low-power mode. V _{OUT_BUCKx} = 1.0V (VOUT_BUCKx[7:0] = 0 x 40)	-4		+4	%
Output Brownout Hysteresis Range	V _{BO_HYS_} BUCKx	2-bit control over I ² C. Max rising threshold limited to 96%	5		20	%
Brownout Voltage Hysteresis Programming Step Size		Programmable with BUCKx_BO_HYS[1:0]		5		%
Output Brownout Hysteresis (Low-Power Mode)	V _{BO_HYS_} BUCKx_LPM			5		%

 $(V_{SYS} = 3.6V, T_A = -40^{\circ}C$ to +85°C, limits are 100% tested at $T_A = +25^{\circ}C$. Limits over the operating temperature range are guaranteed by design and characterization.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Output Brownout Response Time		BUCKx_BO_PR[1:0] = 0b00 (fast), buck in normal-power mode, 100mV under-drive with falling slew rate of 150mV/µs. Time from V _{OUT_BUCKx} falling to PGOOD pin falling		1.04		
		BUCKx_BO_PR[1:0] = 0b01 (med- fast), buck in normal-power mode, 100mV under-drive with falling slew rate of 150mV/µs. Time from V _{OUT} _ BUCKx falling to PGOOD pin falling		1.14		
	^t BO_BUCKx	BUCKx_BO_PR[1:0] = 0b10 (med-slow), buck in normal-power mode, 100mV under-drive with falling slew rate of 150mV/µs. Time from V _{OUT} _BUCKx falling to PGOOD pin falling		1.30		μs
		BUCKx_BO_PR[1:0] = 0b11 (slow), buck in normal-power mode, 100mV under-drive with falling slew rate of 150mV/µs. Time from V _{OUT_BUCKx} falling to PGOOD pin falling		1.68		
		Buck in low-power mode, 100mV under-drive with falling slew rate of 150mV/µs. Time from V _{OUT_BUCKx} falling to PGOOD pin falling		3.18		
		Normal-power mode, BUCKx_BO_PR[1:0] = 0b00 (fast)		13.4		
	IQNM_BO_	Normal-power mode, BUCKx_BO_PR[1:0] = 0b01 (med-fast)		10.4		
Output Brownout Supply Current	BUCKx	Normal-power mode, BUCKx_BO_PR[1:0] = 0b10 (med-slow)		7.4		μΑ
		Normal-power mode, BUCKx_BO_PR[1:0] = 0b11 (slow)		4.4		
	I _{QLPM_BO_} BUCKx	Low-power mode		1.3		

 $(V_{SYS} = 3.6V, T_A = -40^{\circ}C \text{ to } +85^{\circ}C, \text{ limits are } 100\% \text{ tested at } T_A = +25^{\circ}C. \text{ Limits over the operating temperature range are guaranteed by design and characterization.)}$

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
OV COMPARATOR						
Output OV Trip Level	V _{OUTBUCKx_OV}	Rising edge, BUCKx_OV_THR = 1, referenced to output voltage setting		116.6		%
Output OV Hysteresis		BUCKx_OV_THR = 1		9.1		%
Output OV Trip Level	V _{OUTBUCKx_OV}	Rising edge, BUCKx_OV_THR = 0, referenced to output voltage setting		108.3		%
Output OV Hysteresis		BUCKx_OV_THR = 0		2.8		%
Output OV Trip Level (Low-Power Mode)	V _{OUTBUCKx_OV}	Rising edge, low-power mode		108.3		%
Output OV hysteresis (Low-Power Mode)		Low-power mode		2.8		%
Output Over-Voltage Response Time	tov_Buckx	Normal-power mode, 100mV over- drive with rising slew rate of 150mV/ µs. Time from V _{OUT_BUCKx} rising to PGOOD pin falling (Note 5)		1.68		μs
Output Over-Voltage Supply current	I _{Q_OV_BUCKx}	Normal-power mode		4.4		μA
Output Over-Voltage Response Time (Low-Power Mode)	tov_Buckx	Low-power mode, 100mV over-drive with rising slew rate of 150mV/µs. Time from V _{OUT_BUCKx} rising to PGOOD pin falling (Note 5)		3.18		μs
Output Over-Voltage Supply Current (Low-Power Mode)	IQ_OV_BUCKx	Low-power mode		1.3		μA

- Note 5: Design guidance only and is not production tested.
- Note 6: Individual buck Iq is not production tested. It is covered by a combined test by turning on all bucks.
- Note 7: There is an n-channel MOSFET in series with the output active-discharge resistance. This NMOS requires V_{SYS} > 1.2V to be enhanced.
- Note 8: The ramp down slew rate when the output voltage is decreased through I²C is a function of the negative current limit and the output capacitance. With no load, forced PWM mode and 22μF output capacitor, the ramp-down slew rate is dv/dt = i / C = 0.4A / 22μF = 18mV/μs.
- Note 9: DVS and soft-start ramp rates can be expected to vary by up to 30%.

 $(V_{SYS} = 5.0V, T_A = -40^{\circ}C \text{ to } +85^{\circ}C, \text{ Limits are } 100\% \text{ tested at } T_A = +25^{\circ}C. \text{ Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Specifications marked "GBD" are guaranteed by design and not production tested.)$

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
SUPPLY VOLTAGE AND	CURRENT					
Input Voltage Range	V _{INBUCK3}		2.6	-	5.5	V
Shutdown Supply Current	I _{QSHDN} _ BUCK3	(Note 10)		0.1		μA
	IQ_SKIP_NM_ BUCK3	No switching, no load (Note 10)		26	40	μA
Supply Quiescent Current	I _{Q_FPWM} _ BUCK3	FPWM mode, no load (Note 10)		10		mA
	I _{Q_SKIP_LPM_} BUCK3	Low-power mode (no switching), no load (Note 10)		10	19	μA
OUTPUT VOLTAGE						
Output Voltage Range	V _{OUT_BUCK3}	I ² C programmable in 10mV Steps (BUCK3VOUT[6:0] = 0x01 to 0x7F)	0.26		1.52	V
Output Voltage Accuracy	VOUT_ACC_ NM_BUCK3	FPWM mode, normal mode, no load, T _A = +25°C, V _{OUT_BUCK3} = 1.0V	-2		+2	- %
	V _{OUT_ACC_} LPM_BUCK3	Low-power mode, no load, T _A = +25°C, V _{OUT_BUCK3} = 1.000V	-4		+4	
PERFORMANCE PARAM	ETERS					
Cuitabina Francisco	£	V _{SYS} = 3.3V	1.8	2	2.2	MHz
Switching Frequency	f _{SW}	V _{SYS} = 5V	1.8	2	2.2	
Line Regulation		V _{INBUCK3} = 2.6V to 5.5V, V _{OUT_BUCK3} = 1.0V		0.2		%/V
Load Regulation		V _{OUT_BUCK3} = 1.0V, (Note 10), load = 0 to 1A, FPWM mode		0.125		%/A
Load Transient Response (Droop)		Skip mode, V_{OUT} = default, L = 1 μ H, C_{OUT} = 28 μ F effective ΔI_{OUT} = 20mA to 500mA, Δt = 0.8 μ s (Note 10)		45		mV
эропае (Бтоор)		Skip mode, V_{OUT} = default, L = 1 μ H, C_{OUT} = 28 μ F effective ΔI_{OUT} = 20mA to 3A, Δt = 4.8 μ s (Note 10)		70		
Soft-Start Slew Rate		BUCK3SSRAMP = 0		2.5		m\//uc
Suit-Start Siew Rate		BUCK3SSRAMP = 1		10		mV/µs
Output Voltage Ramp-Up/Down Slew Rate (DVS)				10		mV/µs

