

Three Phase Brushless Motor Pre-driver

BD63003MUV

General Description

BD63003MUV is a Pre-driver IC for 3-phase brushless motor. It generates a driving signal from the Hall sensor and applies PWM in the input control signal for motor control. It has a built-in booster circuit which allows Nch-Nch MOS transistors to be used as the external power Transistor. It is compatible with 12 V or 24 V power supply and has various controls and built-in protection functions, making it useful for variety of purposes. Since the IC adopts a small package, it can also be used on small diameter motors.

Key Specifications

- Power Supply Voltage Rating: 40 V
- Low Side Gate Drive Voltage: 10 V (Typ)
- High Side Gate Drive Voltage: 10 V (Typ)
- Operating Temperature Range: -40 °C to +85 °C
- Current Limit Detect Voltage: 0.2 V±10 %
- UVLO Lockout Voltage: 6.0 V (Typ)
- OVLO Lockout Voltage: 28.5 V (Typ)

Package

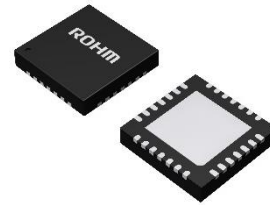
VQFN032V5050

W (Typ) x D (Typ) x H (Max)

5.0 mm x 5.0 mm x 1.0 mm

Features

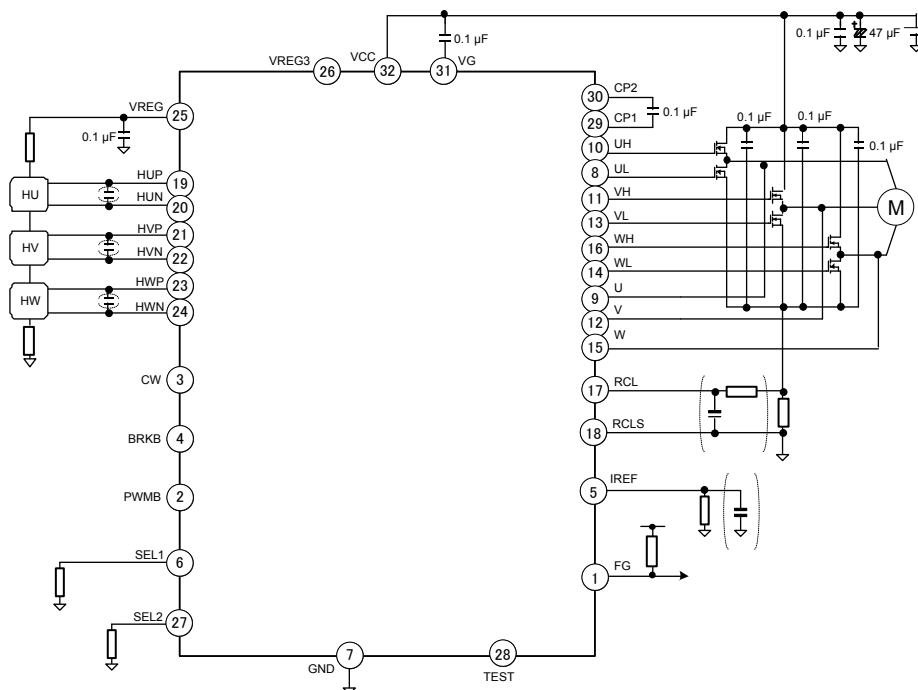
- Built in 120° Commutation Logic Circuit
- Driving with Nch-Nch MOS Transistors
- Built in Peak Current Control Function
- For Controller Input of 3.3 V and 5 V
- PWM Control Mode (lower arm switching)
- CW/CCW Function
- Short Brake Function
- FG Output (1FG / 3FG conversion)
- Built-in Protection Circuit for Current Limit (CL), Overheating (TSD), Under Voltage (UVLO), Over Voltage (OVLO), Motor Lock (MLP)



Applications

- OA Machines
- Other General Civil Equipment

Typical Application Circuit

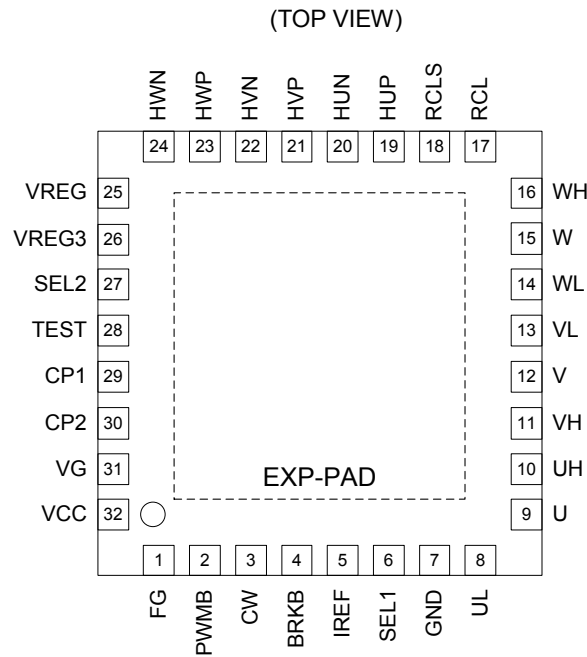


○Product structure : Silicon integrated circuit ○This product has no designed protection against radioactive rays.

