

Product Change Notification: SYST-18ZEKG692

Date:

05-Nov-2024

Product Category:

Switchmode PWM Controller Ics

Notification Subject:

Data Sheet - HV911X Family Data Sheet

Affected CPNs:

SYST-18ZEKG692_Affected_CPN_11052024.pdf SYST-18ZEKG692_Affected_CPN_11052024.csv

Notification Text:

SYST-18ZEKG692
Microchip has released a new Datasheet for the HV911X Family Data Sheet of devices. If you are using one of these devices please read the document located at HV911X Family Data Sheet .
Notification Status: Final
Description of Change:
Updated section Features.
• Updated Functional Block Diagram for HV9110/HV9112 and Functional Block Diagram for HV9113.
 Updated package drawing in section Packaging Information.
Impacts to Data Sheet: See above details
Reason for Change: To Improve Productivity

Change Implementation Status: Complete

Date Document Changes Effective: 05 November 2024

NOTE: Please be advised that this is a change to the document only the product has not been changed.

Markings to Distinguish Revised from Unrevised Devices: N/A

Attachments:

HV911X Family Data Sheet

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Affected Catalog Part Numbers (CPN)

HV9110NG-G HV9110NW-8 HV9110X-AD1 HV9110X-AD1 HV9110X-AD0 HV9112NG-G HV9113NG-G HV9113X HV9113X-AD0



High-Voltage, Current-Mode, PWM Controller

Features

- Input Voltage Range of V_{DD} Regulator
 - HV9110: 10V to 120V
 - HV9112: 9V to 80V
 - HV9113: 10V to 120V
- Maximum Duty, Feedback Accuracy
 - HV9110: 49%, 1%
 - HV9112: 49%, 2%
 - HV9113: 99%, 1%
- Current Mode Control
- <1 mA Supply Current
- ≤1 MHz Clock

Applications

DC/DC Power Converters

Description

HV9110/HV9112/HV9113 are Switch-Mode Power Supply (SMPS) controllers suitable for the control of a variety of converter topologies, including the flyback converter and the forward converter.

The V_{DD} regulator supports an input voltage as high as 80V or 120V.

HV9110/HV9112/HV9113 controllers include all essentials for a power converter design, such as a bandgap reference, an error amplifier, a ramp generator, a high-speed PWM comparator, and a gate driver. A shutdown latch provides on/off control.

The HV9110 and HV9113 feature an input voltage range of 10V to 120V, and the HV9112 has an input voltage range of 9V to 80V. The HV9110 and HV9112 have a maximum duty of 49%, while the HV9113 has a maximum duty of 99%.

Package Type



Functional Block Diagram



Functional Block Diagram



1.0 ELECTRICAL CHARACTERISTICS

ABSOLUTE MAXIMUM RATINGS[†]

Input Voltage, V _{IN}	
HV9110/HV9113	
HV9112	
Device Supply Voltage, V _{DD}	
Logic Input Voltage Range	-0.3V to V _{DD} + 0.3V
Linear Input Voltage Range	-0.3V to V _{DD} + 0.3V
Storage Temperature Range	-65°C to +150°C
Operating Temperature Range	
Power Dissipation: 14-lead SOIC	

† Notice: Stresses above those listed under "Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operational listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

Electrical Specifications: V_{DD} = 10V, V_{IN} = 48V, V_{DISC} = 0V, R_{BIAS} = 390 k Ω , R_{OSC} = 330 k Ω , T_A = 25°C, unless otherwise noted.								
Parame	Sym.	Min.	Тур.	Max.	Units	Conditions		
REFERENCE								
Output Voltage	HV9110/13	V _{REF}	3.92	4	4.08	V	R _L = 10 MΩ	
	HV9112		3.88	4	4.12			
	HV9110/13		3.82	4	4.16		R _L = 10 MΩ, T _A = -55°C to +125°C	
Output Impedance		Z _{OUT}	15	30	45	kΩ	(Note 1)	
Short Circuit Current		I _{SHORT}	—	125	250	μA	V _{REF} = GND	
Change in V_{REF} with Temperature		ΔV_{REF}		0.25		mV/°C	T _A = -55°C to +125°C (Note 1)	
OSCILLATOR								
Oscillator Frequency		f _{MAX}	1.0	3	—	MHz	R _{OSC} = 0Ω	
Initial Accuracy		f _{OSC}	80	100	120	kHz	R _{OSC} = 330 kΩ (Note 2)	
			160	200	240		R _{OSC} = 150 kΩ (Note 2)	
V _{DD} Regulation		_		_	15	%	9.5V < V _{DD} < 13.5V	
Temperature Coefficient		_	—	170	-	ppm/°C	T _A = -55°C to +125°C (Note 1)	
PWM					-			
Maximum Duty	HV9110/HV9112	D _{MAX}	49	49.4	49.6	%	(Note 1)	
Cycle	HV9113		95	97	99			
Dead Time HV9113		D _{MIN}	_	225	—	ns	HV9113 only (Note 1)	
Minimum Duty Cycle			_	—	0	%		
Pulse Width where Pulse drops out			_	80	125	ns	(Note 1)	
CURRENT LIMIT			-					
Maximum Input Signal		V _{LIM}	1	1.2	1.4	V	V _{FB} = 0V	
Delay to Output		t _D		80	120	ns	V _{CS} = 1.5V, V _{COMP} ≤ 2V (Note 1)	