 $(V_{SYS} = 5.0V, T_A = -40^{\circ}C \text{ to } +85^{\circ}C$, Limits are 100% tested at $T_A = +25^{\circ}C$. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Specifications marked "GBD" are guaranteed by design and not production tested.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
DMOC ON Desisters	Б	V _{SYS} = V _{INBUCK3} = 5V, I _{OUT} = 150mA		60	90	0
PMOS ON Resistance	R _{ON_PCH}	V _{SYS} = V _{INBUCK3} = 3.6V, I _{OUT} = 150mA		60	90	mΩ
NIMOS ON Posistanos	В	V _{SYS} = V _{INBUCK3} = 5V, I _{OUT} = 150mA		35	60	0
NMOS ON Resistance	R _{ON_NCH}	V _{SYS} = V _{INBUCK3} = 3.6V, I _{OUT} = 150mA		35	60	mΩ
NMOS Zero-Crossing	I _{ZX_SKIP}	SKIP mode		20		Λ
Threshold	I _{ZX_PWM}	PWM mode		20		mA
Output Voltage Ripple In Skip Mode		V _{OUT_BUCK3} = 1.0V, L = 1μH, C _{OUT} = 28μF effective, no load (Note 10)		15		mV
Output Voltage Ripple In PWM Mode		V_{OUT_BUCK3} = 1.0V, L = 1 μ H, C_{OUT} = 28 μ F effective, I_{LOAD} = 0.5 x $I_{OUT_MAX_BUCK3}$ (Note 10)		5		mV
LX Leakage	I _{L_LX_25C}	V _{LXBUCK1} = 5.5V or 0V, T _A = +25°C		0.1	1	
	I _{L_LX_85C}	V _{LXBUCK1} = 5.5V or 0V, T _A = +85°C (Note 10)		1		μA
Output Active Discharge Resistance	R _{DISCHG} BUCK3	Resistance from FBB3 to PGND3, output disabled		100		Ω
Nominal Output Inductance	L _{NOM}			1.0		μΗ
Minimum Effective Output Capacitance	C _{OUT_EFF} _	0mA < I _{OUT} < 3000mA		28		μF
Turn-On Delay Time	^t ON_DLY_ BUCK1	EN signal to LX switching with bias ON		200		μs
Light Load Efficiency	Eff _{LIGHT}	Low-power mode, I_{OUT} = 0.5mA, V_{OUT_BUCKx} = 1.0V, L = 1 μ H, DCR _L = 50m Ω , C _{OUT} = 3 x 22 μ F (Note 10)		75		%
Typical Load Efficiency	Eff _{IOUT_TYP}	I_{OUT} = 0.25 x $I_{OUT_MAX_BUCKx}$, V_{OUT_BUCKx} = 1.0V, L = 1 μ H, DCR _L = 50m Ω , C _{OUT} = 3 x 22 μ F (Note 10)		88		%
Maximum Load Efficiency	EFFIOUT_MAX	$I_{OUT} = I_{OUT_MAX_BUCKx}$, $V_{OUT} = 1.0V$, L = 1μH, DCR _L = $\overline{5}0m\Omega$, $C_{OUT} = 3 \times 22\mu F$ (Note 10)		77		%

 $(V_{SYS} = 5.0V, T_A = -40^{\circ}C \text{ to } +85^{\circ}C$, Limits are 100% tested at $T_A = +25^{\circ}C$. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Specifications marked "GBD" are guaranteed by design and not production tested.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
OUTPUT CURRENT						
Maximum Output	IOUT_MAX_ NM_BUCK3	RMS, normal mode	3000			mA
Current	IOUT_MAX_ LPM_BUCK3	RMS, low-power mode		10		IIIA
PMOS Peak Current Limit	I _{LIMP}	$T_A = -40$ °C to +85°C, $V_{SYS} = 3.6V$	3825	4250	4675	mA
NMOS Valley Current Limit	I _{LIMV}			3750		mA
NMOS (Negative) Current Limit	I _{LIMN}			2000		mA
BROWNOUT COMPARA	TOR					
Output Brownout		Normal-power mode, falling threshold, BUCK3_BO_THR[1:0] = 0b00		77		
		Normal-power mode, falling threshold, BUCK3_BO_THR[1:0] = 0b01		81		0,
Threshold	Vво_вискз	Normal-power mode, falling threshold, BUCK3_BO_THR[1:0] = 0b10		85.7		%
		Normal-power mode, falling threshold, BUCK3_BO_THR[1:0] = 0b11		91		
Output Brownout Accuracy		Normal-power mode. V _{OUT_BUCK3} = 1.0V (VOUT_BUCK3[7:0] = 0x4B)	-4.5		+4.5	%
Output Brownout Threshold (Low-Power Mode)	V _{BO_BUCKx}	Falling threshold, low-power mode		86.0		%
Output Brownout Accuracy		Low-power mode, V _{OUT_BUCK3} = 1.0V (VOUT_BUCK3[7:0] = 0x4B)	-4		+4	%
Output Brownout Hysteresis Range	V _{BO_HYS_} BUCKx	2-Bit control over I ² C. Max rising threshold limited to 96%	5		20	%
Brownout Voltage Hysteresis Programming Step Size		Programmable with BUCKx_BO_HYS[1:0]		5		%
Output Brownout Hysteresis (Low-Power Mode)	V _{BO_HYS_} BUCKx_LPM			5		%

 $(V_{SYS} = 5.0V, T_A = -40^{\circ}C \text{ to } +85^{\circ}C, \text{ Limits are } 100\% \text{ tested at } T_A = +25^{\circ}C. \text{ Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Specifications marked "GBD" are guaranteed by design and not production tested.)$

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Output Brownout Response Time		BUCKx_BO_PR[1:0] = 0b00 (fast), buck in normal-power mode, 100mV under-drive with falling slew rate of 150mV/µs. Time from V _{OUT_BUCKx} falling to PGOOD pin falling		1.04		
		BUCKx_BO_PR[1:0] = 0b01 (med-fast), buck in normal-power mode, 100mV under- drive with falling slew rate of 150mV/µs. Time from V _{OUT_BUCKx} falling to PGOOD pin falling		1.14		
	tBO_BUCKx	BUCKx_BO_PR[1:0] = 0b10 (med-slow), buck in normal-power mode, 100mV under-drive with falling slew rate of 150mV/µs. Time from V _{OUT_BUCKx} falling to PGOOD pin falling		1.30		μs
		BUCKx_BO_PR[1:0] = 0b11 (slow), buck in normal-power mode, 100mV under-drive with falling slew rate of 150mV/µs. Time from V _{OUT_BUCKx} falling to PGOOD pin falling		1.68		
		Buck in Low-power mode, 100mV underdrive with falling slew rate of 150mV/µs. Time from V _{OUT_BUCKx} falling to PGOOD pin falling		3.18		
		Normal-power mode, BUCKx_BO_PR[1:0] = 0b00 (fast)		13.4		
	I _{QNM_BO_}	Normal-power mode, BUCKx_BO_PR[1:0] = 0b01 (med-fast)		10.4		
Output Brownout Supply Current	BUCKx	Normal-power mode, BUCKx_BO_PR[1:0] = 0b10 (med-slow)		7.4		μA
		Normal-power mode, BUCKx_BO_PR[1:0] = 0b11 (slow)		4.4		
	IQLPM_BO_ BUCKx	Low-power mode		1.3		

 $(V_{SYS} = 5.0V, T_A = -40^{\circ}C \text{ to } +85^{\circ}C$, Limits are 100% tested at $T_A = +25^{\circ}C$. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Specifications marked "GBD" are guaranteed by design and not production tested.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
OV COMPARATOR						
Output OV Trip Level	VOUTBUCK3_ OV	Rising edge, BUCK3_OV_THR = 1		117.1		%
Output OV hysteresis		BUCK3_OV_THR = 1		8.6		%
Output OV Trip Level	V _{OUTBUCKx} _	Rising edge, BUCK3_OV_THR = 0		108.5		%
Output OV Hysteresis		BUCK3_OV_THR = 0		3.9		%
Output OV Trip Level (Low-Power Mode)	V _{OUTBUCK3} _	Rising edge, low-power mode		108.3		%
Output OV Hysteresis (Low-Power Mode)		Low-power mode		3.9		%
Output Over-Voltage Response Time	tov_вискз	Buck in normal-power mode, 100mV over-drive with rising slew rate of 150mV/µs. Time from V _{OUT_BUCK3} rising to PGOOD pin falling (Note 10)		1.68		μs
Output Over-Voltage Supply current	IQ_OV_BUCKx	Buck in normal-power mode		4.4		μΑ
Output Over-Voltage Response Time (Low-Power Mode)	tov_Buckx	Buck in low-power mode, 100mV over-drive with rising slew rate of 150mV/µs. Time from V _{OUT_BUCKx} rising to PGOOD pin falling (Note 10)		3.18		μs
Output Over-Voltage Supply current (Low-Power Mode)	IQ_OV_BUCK3	Buck in low-power mode		1.3		μΑ

Note 10: Design guidance only and is not production tested.