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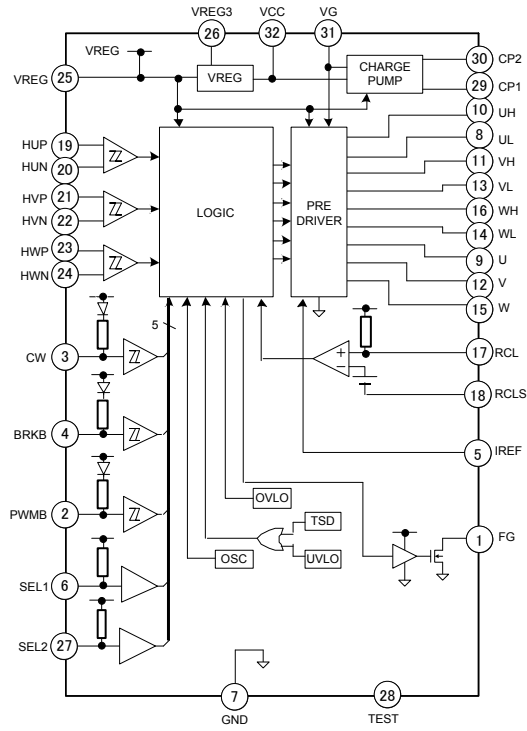
Pin Configuration



Pin Descriptions

Pin No.	Pin Name	Function	Pin No.	Pin Name	Function
1	FG	1FG / 3FG output	17	RCL	Detect voltage input for over current
2	PWMB	PWM input (negative logic)	18	RCLS	RCL sense input
3	CW	Changing direction of rotation (H: CW, L: CCW)	19	HUP	U phase hall input +
4	BRKB	Brake input (negative logic)	20	HUN	U phase hall input -
5	IREF	Output driving current setting	21	HVP	V phase hall input +
6	SEL1	Function setting input 1	22	HVN	V phase hall input -
7	GND	GND	23	HWP	W phase hall input +
8	UL	U phase lower output	24	HWN	W phase hall input -
9	U	U phase output feedback	25	VREG	VREG output
10	UH	U phase upper output	26	VREG3	VREG3 output
11	VH	V phase upper output	27	SEL2	Function setting input 2
12	V	V phase output feedback	28	TEST	TEST input (GND)
13	VL	V phase lower output	29	CP1	Charge pump setting 1
14	WL	W phase lower output	30	CP2	Charge pump setting 2
15	W	W phase output feedback	31	VG	Charge pump output
16	WH	W phase upper output	32	VCC	Power supply
-	EXP-PAD	Connect EXP-PAD to GND			

Block Diagram



Description of Blocks

Commutation logic (120° commutation)

Truth Table

HU	HV	HW	CW (CW = H or OPEN)						FG	
			UH	UL	VH	VL	WH	WL	1FG	3FG
H	L	H	PWM	PWM ^(Note 1)	H	L	L	L	L	Hi-z
H	L	L	$\overline{\text{PWM}}$	PWM ^(Note 1)	L	L	H	L	L	L
H	H	L	L	L	$\overline{\text{PWM}}$	PWM ^(Note 1)	H	L	L	Hi-z
L	H	L	H	L	$\overline{\text{PWM}}$	PWM ^(Note 1)	L	L	Hi-z	L
L	H	H	H	L	L	L	$\overline{\text{PWM}}$	PWM ^(Note 1)	Hi-z	Hi-z
L	L	H	L	L	H	L	$\overline{\text{PWM}}$	PWM ^(Note 1)	Hi-z	L
H	H	H	L	L	L	L	L	L	L	L
L	L	L	L	L	L	L	L	L	Hi-z	Hi-z

HU	HV	HW	CCW (CW = L)						FG	
			UH	UL	VH	VL	WH	WL	1FG	3FG
H	L	H	H	L	$\overline{\text{PWM}}$	PWM ^(Note 1)	L	L	L	Hi-z
H	L	L	H	L	L	L	PWM	PWM ^(Note 1)	L	L
H	H	L	L	L	H	L	$\overline{\text{PWM}}$	PWM ^(Note 1)	L	Hi-z
L	H	L	$\overline{\text{PWM}}$	PWM ^(Note 1)	H	L	L	L	Hi-z	L
L	H	H	$\overline{\text{PWM}}$	PWM ^(Note 1)	L	L	H	L	Hi-z	Hi-z
L	L	H	L	L	$\overline{\text{PWM}}$	PWM ^(Note 1)	H	L	Hi-z	L
H	H	H	L	L	L	L	L	L	L	L
L	L	L	L	L	L	L	L	L	Hi-z	Hi-z

(Note 1) When PWM = "L", $\overline{\text{PWM}}$ = "H". When PWM = "H", $\overline{\text{PWM}}$ = "L".

- 1. Regulator Output Pin (VREG)**
 This is constant voltage output pin of 5 V (Typ). Connect capacitors of 0.01 μF to 1 μF. Be careful that VREG current does not exceed absolute maximum ratings in case of being used for bias power supply of hall elements.
- 2. Regulator Output Pin (VREG3)**
 This is constant voltage output pin of 3.3 V (Typ). VREG3 can be used for bias voltage of hall elements. Be careful that VREG3 current does not exceed absolute maximum ratings in case of being used.

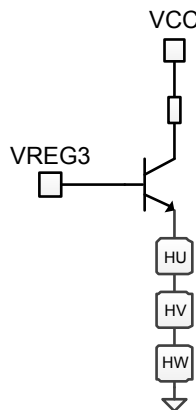


Figure 1. VREG3 reference circuit

Description of Blocks - continued

3. PWM Input Pin (PWMB)

Speed can be controlled by inputting PWM signal into the PWMB pin (negative logic). Synchronous rectifier PWM can be achieved through lower switching. When PWMB = "L", lower external FET that matches to Hall input logic is "L". When PWMB = "H" or open, lower external FET is "H". When PWMB = "H" or OPEN status is detected for 104 μ s (Typ), the synchronous rectifier is OFF. Synchronous rectifier turns ON through the falling edges of subsequent PWMB signals. At startup, External FET keeps "Hi-z" states in which all phase is OFF (stand-by) until PWMB = "L" status is detected 2 μ s (Typ) or more [figure 2]. However, the internal regulator of VREG, VG, the other regulator, protection function of OVLO, and the other ones are operated in the stand-by. Additionally, the PWMB pin is pulled up by internal 3.3V (Typ) through a resistance of 100 k Ω \pm 30 k Ω and pulled down by GND through a resistance of 1000 k Ω \pm 300 k Ω .