Electrical Characteristics—Load Switch Driver (LSW1/2)

 $(V_{SYS} = 3.6V, T_A = -40^{\circ}C \text{ to } +85^{\circ}C, \text{ Limits are } 100\% \text{ tested at } T_A = +25^{\circ}C. \text{ Limits over the operating temperature range are guaranteed by design and characterization. Specifications marked "GBD" are guaranteed by design and not production tested. x is used to represent multiple instances of similar resources, for this section x = 1, 2 unless specified for e.g., LSWx represents LSW1, LSW2.)$

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
POWER SUPPLY						
Supply Current	ISYS_LSW_SS	V _{INLSWx} = V _{SYS} = 5V, LSWxDRV_ FREQ=0b111 (800kHz), C _{OUTLSWx} = 20µF		68		μA
Supply Voltage Range	V _{SYS}		2.6		5.5	V
Supply Current		$V_{\text{INLSWx}} = V_{\text{SYS}} = 3.3V,$ LSWxDRV_FREQ = 0b111 (1.6MHz), $C_{\text{OUTLSWx}} = 20\mu\text{F}$		43		μΑ
Supply Current	lsys_lsw_ss	$V_{\text{INLSWx}} = V_{\text{SYS}} = 5V$, LSWxDRV_FREQ = 0b101 (400kHz), COUTLSWx = 20 μ F		20		
Lookogo		VSYS = 5.5V, V_{LSWx_DRV} = 0V and 11V, T_A = +25°C		0.001	1	μA
Leakage	ILKG_LSWx_DRV	V_{SYS} = 5.5V, V_{LSWx_DRV} =0 V and 11V, T_A = +85°C		0.01		μΛ
NMOS SWITCH DRIVER						•
Gate Drive Voltage	V _{LSWx_DRV}	V _{SYS} = 5V	8.5		11	V
Gate Drive Current	I _{LSWx_DRV}	V _{SYS} = 3.3V, 1X gate drive frequency setting (LSWx_DRV_FREQ[2:0] = 0b101), C _{OUTLSWx} = 20µF	1.85	3.7	5.55	μA
4x Gate Drive Oscillator Frequency	fLSWx_DRV_4X	LSWxDRV_FREQ[2:0] = 0b111, V _{SYS} = 3.3V, V _{SYS} = 5V		1600		kHz
2x Gate Drive Oscillator Frequency	fLSWx_DRV_2X	LSWx_DRV_FREQ[2:0] = 0b110, V _{SYS} = 3.3V, V _{SYS} = 5V		800		kHz
1x Gate Drive Oscillator Frequency	f _{LSWx_DRV_1X}	LSWx_DRV_FREQ[2:0] = 0b101 (nominal gate drive strength), V _{SYS} = 3.3V, V _{SYS} = 5V	200	400	600	kHz
0.5x Gate Drive Oscillator Frequency	fLSWx_DRV_0.5X	LSWx_DRV_FREQ[2:0] = 0b100, V _{SYS} = 3.3V, V _{SYS} = 5V		200		kHz
0.25x Gate Drive Oscillator Frequency	f _{LSWx} _ DRV_0.25X	LSWx_DRV_FREQ[2:0] = 0b011, V _{SYS} = 3.3V, V _{SYS} = 5V		100		kHz
0.125x Gate Drive Oscillator Frequency	fLSWx_ DRV_0.125X	LSWx_DRV_FREQ[2:0] = 0b010, V _{SYS} = 3.3V, V _{SYS} = 5V	25	50	75	kHz
0.0625x Gate Drive Oscillator Frequency	fLSWx_ DRV_0.0625X	LSWx_DRV_FREQ[2:0] = 0b001, V _{SYS} = 3.3V, V _{SYS} = 5V		25		kHz

Electrical Characteristics—Load Switch Driver (LSW1/2) (continued)

 $(V_{SYS} = 3.6V, T_A = -40^{\circ}C \text{ to } +85^{\circ}C, \text{ Limits are } 100\% \text{ tested at } T_A = +25^{\circ}C. \text{ Limits over the operating temperature range are guaranteed by design and characterization. Specifications marked "GBD" are guaranteed by design and not production tested. x is used to represent multiple instances of similar resources, for this section x = 1, 2 unless specified for e.g., LSWx represents LSW1, LSW2.)$

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
0.03125x Gate Drive Oscillator Frequency	f _{LSWx} _ DRV_0.03125X	LSWx_DRV_FREQ[2:0] = 0b000, VSYS = 3.3V, V _{SYS} = 5V		12.50		kHz
Output Active Discharge Resistance	R _{DISCHG_LSW}		50	100	150	Ω
Gate Drive Discharge	D	Resistance from LSWx_DRV to FBLSWx, VLSWx_DRV-FBLSWx = 4V		74		0
Resistance	RLSW_GDRV_DIS	Resistance from LSWx_DRV to FBLSWx, VLSWx_DRV-FBLSWx = 3.3V		100		Ω
TIMING						
Soft-Start Done Time	tss_done_lsw	Duration from MOSFET drive circuit being enabled to the internal soft-start done signal being asserted. Based on default gate drive frequency (f _{LSWx_DRV_FREQ}) selected by LSWx_DRV_FREQ[2:0] to program the default gate drive frequency		256		cycles of fLSWx_ DRV_ FREQ
POWER-OK COMPARAT	OR					
Output Power-OK Threshold	V _{LSWx_} OUT_ POK_INT	Rising edge, input to the load switch is either one of the three internal buck regulator outputs or V _{SYS} as selected by LSWx_INP_EXT	0.85 x V _{IN} _ LSWx	0.90 x V _{IN} _ LSWx	0.95 x V _{IN} _ LSWx	V
Output Power-OK Hysteresis				3		%
Power-OK Response Time		V _{SYS} = 3.3V, V _{INLSWx} = 1.8V, LSWx is enabled, 100mV under-drive with falling slew rate of 150mV/µs		1.20		μs
Power-OK Comparator Active Current	I _{Q_POK_LSWx}	V _{SYS} = 3.3V, V _{INLSWx} = 1.8V, LSWx is enabled		1.1		μА

Electrical Characteristics—Load Switch Driver (LSW3)

 $(V_{SYS} = 3.6V, T_A = -40^{\circ}C \text{ to } +85^{\circ}C, \text{ limits are } 100\% \text{ tested at } T_A = +25^{\circ}C. \text{ Limits over the operating temperature range are guaranteed}$ by design and characterization. Specifications marked "GBD" are guaranteed by design and not production tested. This section is applicable when OTP_INRUSH_DISABLE = 1 and LSW_OTP_SEL = 1.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
POWER SUPPLY						
Supply Voltage Range	V _{SYS}		2.6		5.5	V
Supply Current	lsys_Lsw_ss	$V_{\text{INLSW3}} = V_{\text{SYS}} = 3.3\text{V, OTP}_{\text{INR_FREQ[2:0]}} = 0\text{b111 (720kHz),}$ $C_{\text{OUTLSW3}} = 20\mu\text{F}$		40		· μΑ
Зирріу Сипепі		$V_{\text{INLSW3}} = V_{\text{SYS}} = 5V,$ OTP_INR_FREQ[2:0] = 0b111 (720kHz), $C_{\text{OUTLSW3}} = 20\mu\text{F}$		67		μΑ
Lookogo		V_{SYS} = 5.5V, V_{INR_DRV} = 0V and 11V, T_A = +25°C		0.001	1	
Leakage	I _{LKG_LSW3_DRV}	V_{SYS} = 5.5V, V_{INR_DRV} = 0V and 11V, T_A = +85°C		0.01		- μΑ
NMOS SWITCH DRIVER						
Gate Drive Voltage	V _{LSW3_DRV}	V _{SYS} = 5V	8.5		11	V
Gate Drive Current	I _{LSW3_DRV}	V _{SYS} = 3.3V, 1X gate drive frequency setting (OTP_INR_FREQ[2:0] = 0b101), C _{OUTLSW3} = 20µF	1.8	3	4.2	μA
4x Gate Drive Oscillator Frequency	fLSW3_DRV_4X	OTP_INR_FREQ[2:0] = 0b111, V _{SYS} = 3.3V, V _{SYS} = 5V		720		kHz
2x Gate Drive Oscillator Frequency	fLSW3_DRV_2X	OTP_INR_FREQ[2:0] = 0b110, V _{SYS} = 3.3V, V _{SYS} = 5V		360		kHz
1x Gate Drive Oscillator Frequency	fLSW3_DRV_1X	OTP_INR_FREQ[2:0] = 0b101 (nominal gate drive strength), V _{SYS} = 3.3V, V _{SYS} = 5V	120	180	240	kHz
0.5x Gate Drive Oscillator Frequency	fLSW3_DRV_0.5X	OTP_INR_FREQ[2:0] = 0b100, V _{SYS} = 3.3V, V _{SYS} = 5V		90		kHz
0.25x Gate Drive Oscillator Frequency	fLSW3_DRV_0.25X	OTP_INR_FREQ[2:0] = 0b011, V _{SYS} = 3.3V, V _{SYS} = 5V		45		kHz
0.125x Gate Drive Oscillator Frequency	fLSW3_ DRV_0.125X	OTP_INR_FREQ[2:0] = 0b010, V _{SYS} = 3.3V, V _{SYS} = 5V	15	23	32	kHz
0.0625x Gate Drive Oscillator Frequency	fLSW3_ DRV_0.0625X	OTP_INR_FREQ[2:0] = 0b001, V _{SYS} = 3.3V, V _{SYS} = 5V		11.25		kHz

Electrical Characteristics—Load Switch Driver (LSW3) (continued)

 $(V_{SYS} = 3.6V, T_A = -40^{\circ}C \text{ to } +85^{\circ}C, \text{ limits are } 100\% \text{ tested at } T_A = +25^{\circ}C. \text{ Limits over the operating temperature range are guaranteed}$ by design and characterization. Specifications marked "GBD" are guaranteed by design and not production tested. This section is applicable when OTP_INRUSH_DISABLE = 1 and LSW_OTP_SEL = 1.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
0.03125x Gate Drive Oscillator Frequency	f _{LSW3} _ DRV_0.03125X	OTP_INR_FREQ[2:0] = 0b000, V _{SYS} = 3.3V, V _{SYS} = 5V		5.625		kHz
Gate Drive Discharge	Б	Resistance from LSW3_DRV to FBLSW3, V _{LSW3_DRV-FBLSW3} = 4V		74		Ω
Resistance	RLSW_GDRV_DIS	Resistance from LSWx_DRV to FBLS- Wx, V _{LSWx_DRV-FBLSWx} = 3.3V		100	Ω	Ω
TIMING						
Soft-Start Done Time	tss_done_lsw	Duration from MOSFET drive circuit being enabled to the internal soft-start done signal being asserted. Based on default gate drive frequency (f _{LSW3_DRV_FREQ}) selected by (OTP_INR_FREQ[2:0], to program the default gate drive frequency		512		cycles of fLSW3_ DRV_ FREQ
POWER-OK COMPARAT	OR					
Output Power-OK Threshold	VLSW3_OUT_ POK_INT	Rising edge, input to the load switch is either one of the four internal buck regulator outputs or V _{SYS} as selected by LSW3_INP_EXT	0.85 x V _{IN} _ LSW3	0.90 x V _{IN} _ LSW3	0.95 x V _{IN} _ LSW3	V
Output Power-OK Hysteresis				3		%
Power-OK Response Time		V_{SYS} = 3.3V, V_{INLSWX} = 1.8V, LSWx is enabled, 100mV under-drive with falling slew rate of 150mV/ μ s		1.20		μs
Power-OK Comparator Active Current	IQ_POK_LSWx	V _{SYS} = 3.3V, V _{INLSWx} = 1.8V, LSWx is enabled		1.1		μA

Electrical Characteristics—Linear Regulator

 $(V_{SYS} = 3.6V, V_{IN_LDO} = 3.6V, V_{OUT_LDO} = 1.8V, C_{IN_LDO} = 1\mu F, C_{OUT_LDO} = 2.2\mu F$, limits are 100% production tested at $T_A = +25^{\circ}C$, limits over the operating temperature range $(T_A = -40^{\circ}C \text{ to } +85^{\circ}C)$ are guaranteed by design and characterization, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
GENERAL CHARACTER	RISTICS					
Input Voltage Range	V _{IN_LDO}	Guaranteed by Output Voltage Accuracy tests (Notes 11, 12)	1.7		V_{SYS}	V
LDO Shutdown IN_LDO Current	lin_LDO	Current measured into IN_LDO, LDO output disabled, V _{SYS} = 3.6V, V _{IN_LDO} = 3.6V. Production tested in combination with other blocks as shown in the <i>OFF State Quiescent Current</i> parameter of the <i>Electrical Characteristics—Linear Regulator</i> table		<0.1	1	μА
LDO Shutdown SYS Current	I _{SYS}	Current measured into IN_LDO, LDO output disabled, V _{SYS} = 3.6V, V _{IN_LDO} = 3.6V. Production tested in combination with other blocks as shown in the global resources "OFF State Quiescent Current" parameter		<0.1		μА
LDO Normal Mode Quiescent Supply IN_LDO Current (Not in Dropout)	I _{IN_LDO}	Normal mode of operation, current measured into IN_LDO, LDO output enabled and in regulation, V _{SYS} = 3.6V, V _{IN_LDO} = 3.6V, V _{OUT_LDO} = 2.5V, I _{OUT_LDO} = 0mA. Production tested in a combination with all other blocks as shown in the global resources "ON State Quiescent Current" parameter		16	20	μА
LDO Normal Mode Quiescent Supply SYS Current (Not in Dropout)	I _{SYS}	Normal mode of operation, current measured into IN_LDO, LDO output enabled and in regulation, V _{SYS} = 3.6V, V _{IN_LDO} = 3.6V, V _{OUT_LDO} = 2.5V, I _{OUT_LDO} = 0mA. Production tested in combination with other blocks as shown in the global resources "ON State Quiescent Current" parameter		5	7	μА
LDO Normal Mode Quiescent Supply IN_LDO Current (In Dropout)	I _{IN_LDO}	Normal mode of operation, current measured into IN_LDO, LDO output enabled and in regulation, V _{SYS} = 3.6V, V _{IN_LDO} = 2.0V, V _{OUT_LDO_TARGET} = 2.5V, I _{OUT_LDO} = 0mA (Note 13)		20		μА
LDO Normal Mode Quiescent Supply SYS Current (In Dropout)	I _{SYS}	Normal mode of operation, current measured into IN_LDO, LDO output enabled and in regulation, V _{SYS} = 3.6V, V _{IN_LDO} = 2.0V, V _{OUT_LDO_TARGET} = 2.5V, I _{OUT_LDO} = 0mA (Note 13)		5		μА

 $(V_{SYS} = 3.6V, V_{IN_LDO} = 3.6V, V_{OUT_LDO} = 1.8V, C_{IN_LDO} = 1\mu F, C_{OUT_LDO} = 2.2\mu F$, limits are 100% production tested at $T_A = +25^{\circ}C$, limits over the operating temperature range $(T_A = -40^{\circ}C \text{ to } +85^{\circ}C)$ are guaranteed by design and characterization, unless otherwise noted.)