PWMB	PWM phase Lower External FET
H or OPEN	OFF
L	ON

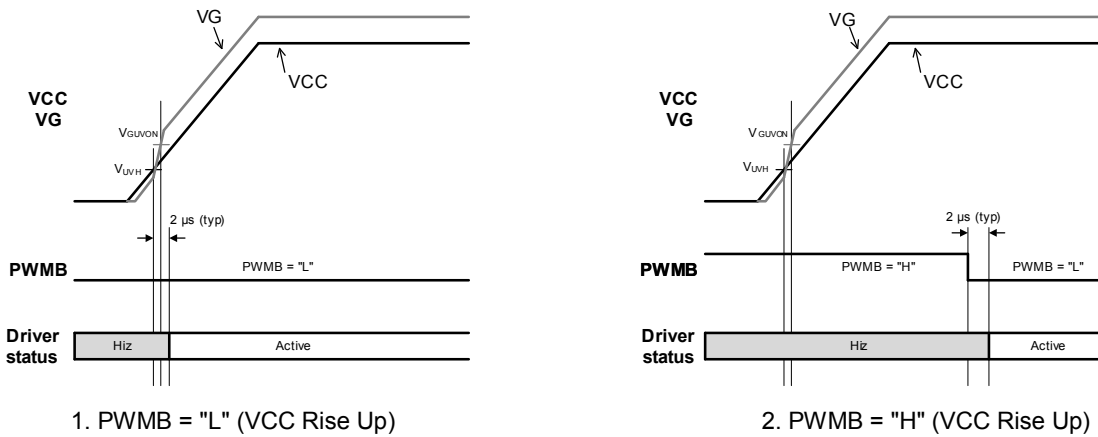


Figure 2. PWMB Status

4. BRKB Pin (BRKB)

Motor rotation can be quickly stopped using the BRKB Pin (negative logic). When BRKB = "L", this causes all the upper external FET to turn "OFF" and all the lower external FET to turn "ON", initiating short brake operation. When BRKB = "H" or OPEN, then short brake operation will be released. Additionally, the BRKB pin is pulled up by internal 3.3V (Typ) through a resistance of 100 k Ω \pm 30 k Ω and pulled down by GND through a resistance of 1000 k Ω \pm 300 k Ω .

BRKB	Operation
H or OPEN	Normal
L	Short brake

Description of Blocks - continued

5. Rotatory Direction Change Pin (CW)

Rotation direction can be switched with the CW pin. When CW = "H" or OPEN, the direction is Clockwise (CW). When CW = "L", the direction will be Counter Clockwise (CCW). We do not recommend switching rotation direction when motor is rotating. If rotation direction is switched when rotating, the operation is the following due to the condition of SEL1.

(1) SBRK = "Enable" (SEL1 = "H" or "M1")

After having performed short brakes movement until hall frequency becomes approximately 40 Hz (Typ) or less, rotatory direction is replaced. In the case of this condition, do not change the logic of CW for 10 ms after brakes cancellation by the BRKB input (Figure 3).

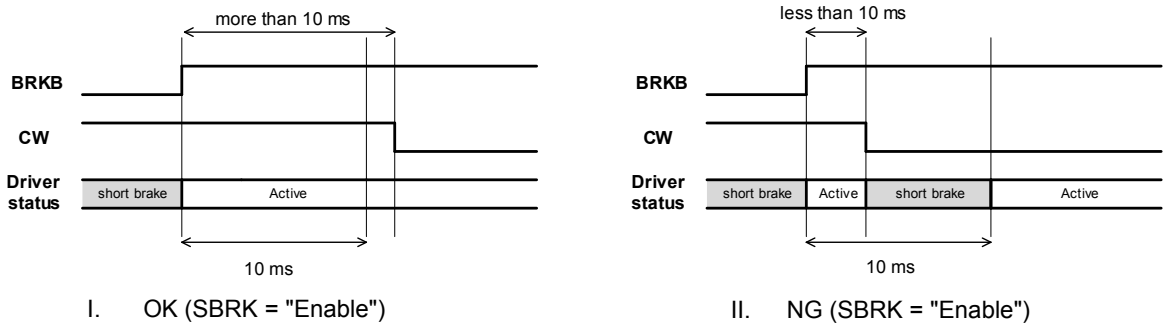


Figure 3. BRKB, CW Control Timing Limitation 1

(2) SBRK = "Disable" (SEL1 = "M2" or "L")

Without the short brake such as (1), direction is replaced. In this case, be careful since high current may sometimes flow in the external FET when the direction is replaced as described. In addition, there is no limitation in timing of BRKB and CW such as (1) in this condition (Figure 4).

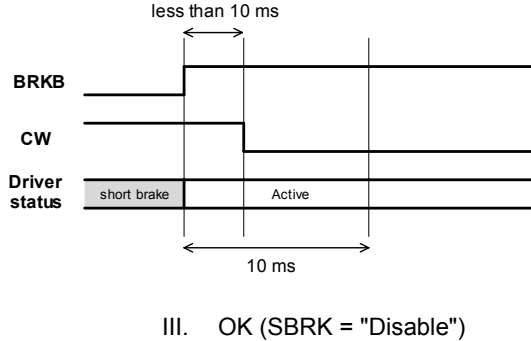


Figure 4. BRKB, CW Control Timing Limitation 2

In addition, the CW pin is pulled up by internal 3.3 V (Typ) through resistance of 100 kΩ ± 30 kΩ and pulled down by GND through a resistance of 1000 kΩ ± 300 kΩ.

CW	Direction
H or OPEN	CW
L	CCW

Description of Blocks - continued

6. Select Pin 1 (SEL1)

The SEL1 pin can be used to select 1FG / 3FG and "Enable" or "Disable" of the short brakes (SBRK) at the direction change with the CW pin. The SEL1 pin is a 4 input pin and can be set by being OPEN, 220 kΩ [$\pm 5\%$] to GND, 47 kΩ [$\pm 5\%$] to GND, and tied to GND. In addition, the SEL1 pin is pulled up by VREG through resistance of 100 kΩ ± 30 kΩ.

SEL1	FG	SBRK(CW/CCW)
H (OPEN)	3FG	Enable
M1 (220 kΩ [$\pm 5\%$] to GND)	1FG	Enable
M2 (47 kΩ [$\pm 5\%$] to GND)	1FG	Disable
L (tied to GND)	3FG	Disable

7. Select Pin 2 (SEL2)

The SEL2 pin can be used to select Enable/Disable of MLP and OVLO. In addition, the SEL2 pin is pulled up by VREG through resistance of 100 kΩ ± 30 kΩ.