3	μА
3	μА
	μА
	μА
	mA
	IIIA
	mA
20	μF
	V
	V
	Bits
	mV
	20

 $(V_{SYS} = 3.6V, V_{IN_LDO} = 3.6V, V_{OUT_LDO} = 1.8V, C_{IN_LDO} = 1\mu F, C_{OUT_LDO} = 2.2\mu F$, limits are 100% production tested at $T_A = +25^{\circ}C$, limits over the operating temperature range $(T_A = -40^{\circ}C \text{ to } +85^{\circ}C)$ are guaranteed by design and characterization, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	
STATIC CHARACTERIS	STICS						
		V_{SYS} = 5.5V, V_{IN_LDO} = 1.7V, V_{OUT_LDO} = 0.8V, I_{OUT_LDO} = 150mA, normal-power mode	-2		+2		
		V_{SYS} = 5.5V, V_{IN_LDO} = 1.7V, V_{OUT_LDO} = 0.8V, I_{OUT_LDO} = 5mA, low-power mode	-6.5		+6.5		
Output Voltage Accuracy		V_{SYS} = 2.8V, V_{IN_LDO} = 2.8V, V_{OUT_LDO} = 2.5V, I_{OUT_LDO} = 150mA, normal-power mode	-2		+2		
		V_{SYS} = 2.8V, V_{IN_LDO} = 2.8V, V_{OUT_LDO} = 2.5V, I_{OUT_LDO} = 5mA, low-power mode	-6.5		+6.5	%	
		V_{SYS} = 5.5V, V_{IN_LDO} = 5.5V, V_{OUT_LDO} = 3.975V, I_{OUT_LDO} = 0.1mA, normal-power mode	-2		+2		
		V_{SYS} = 5.5V, V_{IN_LDO} = 5.5V, V_{OUT_LDO} = 3.975V, I_{OUT_LDO} = 0.1mA, low-power mode	-6.5		+6.5		
		V_{SYS} = 5.5V, V_{IN_LDO} = 1.7V to 5.5V, V_{OUT_LDO} = 0.8V, I_{OUT_LDO} = 0.1mA, normal-power mode		0.05			
		V_{SYS} = 5.5V, V_{IN_LDO} = 1.7V to 5.5V, V_{OUT_LDO} = 0.8V, I_{OUT_LDO} = 0.1mA, low-power mode		0.05		0/ /\ /	
Line Regulation		V_{SYS} = 5.5V, V_{IN_LDO} = 4.4V to 5.5V, V_{OUT_LDO} = 3.975V, I_{OUT_LDO} = 0.1mA, normal-power mode		0.05		- %/V	
		V_{SYS} = 5.5V, V_{IN_LDO} = 4.4V to 5.5V, V_{OUT_LDO} = 3.975V, I_{OUT_LDO} = 0.1mA, low-power mode		0.05			
		V _{SYS} = 5.5V, V _{IN_LDO} = 1.7V, V _{OUT_LDO} = 0.8V, I _{OUT_LDO} = 0.1mA to 150mA, normal-power mode		0.5			
Load Regulation		V_{SYS} = 5.5V, V_{IN_LDO} = 1.7V, V_{OUT_LDO} = 0.8V, I_{OUT_LDO} = 0.1mA to 5mA, low-power mode		0.5		0/	
		V _{SYS} = 5.5V, V _{IN_LDO} = 3.975V, V _{OUT_LDO} = 0.8V, I _{OUT_LDO} = 0.1mA to 150mA, normal-power mode		0.5		%	
	V _{SYS} = 5.5V, V _{IN LDO} = 3.	V_{SYS} = 5.5V, V_{IN_LDO} = 3.975V, V_{OUT_LDO} = 0.8V, I_{OUT_LDO} = 0.1mA to 5mA, low-power mode		0.5			

 $(V_{SYS} = 3.6V, V_{IN_LDO} = 3.6V, V_{OUT_LDO} = 1.8V, C_{IN_LDO} = 1\mu F, C_{OUT_LDO} = 2.2\mu F$, limits are 100% production tested at $T_A = +25^{\circ}C$, limits over the operating temperature range $(T_A = -40^{\circ}C \text{ to } +85^{\circ}C)$ are guaranteed by design and characterization, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS	MIN TYP	MAX	UNITS
		Normal-power mode, f = 10Hz to 100kHz, I_{LDO_OUT} = 15mA, V_{SYS} = 2.7V, V_{IN_LDO} = 1.7V, V_{LDO} = 0.8V	100		
		Normal-power mode, f = 10Hz to 100kHz, I_{LDO_OUT} = 15mA, V_{SYS} = 2.7V, V_{IN_LDO} = 1.7V, V_{LDO} = 1.0V	150		
Output Noise		Normal-power mode, f = 10Hz to 100kHz, I_{LDO_OUT} = 15mA, V_{SYS} = 2.7V, V_{IN_LDO} = 2.7V, V_{LDO} = 2.0V	200		μV _{RMS}
DVNAMIC CHARACTER		Normal-power mode, f = 10Hz to 100kHz, I_{LDO_OUT} = 15mA, V_{SYS} = 3.6V, V_{IN_LDO} = 3.6V, V_{LDO} = 3.0V	300		
		Normal-power mode, f = 10Hz to 100kHz, I_{LDO_OUT} = 15mA, V_{SYS} = 5.5V, V_{IN_LDO} = 5.5V, V_{LDO} = 3.975V	400		
DYNAMIC CHARACTER	ISTICS				
Power-Supply Rejection Ratio	PSRR	Normal-power mode, $V_{SYS} = 3.6V$, $V_{IN_LDO} = 2.8V + 20 mVpp$, $f = 10 Hz$ to $10 kHz$, $V_{OUT_LDO} = 1.8V$, $I_{OUT_LDO} = 15 mA$	60		dB
Line Transient		Normal-power mode, V_{OUT_LDO} = 1.2V, I_{OUT_LDO} = 1mA, V_{SYS} = V_{IN_LDO} = 3.6V to 3.2V to 3.6V with 5µs transition times	5		
Line Hansient		Normal-power mode, V_{OUT_LDO} = 1.2V, I_{OUT_LDO} = 1mA, V_{SYS} = 3.6V, V_{IN_LDO} = 3.6V to 3.2V to 3.6V with 5 μ s transition times	5		- mV
Load Transient		Normal-power mode, V _{OUT_LDO} = 2.5V, I _{OUT_LDO} = 1mA to 75mA to 1mA with 1µs transition times, C _{OUT_LDO} = 2.2µF	±5		- %
Loau Hansietti		Normal-power mode, V_{OUT_LDO} = 2.5V, I_{OUT_LDO} = 1mA to 75mA to 1mA with 1µs transition times, C_{OUT_LDO} = 10µF	±3		70
Output Over-Shoot During Startup			50		mV
TIMING CHARACTERIS	TICS				
Maximum Turn-On Delay		From the LDO receiving an enable signal to when the output voltage starts to rise	20		μs
Maximum Soft-Start Time		V _{OUT_LDO} from 10% to 90% of 2.5V final value	40		μs

 $(V_{SYS} = 3.6V, V_{IN_LDO} = 3.6V, V_{OUT_LDO} = 1.8V, C_{IN_LDO} = 1\mu F, C_{OUT_LDO} = 2.2\mu F$, limits are 100% production tested at $T_A = +25^{\circ}C$, limits over the operating temperature range $(T_A = -40^{\circ}C \text{ to } +85^{\circ}C)$ are guaranteed by design and characterization, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
POWER STAGE CHARA	CTERISTCIS					
Drangut Valtage	V	Normal-power mode, V_{SYS} = 3.6V, 3.3V programmed output voltage, V_{IN_LDO} = 3.1V, I_{OUT_LDO} = 150mA (Note 13)		100	200	- mV
Dropout Voltage	V _{LDO_DO}	Normal-power mode, V_{SYS} = 3.6V, 1.8V programmed output voltage, V_{IN_LDO} = 1.7V, I_{OUT_LDO} = 150mA (Note 13)		100		IIIV
Disabled Output Impedance	R _{AD_LDO}	Regulator disabled, active-discharge enabled (ADE_LDO = 1) (Note 16)	50	100	150	Ω
Disabled Output Leakage Current		Regulator disabled, active-discharge disabled (ADE_LDO = 0), V _{SYS} = V _{IN_LDO} = 5.5V, V _{OUT_LDO} = 5.5V and 0V, T _A = +25°C (Note 17)	-1.0	±0.1	+1.0	μA
POWER-OK COMPARAT	ΓOR					
Output Power-OK Trim Level		Rising edge, V _{OUT_LDO} = 2.0V	82.5	87.5	92.5	%
Output Power-OK Hysteresis		V _{OUT_LDO} = 2.0V		3		%

- Note 11: When the input voltage is within the specified range, the LDO tries to regulate the output voltage. However, the regulator may be in dropout. For example, if the output voltage is fixed at 1.85V and a 1.7V input is provided, the output is 1.7V minus the dropout voltage (V_{LDO} = V_{IN} _LDO-V_{LDO}_DO). To achieve the specified output voltage, the input voltage must be the output voltage plus the dropout voltage (V_{IN}_LDO ≥ V_{LDO} + V_{LDO}_DO_MAX).
- Note 12: V_{IN_LDO} must be lower than or equal to V_{SYS}. The V_{SYS} maximum operating voltage range is 5.5V. For example, if V_{SYS} is 4.2V, then the maximum voltage for V_{IN_LDO} is 4.2V. Similarly, if V_{SYS} is 5.5V, then the maximum voltage for V_{IN_LDO} is 5.5V.
- Note 13: The dropout voltage is the difference between the input voltage and the output voltage, when the input voltage is inside the specified "input voltage" range but below the "output voltage" set point. For example, if the output voltage set point is 1.85V, the input voltage is 1.7V, and the actual output voltage is 1.65V, then the dropout voltage is 50mV (V_{LDO_DO} = V_{IN_LDO}-V_{OUT_LDO}).
- Note 14: The "Maximum Output Current" is guaranteed by the "Output Voltage Accuracy" tests.
- Note 15: Current limit is provided for thermal concerns as a system fail safe feature, minor (50mA) oscillations of current when the LDO is at current limit are normal. Over process corner current limit is not expected to exceed 560mA.
- Note 16: There is an n-channel MOSFET in series with the output active discharge resistance. This NMOS requires V_{SYS} > 1.2V to be enhanced.
- **Note 17:** Guaranteed by design and characterization but not directly production tested. The ability to disconnect the active discharge resistance is functionally checked in a production test.

Electrical Characteristics—I²C Interface

 $(T_A = -40^{\circ}C \text{ to } +85^{\circ}C \text{ (unless otherwise specified)}, V_{IO} = 1.8V, V = 3.6V)$

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
POWER SUPPLY			,			
VIO Voltage Range	V _{IN_VIO_I2C}		1.7		3.6	V
SDA AND SCL I/O STAGE		·	-			
SCL, SDA Input HIGH Voltage	V _{IH}	V _{IN_VIO_I2C} = 1.7V to 3.6V	0.7 x V _{IN_VIO_} I2C			V
SCL, SDA Input LOW Voltage	V _{IL}	V _{IN_VIO_I2C} = 1.7V to 3.6V			0.3 x V _{IN} _ VIO_I2C	
SCL, SDA Input Hysteresis	V _{HYS}			0.1 x V _{IN} _ VIO_I2C		V
SCL, SDA Input Leakage Current	I _I	$V_{IN_VIO_I2C}$ = 3.6V, V_{SCL} = V_{SDA} = 0V and 3.6V	-10		+10	μΑ
SDA Output LOW Voltage	V _{OL}	IOL = 20mA			0.4	V
SCL, SDA Pin Capacitance	CI			10		pF
Output Fall Time from V _{IH} to V _{IL}	toF	(Note 18)			120	ns
		OTP_INT_PU[0] = 0b0		Open		
Internal Pullup	R _{PU_SDA} , R _{PU_SCL}	Pulled up to V _{IN_VIO_I2C} , OTP_INT_PU[0] = 0b1	2.5	5.0	7.5	kΩ
WATCHDOG TIMER						
Watchdog Timer Period	t _{WD}			35		ms
I ² C-COMPATIBLE INTERFACE	TIMING (STANI	DARD, FAST, AND FAST-MODE PL	.US) (Note 18)			
Clock Frequency	f _{SCL}		0		1000	kHz
Bus Free Time between STOP and START Condition	t _{BUF}		0.5			μs
Hold Time (REPEATED) START Condition	t _{HD;STA}		0.26			μs
SCL LOW Period	t _{LOW}		0.5			μs
SCL HIGH Period	t _{HIGH}		0.26			μs
Setup Time REPEATED START Condition	tsu_sta		0.26			μs

Electrical Characteristics—I²C Interface (continued)

 $(T_A = -40^{\circ}C \text{ to } +85^{\circ}C \text{ (unless otherwise specified)}, V_{IO} = 1.8V, V = 3.6V)$

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
DATA Hold Time	t _{HD_DAT}	Receive mode	0			μs
DATA Setup Time	tsu_dat		50			ns
Setup Time for STOP Condition	t _{SU_STO}		0.26			μs
Data Valid Time	t _{VD_DAT}	Transmit mode			0.45	μs
Data Valid Acknowledge Time	t _{VD_ACK}				0.45	μs
Bus Capacitance	C _B				550	pF
Pulse Width of Suppressed Spikes	t _{SP}			50		ns
I ² C-COMPATIBLE INTERFACE	TIMING (HIGH-	SPEED MODE, CB = 100pF) (Not	e 18)			
Clock Frequency	f _{SCL}				3.4	MHz
Hold Time (REPEATED) START Condition	t _{HD_STA}		160			ns
Setup Time REPEATED START Condition	t _{SU_STA}		160			ns
SCL LOW Period	t _{LOW}		160			ns
SCL HIGH Period	t _{HIGH}		60			ns
DATA Hold Time	t _{HD_DAT}		0			ns
DATA Setup Time	t _{SU_DAT}		10			ns
SCL Rise Time	t _{rCL}	T _A = +25°C (Note 18)	10		40	ns
Rise Time of SCL Signal After REPEATED START Condition and After Acknowledge Bit	^t rCL1	T _A = +25°C (Note 18)	10		80	ns
SCL Fall Time	t _{fCL}	T _A = +25°C (Note 18)	10		40	ns
SDA Rise Time	t _{rDA}	T _A = +25°C (Note 18)	10		80	ns
SDA Fall Time	t _{fDA}	T _A = +25°C (Note 18)			80	ns
Setup Time for STOP Condition	t _{SU_STO}		160			ns
Bus Capacitance	C _B				100	pF
Pulse Width of Suppressed Spikes	t _{SP}			10		ns

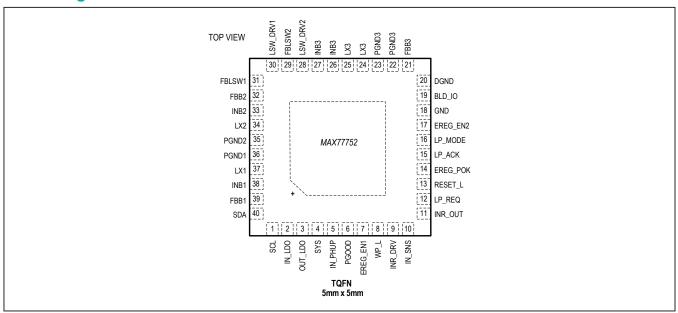
Electrical Characteristics—I²C Interface (continued)