SEL2	MLP	OVLO
H (OPEN)	2.2 s (Typ)	28.5 V (Typ)
M1 (220 kΩ [$\pm 5\%$] to GND)	Disable	28.5 V (Typ)
M2 (47 kΩ [$\pm 5\%$] to GND)	2.2 s (Typ)	Disable
L (tied to GND)	Disable	Disable

8. FG Output pin (FG)

FG signal is composed by a hall signal and is outputted by the FG pin. Changing between 1FG and 3FG can be done with SEL1. In addition, the FG pin, which is an open drain output, should be externally pulled-up by a resistance of the about 10 kΩ to 100 kΩ.

9. Hall Input (Hall: HUP, HUN, HVP, HVN, HWP, HWN)

Hall comparator inside the IC is designed with hysteresis (± 15 mV (Typ)) in order to prevent malfunction due to noise. Always set correct bias current for the Hall element so that the amplitude of Hall input voltage will be the minimum input voltage ($V_{HALLMIN}$) or more. Furthermore, the output of the comparator has a digital filter of 2 μ s (Typ). However, if it can't prevent the noise, it is recommended to connect a ceramic capacitor with about 100 pF to 0.01 μ F value between the input pins of the Hall comparator. The in-phase input voltage range (V_{HALLCM} : 0 V to $V_{REG}-1.7$ V (Typ)) is designed for Hall comparator, set within this range when applying bias to the Hall element. When all Hall inputs become "H" or "L", all external FETs will be "OFF" by the hall input abnormal detection circuit.

10. Booster Circuit

There is built-in booster circuit used to drive upper Nch MOS transistor. The VG pins can produce a boost voltage (the VCC voltage+10 V (Typ)) through connecting capacitors between CP1 - CP2 and between VG - VCC. We recommend connected capacitors to be 0.1 μ F or more. Because CP1 and CP2 are oscillated, the capacitors have to be located near IC. If need, add GND line for the shield. In addition, because VG voltage are boosted from the voltage that based on VCC, if VCC voltage is instability, it can be caused the malfunction such as VG voltage rise up. Therefore, add capacitor between VCC and GND as necessary to stabilize VCC voltage when using large current and motor with large BEMF. Because there is built-in protection circuit for insufficient booster, when VG voltage is V_{GUVON} ($V_{CC}+7$ V (Typ)) or less, all external FETs will be "OFF".

11. Current Limit Circuit (CL Circuit)

Output current limit (Current Limit: CL) circuit can be formed by connecting a low resistance used for detecting current between the RCL pin and the RCLS pin. When RCL voltage is detected 0.2 V (Typ) or more, all lower external FET will be "OFF". It returns by itself after a set amount of time (32 μ s (Typ)). This operation does not synchronize with PWM signal input into the PWMB pin. In addition, in order to avoid misdetection of output current due to RCL noise, the IC sets up a noise-masking period (1 μ s to 2 μ s (Typ)). During the noise-masking period, current detection is disabled. RCLS is the sense line of RCL. If RCLS becomes OPEN, Current Limit may not be normally Function. Connect the RCLS pin to the GND nearest to the current sense resistor's pin.

Description of Blocks - continued

12. Thermal Shutdown Circuit (TSD Circuit)
When chip temperature of driver IC rises and exceeds the set temperature (165 °C (Typ)), the thermal Shutdown circuit (Thermal Shutdown: TSD) begins working. At this time, all external FETs will be "OFF". In addition, the TSD circuit is designed with hysteresis (25 °C (Typ)), and will return to normal working condition when the chip temperature drops. Moreover, the purpose of the TSD circuit is to protect the driver IC from thermal breakdown, therefore, temperature of this circuit will be over working temperature when this circuit operates. Thus, thermal design should have sufficient margin, so do not take continuous use and operation of the circuit as precondition.
13. Under Voltage Lock Out Circuit (UVLO Circuit)
There is a built-in under voltage lockout circuit (Under Voltage Lockout: UVLO circuit) used to ensure the minimum power supply voltage for drive IC to work and to prevent error in the operation of IC. When VCC voltage declined to V_{UVL} (6 V (Typ)), all external FETs should be "OFF". At the same time, UVLO circuit is designed with hysteresis (1 V (Typ)), so when VCC voltage reaches V_{UVH} (7 V (Typ)) or more, it will enter normal operation.
14. Over Voltage Lock Out Circuit (OVLO Circuit)
There is built-in over voltage lockout circuit (Over Voltage Lockout: OVLO circuit) used to restrain the increase of VCC voltage when motor is decelerating. When VCC voltage is 28.5 V (Typ) or more, short brake action will be conducted. In order to avoid misdetection, the IC sets up a noise-masking period (2 μ s to 3 μ s (Typ)). The short brake operation is released after a certain period of time (4 ms) when the VCC voltage is less than or equal to 27.5 V (Typ) and returns to normal operation. OVLO function does not work in case of SEL2 = "Disable"
15. Motor Lock Protection Circuit (MLP Circuit)
There is built-in motor lock protection circuit (Motor Lock Protection: MLP), ON/OFF of MLP circuit can be set by the SEL2 pin. When the MLP setting of SEL2 = "Enable" and the Hall signal logic does not change for 2.2 s (Typ) or more, all external FETs will all be latched as "OFF". Latch can be released through switching BRKB/CW logic. Moreover, when PWMB = "H" or OPEN state is detected for about 15 ms (Typ), latch can be released by the falling edges of subsequent PWMB. However, the MLP circuit does not operate when the MLP setting of the SEL2 pin is "Disable" and when the short brake (including when switching the direction of rotation) or the TSD circuit is in operation.
16. Predriver Output
The drive signal generated by the internal logic outputs the drive signal to the external output power transistor. Driving voltage of upper gate is VG voltage ($V_{CC}+10$ V (Typ)) and driving voltage of lower gate is the internal REG voltage (10 V (Typ)). In addition, a dead time (0.2 μ s (Typ)) is designed between the driving signals of upper gate and lower gate in order to prevent the upper and lower FET from being set to ON at the same time during synchronous rectifier PWM operation. Due to the influence of the motor's counter electromotive force, the output feedback pin (U, V, W) might swing under GND potential, which can cause malfunctions or destruction. When negative potential exceeds -2 V (min), Schottky diode can be inserted to prevent malfunction or destruction.
17. Pre-driver Output Peak Current Setting Pin (IREF)
A current of the pre-driver output can be set by connecting a resistor between the IREF Pin and GND. Note that if the IREF pin is connected with GND or open, it may cause malfunction. The range of the resistance is 27 k Ω [± 5 %] to 150 k Ω [± 5 %]. About the approximate value of the output current, refer to the following table.

Resistor Value [k Ω]	Output Source Current [mA]	Output Sink Current [mA]
150	16	27
120	18	33
100	22	40
82	26	48
68	31	58
56	36	68
47	42	84
39	48	96
33	55	113
27	63	136

Figure 5. The reference value of Pre-driver output current

Description of Blocks - continued

18. Control Signal Sequence

Though we recommend you to input control signals of the CW, PWMB, and BRKB pins after inputting VCC, there won't be any problem if done otherwise. However, if MLP = "Enable" is set at startup, the MLP circuit will not be able to start the motor if the rotation of the motor is not detected within the set time (the edge of the FG signal is not input). Moreover, the control signal and the IC internal signal are given priority. Refer to the table below.

Priority of Control Signal

Priority	Input / Internal signals
1st	UVLO
2nd	BRKB $\uparrow\downarrow$ (Note 2), CW $\uparrow\downarrow$ (Note 2), PWMB \downarrow (Note 2)
3rd	TSD, MLP, HALLERR
4th	OVLO
5th	VG_UVLO, stand-by
6th	BRKB
7th	CL
8th	PWMB, CW,

(Note 2) $\uparrow\downarrow$ means rising and falling edge of the signal.
Refer to a figure of in condition transition for the signal name.

Absolute Maximum Ratings (Ta = 25 °C)

Parameter	Symbol	Rating	Unit
Power Supply Voltage	V _{CC}	-0.3 to +40.0	V
VG Voltage	V _G	-0.3 to +52.0	V
External FET Output Feedback Voltage	V _(U, V, W)	40	V
FG Voltage	V _{FG}	-0.3 to +7.0	V
RCL Voltage	V _{RCL}	-0.3 to +5.5	V
RCLS Voltage	V _{RCLS}	-0.3 to +0.3	V
Voltage of Input of Control and Hall Pins	V _{I/O}	-0.3 to +7.0	V
FG Output Current	I _{FG}	5	mA
VREG Output Current	I _{VREG}	-30	mA
VREG3 Output Current	I _{VREG3}	-5	mA
Maximum Junction Temperature	T _{jmax}	150	°C
Storage Temperature Range	T _{stg}	-55 to +150	°C

Caution 1: Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.

Caution 2: Should by any chance the maximum junction temperature rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. In case of exceeding this absolute maximum rating, design a PCB with thermal resistance taken into consideration by increasing board size and copper area so as not to exceed the maximum junction temperature rating.

Thermal Resistance (Note 3)

Parameter	Symbol	Thermal Resistance (Typ)		Unit
		1s ^(Note 5)	2s2p ^(Note 6)	
VQFN032V5050				
Junction to Ambient	θ _{JA}	138.9	39.1	°C/W
Junction to Top Characterization Parameter ^(Note 4)	Ψ _{JT}	11	5	°C/W

(Note 3) Based on JE51-2A (Still-Air).

(Note 4) The thermal characterization parameter to report the difference between junction temperature and the temperature at the top center of the outside surface of the component package.

(Note 5) Using a PCB board based on JE51-3.

(Note 6) Using a PCB board based on JE51-5, 7.

Layer Number of Measurement Board	Material	Board Size
Single	FR-4	114.3 mm x 76.2 mm x 1.57 mmt

Top	
Copper Pattern	Thickness
Footprints and Traces	70 μm

Layer Number of Measurement Board	Material	Board Size	Thermal Via ^(Note 7)	
			Pitch	Diameter
4 Layers	FR-4	114.3 mm x 76.2 mm x 1.6 mmt	1.20 mm	Φ0.30 mm

Top		2 Internal Layers		Bottom	
Copper Pattern	Thickness	Copper Pattern	Thickness	Copper Pattern	Thickness
Footprints and Traces	70 μm	74.2 mm x 74.2 mm	35 μm	74.2 mm x 74.2 mm	70 μm

(Note 7) This thermal via connects with the copper pattern of all layers.

Recommended Operating Conditions

Parameter	Symbol	Min	Typ	Max	Unit
Operating Temperature	T _{opr}	-40	+25	+85	°C
Power Supply Voltage	V _{CC}	10.8	24.0	26.4	V