 $(T_A = -40^{\circ}C \text{ to } +85^{\circ}C \text{ (unless otherwise specified)}, V_{IO} = 1.8V, V = 3.6V)$

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	
I ² C-COMPATIBLE INTERFACE TIMING (HIGH-SPEED MODE, CB = 400pF) (Note 18)							
Clock Frequency	f _{SCL}				1.7	MHz	
Hold Time (REPEATED) START Condition	t _{HD_STA}		160			ns	
Setup Time REPEATED START Condition	tsu_sta		160			ns	
SCL LOW Period	t _{LOW}		320			ns	
SCL HIGH Period	t _{HIGH}		120			ns	
DATA Hold Time	t _{HD_DAT}		0			ns	
DATA Setup Time	t _{SU_DAT}		10			ns	
SCL Rise Time	t _{rCL}	T _A = +25°C (Note 18)	20		80	ns	
Rise Time of SCL Signal after REPEATED START Condition and after Acknowledge bit	^t rCL1	T _A = +25°C (Note 18)	20		160	ns	
SCL Fall Time	t _{fCL}	T _A = +25°C (Note 18)	20		80	ns	
SDA Rise Time	t _{rDA}	T _A = +25°C (Note 18)	20		160	ns	
SDA Fall Time	t _{fDA}	T _A = +25°C (Note 18)			160	ns	
Setup Time for STOP Condition	t _{SU_STO}		160			ns	
Bus Capacitance	C _B				400	pF	
Pulse Width of Suppressed Spikes	t _{SP}			10		ns	

Note 18: Minimum typical and maximum values are guaranteed by design. Not production tested.

Pin Configuration



Pin Description

PIN	NAME	FUNCTION	TYPE
ТОР			
9	INR_DRV	External Inrush FET Gate Drive. Inrush MOSFET Gate Driver. When using the inrush control feature, connect INR_DRV to the gate of an external NMOS. If the inrush feature is not required, this pin can also be configured as LSW3_DRV using LSW_OTP_SEL = 1. If either use cases do not apply, leave INR_DRV unconnected or connect to ground ONLY after ensuring that the inrush controller is disabled by the appropriate OTP option.	Analog Output
10	IN_SNS	Input Voltage Sense (Preswitch). Input Voltage Sense. When using the inrush control feature, connect IN_SNS to the drain of an external n-channel MOSFET. When the inrush control feature is not needed, connect IN_SNS to V _{SYS} .	Power Input
11	INR_OUT	Inrush Control Output Sense. This pin must be connected to the source of the inrush control MOSFET. If the inrush controller is not required, this pin can also be configured as FBLSW3 by setting LSW_OTP_SEL = 1. If either use cases do not apply, then this pin must be connected to the SYS node.	Power Input
4	sys	System Power Input. SYS is the voltage sense input for the inrush controller, system voltage monitors, and other analog circuits. Connect SYS to the same power source as that meant for the voltage regulators in the PMIC. When using the inrush control feature, connect SYS to the source of an external n-channel MOSFET whose drain is connected to the main power input. When the inrush control feature is not needed, connect SYS to IN_SNS. Regardless of the inrush controller configuration, SYS must connect to the buck regulator power inputs (INB1, INB2, INB3).	Power Input

PIN	NAME	FUNCTION	TYPE
18	GND	Ground. GND carries ground current for "quiet" control circuits. GND also carries the current for the OTP programming circuit when the programming sequence is executed.	Ground
8	WP_L	Write Protect (Open Drain, Active Low) to memory. Connect this pin to the appropriate pin on the memory. An optional $100 \text{k}\Omega$ internal pullup resistor is available which is pulled up to an internal $V_{\text{IN_VIO}}$ node.	Digital Output
15	LP_ACK	Low-Power Mode Acknowledge from controller. Connect LP_ACK to the appropriate pin on the controller. LP_ACK acknowledges the LP_REQ output signal in master mode (OTP_SLP_MSTRSLV = 0) by asserting high, which initiates the transition to DevSlp state. LP_ACK initiates the transition to DevSlp state independently in slave mode (OTP_SLP_MSTRSLV = 1) by asserting high. An optional $100k\Omega$ internal pullup resistor is available which is pulled up to an internal V_{IN_VIO} node.	Digital Input
13	RESET_L	Reset Output (Open Drain, Active Low) to controller. Connect to the reset input of the controller. An optional $100k\Omega$ internal pullup resistor is available which is pulled up to an internal node.	Digital Output
6	PGOOD	Power Good Output (Open Drain, Active High). PGOOD indicates the status of all regulators controlled by the PMIC (internal and external) and asserts LOW if any regulator's individual Power-OK (POK) signal is deasserted. Additionally, it also asserts low if the system voltage (V_{SYS}) falls below the brownout threshold. Connect PGOOD to the appropriate pin on the controller. An optional $100 \text{k}\Omega$ internal pullup resistor is available which is pulled up to an internal node.	Digital Output
7	EREG_EN1	External Regulator #1 Enable Output. EREG_EN1 is an open-drain output with optional internal pullup resistor. EREG_EN1 is typically used to drive the enable pin of an external regulator.	Digital Output
17	EREG_EN2	External Regulator #2 Enable Output. EREG_EN2 is an open-drain output with optional internal pullup resistor. EREG_EN2 is typically used to drive the enable pin of an external regulator.	Digital Output
14	EREG_POK	External Regulator Power-OK Input. EREG_POK is a digital input. In the typical application, EREG_POK is derived from the POK outputs of the external regulators that are enabled/disabled by EREG_EN1 and EREG_EN2.	Digital Input
16	LP_MODE	Low-Power Mode Input to PMIC from Connector in Master Mode (OTP_SLP_MSTRSLV = 0). When in slave mode (OTP_SLP_MSTRSLV = 1), it is recommended to connect LP_MODE to ground or to a power supply such that it is logic high. Open-Drain Output. An optional $100k\Omega$ internal pullup resistor is available which is pulled up to an internal node.	Digital Input
12	LP_REQ	Open-Drain (Active High) Output. Low-power mode request to controller in master mode in alternate mode (OTP_SLP_MSTRSLV = 0). Open-Drain Output. An optional $100k\Omega$ internal pullup resistor is available which is pulled up to an internal node.	Digital Output

PIN	NAME	FUNCTION	TYPE
20	DGND	Digital Ground. DGND carries ground current for digital circuits such as the I ² C.	Ground
19	BLD_IO	BLD_IO Pin is a Dedicated Open-Drain Input/Output Pin. In an application, this active-low input discharges the supply rail during a powerup cycle. This pin also senses the voltage on the pin it is connected to, and the function is to discharge a rail lower than 100mV. Connect this pin to GND when this feature is not required in the system.	Analog I/O
5	IN_PHUP	IN_PHUP is a Dedicated Analog Input Pin. This pin is connected to the output of the power holdup IC. In case of a power-fail event, the voltage on this pin drives the internal logic block to sustain the holdup function by maintaining the logic levels of the appropriate pins.	Power Input
LDO			
2	IN_LDO	Input Power for LDO (150mA). Bypass with a $2.2\mu F$ ceramic capacitor to GND with the following parasitic constraints (including capacitor and PCB parasitics) of ESR<100m Ω and ESL<30nH. If the LDO is not used, it is recommended to connect IN_LDO to OUT_LDO and connect them to ground.	Power Input
3	OUT_LDO	150mA PMOS LDO Output. Bypass with a 2.2µF capacitor to GND. If the LDO is not used, it is recommended to either ground OUT_LDO or leave it unconnected.	Power output
BUCK	1		1
38	INB1	BUCK1 Power Input. INB1 is the shared drain connection of BUCK1's main power FET. Connect both INB1 pins together and to the power input to the system. INB1 is a critical discontinuous current node that requires careful PCB layout.	Power Input
37	LX1	BUCK1 Switching Node. Connect the required inductor between LX and the output capacitor. Both LX1 pins must be connected together. LX1 is a critical node that requires careful PCB layout.	Power I/O
36	PGND1	BUCK1 Power Ground are Internally Combined. PGND1 is the source connection of BUCK1's synchronous rectifier. PGND1 is a critical discontinuous current node that requires careful PCB layout.	Ground
39	FBB1	BUCK1 Output Voltage Feedback Node. Connect FBB1 to the local output capacitor at the buck output. In addition to setting the output-voltage regulation threshold, FBB1 can also be programmed to discharge the output capacitor when the converter is shutdown. FBB1 is a critical analog input that requires careful PCB layout.	Analog Input
33	INB2	BUCK2 Power Input. INB2 is the shared drain connection of BUCK2's main power FET. Connect both INB2 pins together and to the power input to the system. INB2 is a critical discontinuous current node that requires careful PCB layout.	Power Input