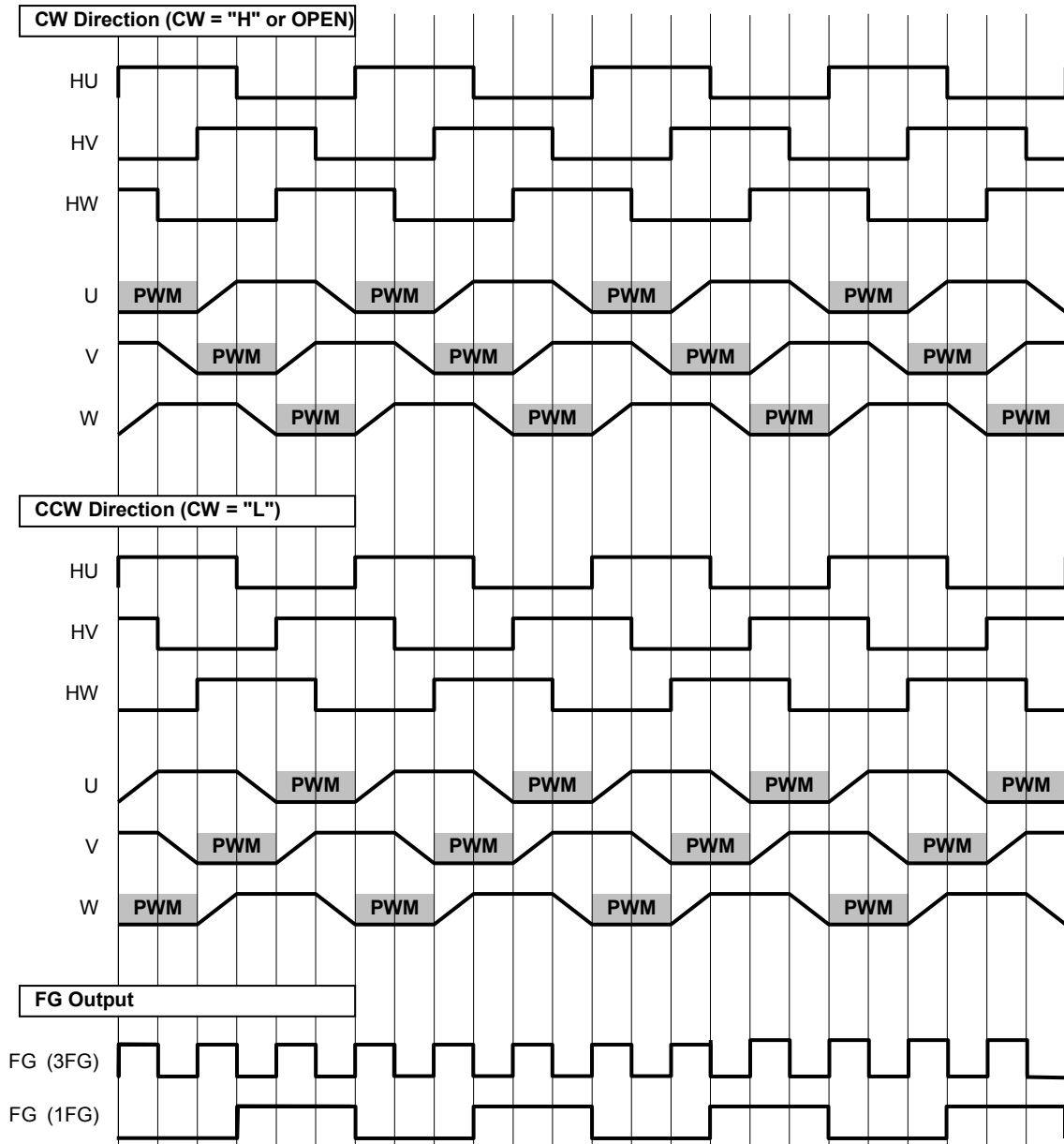
Electrical Characteristics (Unless otherwise specified $V_{CC} = 24\text{ V}$, $T_a = 25\text{ }^\circ\text{C}$)

Parameter	Symbol	Min	Typ	Max	Unit	Conditions
Whole						
Circuit Electric Current	I_{CC}	-	4.5	9	mA	
The VREG Voltage	V_{REG}	4.5	5.0	5.5	V	$I_{VREG} = -10\text{ mA}$
The VREG3 Voltage	V_{REG3}	3.0	3.3	3.6	V	$I_{VREG3} = -1\text{ mA}$
Booster Circuit						
VG Voltage	V_{G1}	$V_{CC}+7$	$V_{CC}+10$	$V_{CC}+11.5$	V	$I_{VG} = 15\text{ mA}$
VG UVLO Voltage	V_{GUVON}	$V_{CC}+6$	$V_{CC}+7$	$V_{CC}+8$	V	
Driver Output						
High Side V_{GS} Gate Drive Voltage	V_{GSH1}	7	10	11.5	V	
Low Side V_{GS} Gate Drive Voltage	V_{GSL1}	7	10	12.5	V	
Source Electric Current Setting Range	I_{RNG1}	-	-	70	mA	$R_{IREF} = 27\text{ k}\Omega$
Sink Electric Current Setting Range	I_{RNG2}	-	-	140	mA	$R_{IREF} = 27\text{ k}\Omega$
Output Peak Source Current	I_{OH}	-	18	-	mA	$R_{IREF} = 120\text{ k}\Omega$
Output Peak Sink Current	I_{OL}	-	33	-	mA	$R_{IREF} = 120\text{ k}\Omega$
IREF Voltage	V_{IREF}		1.2		V	
Hall Input						
Input Bias Current	I_{HALL}	-2.0	-0.1	+2.0	μA	$V_{IN} = 0\text{ V}$
Range of In-phase Input Voltage	V_{HALLCM}	0	-	$V_{REG}-1.7$	V	
Minimum Input Voltage	$V_{HALLMIN}$	50	-	-	mVp-p	
HYS Level +	$V_{HALLHY+}$	5	15	25	mV	
HYS Level -	$V_{HALLHY-}$	-25	-15	-5	mV	
Input of Control: PWMB, CW, BRKB						
Input Electric Current	I_{IN}	-46	-33	-20	μA	$V_{IN} = 0\text{ V}$
The Input H Voltage	V_{INH}	2.0	-	5.5	V	
The Input L Voltage	V_{INL}	0	-	0.8	V	
The Smallest Input Pulse Width	t_{PLSMIN}	1	-	-	ms	CW, BRKB
Input Frequency Range	f_{PWM}	10	-	50	kHz	PWMB
Input of Control: SEL1, SEL2						
Input Current	I_{SEL}	-80	-50	-30	μA	$V_{SEL} = 0\text{ V}$
FG Output						
Output Voltage L	V_{FGOL}	0	0.1	0.3	V	$I_{FG} = 2\text{ mA}$
Leak Current	I_{FGLEAK}	-	0	1	μA	$V_{FG} = 5\text{ V}$
Current Limit						
The Detection Voltage	V_{CL}	0.18	0.20	0.22	V	
Input Bias Electric Current	I_{RCL}	-32	-20	-12	μA	$V_{RCL} = 0\text{ V}$
Input Voltage Range	V_{RCL}	-0.3	-	+1.0	V	