PIN	NAME	FUNCTION	TYPE
34	LX2	BUCK2 Switching Node. Connect the required inductor between LX and the output capacitor. LX2 is a critical node that requires careful PCB layout.	
35	PGND2	BUCK2 Power Ground are Internally Combined. PGND2 is the source connection of BUCK2's synchronous rectifier. PGND2 is a critical discontinuous current node that requires careful PCB layout.	
32	FBB2	BUCK2 Output Voltage Feedback Node. Connect FBB2 to the local output capacitor at the buck output. In addition to setting the output-voltage regulation threshold, FBB2 can also be programmed to discharge the output capacitor when the converter is shutdown. FBB2 is a critical analog nput that requires careful PCB layout.	
26,27	INB3	BUCK3 Power Input. INB3 is the drain connection of BUCK3's main power FET. Connect to the power input to the system. INB3 is a critical discontinuous current node that requires careful PCB layout.	Power Input
24,25	LX3	BUCK3 Switching Node. Connect the required inductor between LX and the output capacitor. LX3 is a critical node that requires careful PCB layout.	
21	FBB3	BUCK3 Output Voltage Feedback Node. Connect FBB3 to the local output capacitor at the Buck output. In addition to setting the output voltage regulation threshold, FBB3 may also be programmed to discharges the output capacitor when the converter is shutdown. FBB3 is a critical analog input that requires careful PCB layout.	
22, 23	PGND3	BUCK3 Power Ground. PGND2 is the shared source connection of BUCK3's synchronous rectifier. Connect both PGND3 pins together. PGND3 is a critical discontinuous current node that requires careful PCB layout.	
I ² C			
40	SDA	Serial Interface Data Bidirectional Open Drain. An optional $5k\Omega$ internal pullup resistor is available which is pulled up to an internal $V_{IN_VIO_I2C}$ node. If the part is in Off state due to HICCUP_CNT_EXPIRE = 1, the I²C power switches from V_{FBB1} to V_{IN_PHUP} . Otherwise, the pin is in Hi-Z state during Off condition.	
1	SCL	Serial Interface Port 0 Clock Input. Open-Drain Output. An optional $5k\Omega$ internal pullup resistor is available which is pulled up to an internal $V_{IN_VIO_I2C}$ node. If the part is in Off state due to HICCUP_CNT_EXPIRE = 1, the I²C power switchs from V_{FBB1} to V_{IN_PHUP} . Otherwise, the pin is in Hi-Z state during Off condition.	

PIN	NAME	FUNCTION	TYPE		
LOAD SWITCH					
30	LSW_DRV1	Gate Drive for Load Switch 1. Connect to the gate of an external n-channel MOSFET used as the load switch. If the load switch is not used, LSW_DRV1 must be left unconnected.	Analog Output		
28	LSW_DRV2	Gate Drive for LSW2. Connect to the gate of an external n-channel MOSFET used as the load switch. If the load switch is not used, LSW_DRV2 must be left unconnected.	Analog Output		
31	FBLSW1	Feedback Input for Load-Switch Controller 1. FBLSW1 is an analog input to the load-switch controller which is used to control soft-start of the load switch and is the input to the output voltage monitor. Connect FBLSW1 to the output (source-side of n-channel MOSFET) of the load switch. If the load switch is not used, FBLSW1 can be left unconnected or tied to ground.	Analog Input		
29	FBLSW2	Feedback Input for Load-Switch Controller 2. FBLSW2 is an analog input to the load-switch controller which is used to control soft-start of the load switch and is the input to the output voltage monitor. Connect FBLSW2 to the output (source-side of n-channel MOSFET) of the load switch. If the load switch is not used, FBLSW2 can be left unconnected or tied to ground.	Analog Input		

Detailed Description— Software Recommendations

Advice for optimizing software is provided throughout this data sheet within the context of the hardware descriptions. This section is dedicated to software recommendations and provides system level software guidance in order to optimally utilize the features of this device.

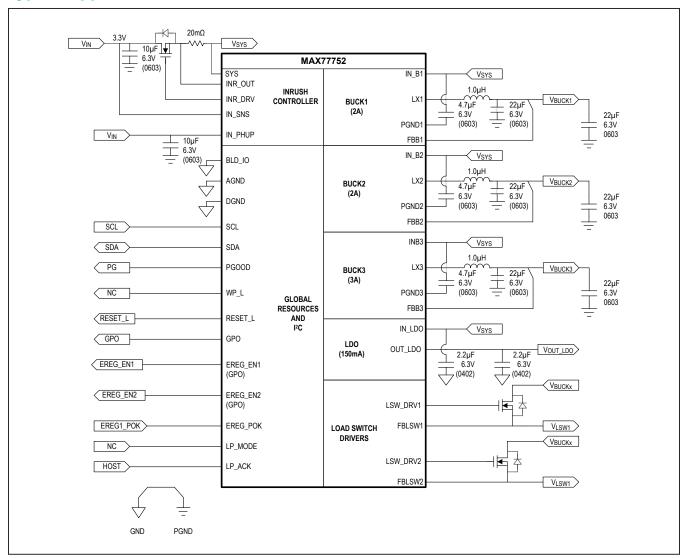
OFF to ON Software Initialization

The system processor typically runs a set of initialization code each time a transition from the OFF to the ON state occurs, the reset output is deasserted (RESET_L = 1), and the PGOOD is asserted (PGOOD = 1).

The following are recommended software steps within this initialization code:

- 1) Read the interrupt bits:
 - 1) Interrupt bits set at this point in time can indicate an issue that previously caused a shutdown.
- 2) Check the values of CID0, CID1, CID2, CID3, and CID4. Consider reporting these values if the product has some form of serial number checking utility. If the SBT bits do not read an appropriate value, then flag the product as bad and do not ship it. Only values of 0b011 and 0b101 should be shipped as production units. If the DRV bits do not match with what was intended for the given product, then flag that product as bad and do not ship it. This device has many OTP options and the DRV bits are set differently for each set of options. If parts got mixed up in the warehouse (i.e., A version confused for C version), then this step helps catch that mistake.
- 3) Set/Clear the mask bits as deemed appropriate for the target platform.

Typical Application Circuits



Ordering Information

PART NUMBER	TEMP RANGE	PIN-PACKAGE	TOP MARKING
MAX77752AETL+	-40°C to +85°C	40 TQFN	MAX77752AETL+
MAX77752BETL+	-40°C to +85°C	40 TQFN	MAX77752BETL+

⁺ Denotes a lead(Pb)-free/RoHS-compliant package.

For a copy of the register map and for further questions, contact techsupport@maximintegrated.com.

MAX77752

Multi-Channel Integrated Power Management IC

Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	12/17	Initial release	
1	1/18	Added conditions statement to the <i>Electrical Characteristics—Current Sense Amplifier</i> table	14
2	1/18	Removed SSD and NAND from <i>Pin Description</i> table, added new part variant to <i>Ordering Information</i> table	38, 42

For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim Integrated's website at www.maximintegrated.com.

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