Electrical Characteristics - continued (Unless otherwise specified $V_{CC} = 24\text{ V}$, $T_a = 25\text{ }^\circ\text{C}$)

Parameter	Symbol	Min	Typ	Max	Unit	Conditions
UVLO						
Release Voltage	V_{UVH}	6.5	7.0	7.5	V	
Lockout Voltage	V_{UVL}	5.5	6.0	6.5	V	
OVLO						
Release Voltage	V_{OVL}	26.5	27.5	28.5	V	OVLO Enable
Lockout Voltage	V_{OVH}	27.5	28.5	29.5	V	OVLO Enable
MLP						
Motor Lock Protection Detect Time	t_{MLP}	1.54	2.20	2.86	s	MLP Enable

Timing Chart



I/O Equivalence Circuits

Pin No	Pin Name	Equivalence Circuit	Pin No	Pin Name	Equivalence Circuit
1	FG		2 3 4	PWM B CW BRKB	
5	IREF		6 27	SEL1 SEL2	
8 13 14	UL VL WL		9 10 11 12 15 16	U UH VH V W WH	
17 18	RCL RCLS		19 20 21 22 23 24	HUP HUN HVP HVN HWP HWN	

I/O Equivalence Circuits - continued

<p>25</p>	<p>VREG</p>		<p>26</p>	<p>VREG3</p>
<p>28</p>	<p>TEST</p>		<p>29</p>	<p>CP1</p>
<p>30 31 32</p>	<p>CP2 VG VCC</p>			

Operational Notes

1. Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply pins.

2. Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

3. Ground Voltage

Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.

4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

5. Recommended Operating Conditions

The function and operation of the IC are guaranteed within the range specified by the recommended operating conditions. The characteristic values are guaranteed only under the conditions of each item specified by the electrical characteristics.

6. Inrush Current

When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

7. Testing on Application Boards

When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned OFF completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

Operational Notes – continued

8. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

9. Unused Input Pins

Input pins of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So unless otherwise specified, unused input pins should be connected to the power supply or ground line.

10. Regarding the Input Pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

- When $GND > Pin A$ and $GND > Pin B$, the P-N junction operates as a parasitic diode.
- When $GND > Pin B$, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be avoided.

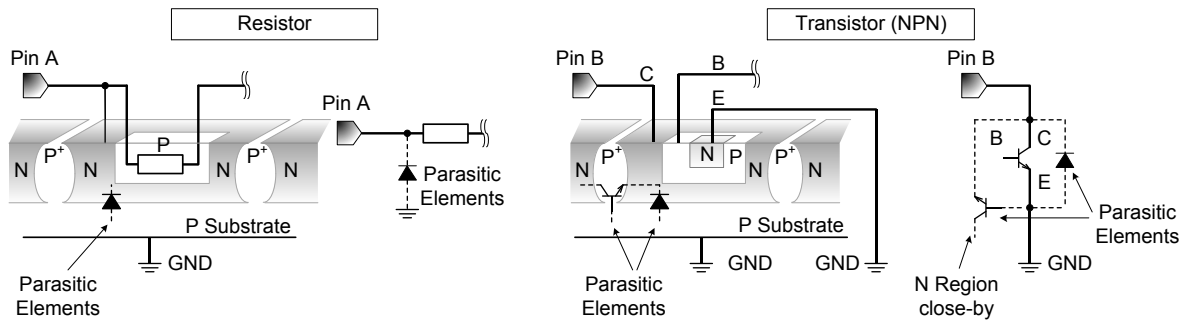


Figure 6. Example of Monolithic IC Structure

11. Ceramic Capacitor

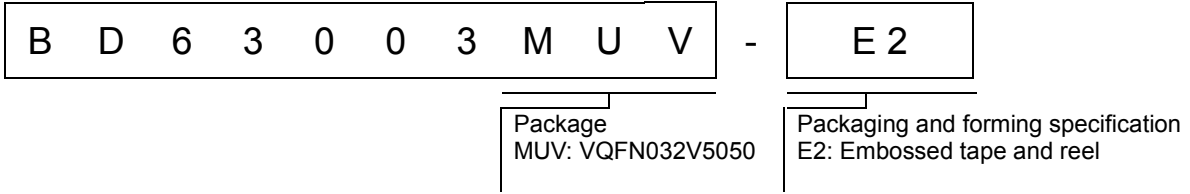
When using a ceramic capacitor, determine a capacitance value considering the change of capacitance with temperature and the decrease in nominal capacitance due to DC bias and others.

12. Thermal Shutdown Circuit (TSD)

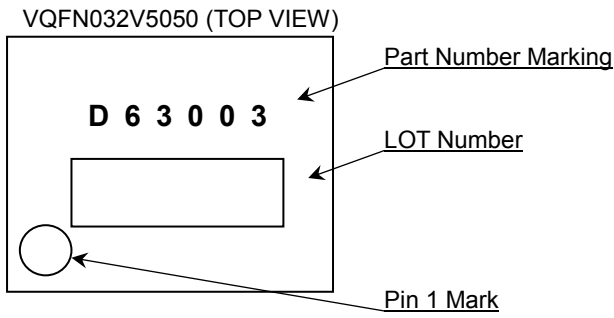
This IC has a built-in thermal shutdown circuit that prevents heat damage to the IC. Normal operation should always be within the IC's maximum junction temperature rating. If however the rating is exceeded for a continued period, the junction temperature (T_j) will rise which will activate the TSD circuit that will turn OFF power output pins. When the T_j falls below the TSD threshold, the circuits are automatically restored to normal operation.

Note that the TSD circuit operates in a situation that exceeds the absolute maximum ratings and therefore, under no circumstances, should the TSD circuit be used in a set design or for any purpose other than protecting the IC from heat damage.

Ordering Information

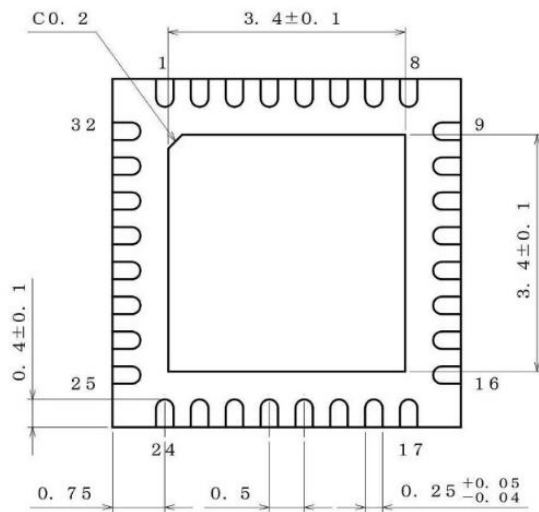
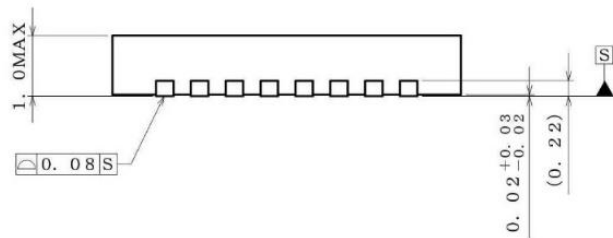
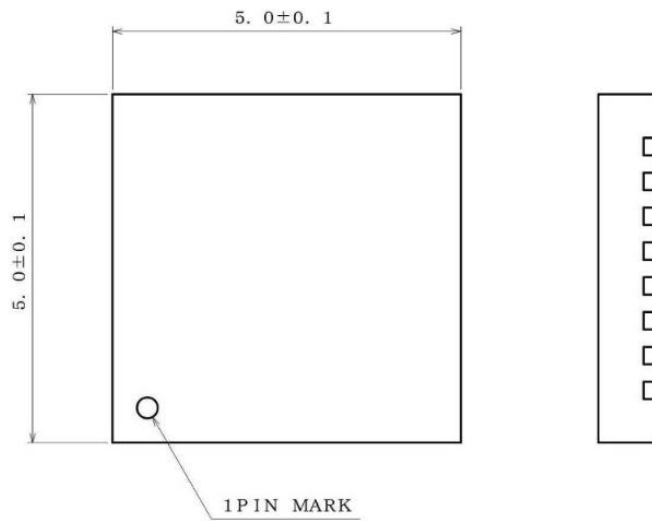


Marking Diagram



Physical Dimension and Packing Information

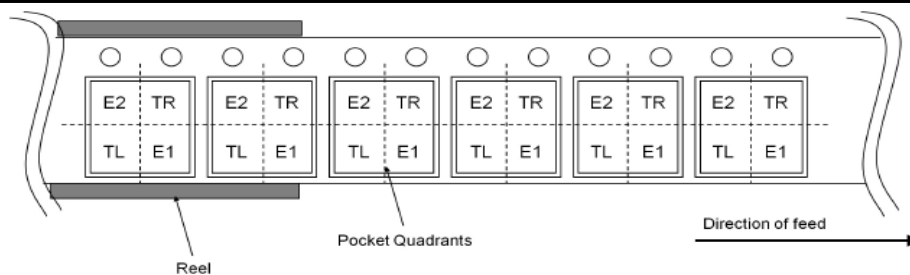
Package Name	VQFN032V5050
--------------	--------------



(UNIT : mm)
 PKG : VQFN032V5050
 Drawing No. EX461-5001-2

< Tape and Reel Information >

Tape	Embossed carrier tape
Quantity	2500pcs
Direction of feed	E2 The direction is the pin 1 of product is at the upper left when you hold reel on the left hand and you pull out the tape on the right hand



Revision History

Date	Revision	Changes
10.Jul.2020	001	New Release

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JAPAN	USA	EU	CHINA
CLASS III	CLASS III	CLASS II b	CLASS III
CLASS IV		CLASS III	

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 - Use of our Products in places where the Products are exposed to static electricity or electromagnetic waves
 - Use of our Products in proximity to heat-producing components, plastic cords, or other flammable items
 - Sealing or coating our Products with resin or other coating materials
 - Use of our Products without cleaning residue of flux (Exclude cases where no-clean type fluxes is used. However, recommend sufficiently about the residue.) ; or Washing our Products by using water or water-soluble cleaning agents for cleaning residue after soldering
 - Use of the Products in places subject to dew condensation
- The Products are not subject to radiation-proof design.
- Please verify and confirm characteristics of the final or mounted products in using the Products.
- In particular, if a transient load (a large amount of load applied in a short period of time, such as pulse, is applied, confirmation of performance characteristics after on-board mounting is strongly recommended. Avoid applying power exceeding normal rated power; exceeding the power rating under steady-state loading condition may negatively affect product performance and reliability.
- De-rate Power Dissipation depending on ambient temperature. When used in sealed area, confirm that it is the use in the range that does not exceed the maximum junction temperature.
- Confirm that operation temperature is within the specified range described in the product specification.
- ROHM shall not be in any way responsible or liable for failure induced under deviant condition from what is defined in this document.

Precaution for Mounting / Circuit board design

- When a highly active halogenous (chlorine, bromine, etc.) flux is used, the residue of flux may negatively affect product performance and reliability.
- In principle, the reflow soldering method must be used on a surface-mount products, the flow soldering method must be used on a through hole mount products. If the flow soldering method is preferred on a surface-mount products, please consult with the ROHM representative in advance.

For details, please refer to ROHM Mounting specification

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Precaution for Storage / Transportation

1. Product performance and soldered connections may deteriorate if the Products are stored in the places where:
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 - [b] the temperature or humidity exceeds those recommended by ROHM
 - [c] the Products are exposed to direct sunshine or condensation
 - [d] the Products are exposed to high Electrostatic
2. Even under ROHM recommended storage condition, solderability of products out of recommended storage time period may be degraded. It is strongly recommended to confirm solderability before using Products of which storage time is exceeding the recommended storage time period.
3. Store / transport cartons in the correct direction, which is indicated on a carton with a symbol. Otherwise bent leads may occur due to excessive stress applied when dropping of a carton.
4. Use Products within the specified time after opening a humidity barrier bag. Baking is required before using Products of which storage time is exceeding the recommended storage time period.

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