Electrical Specifications: V_{DD} = 10V, V_{IN} = 48V, V_{DISC} = 0V, R_{BIAS} = 390 k Ω , R_{OSC} = 330 k Ω , T_A = 25°C, unless otherwise noted. **Parameters** Sym. Min. Тур. Max. Units Conditions ERROR AMPLIFIER Feedback Voltage HV9110/13 3.96 4 4.04 V V_{FB} shorted to COMP V_{FB} HV9112 3.92 4 4.08 Input Bias Current 25 500 nA V_{FB}= 4V IIN Input Offset Voltage Nulled during trim ____ Vos **Open-loop Voltage Gain** 60 (Note 1) 80 dB Avoi Unity Gain Bandwidth GB 1 1.3 MHz (Note 1) ____ **Output Source Current** -1.4 -2 V_{FB} = 3.4V ____ mΑ ISOURCE 0.12 **Output Sink Current** 0.15 mΑ $V_{FB} = 4.5V$ ISINK ____ HIGH-VOLTAGE REGULATOR AND START-UP Input Voltage HV9110/13 120 V I_{IN}< 10 μA; V_{CC} > 9.4V VIN ____ HV9112 80 ____ ____ Input Leakage Current 10 μA V_{DD} > 9.4V IIN Regulator Turn-off Threshold Voltage 8 8.7 9.4 V I_{IN} = 10 μA V_{TH} Undervoltage Lockout VLOCK 7 8.9 V 8.1 SUPPLY Supply Current 0.75 1 mΑ $C_{l} < 75 \, pF$ I_{DD} **Quiescent Supply Current** 0.55 mΑ V_{NSD} = 0V lQ ____ ____ Nominal Bias Current 20 μA I_{BIAS} ____ **Operating Range** V_{DD} 9 ____ 13.5 V SHUTDOWN LOGIC C_L= 500 pF, V_{CS}= 0V Shutdown Delay t_{SD} 50 100 ns (Note 1) NSD Pulse Width 50 (Note 1) t_{SW} ____ ns **RST Pulse Width** 50 (Note 1) t_{RW} ____ ns 25 Latching Pulse Width V_{NSD.} V_{RST} = 0V (Note 1) t_{LW} ____ ns Input Low Voltage V_{IL} 2 V ____ Input High Voltage V_{H} 7 ____ V Input Current, Input High Voltage 1 5 $V_{IN} = V_{DD}$ μA Ι_Η Input Current, Input Low Voltage I_{IL} -25 -35 μA $V_{IN} = 0V$ OUTPUT **Output High Voltage** V_{DD}-0.25 V HV9110/13 I_{OUT} = 10 mA HV9112 V_{DD}-0.3 _ ____ V_{OH} HV9110/13 I_{OUT} = 10 mA, $V_{DD} - 0.3$ $T_A = -55^{\circ}C$ to 125°C **Output Low Voltage** VOL 0.2 V All I_{OUT}= -10 mA _ HV9110/13 I_{OUT}= -10 mA, 0.3 T_A= -55°C to 125°C Output Resistance Pull up 15 25 R_{OUT} I_{OUT}= ±10 mA Ω Pull down 20 8 ____ Pull up 20 30 I_{OUT} = ±10 mA, Ω T_A= -55°C to 125°C Pull down 10 30 _ **Rise Time** 30 75 $C_1 = 500 \text{ pF} (\text{Note 1})$ t_R ns 75 Fall Time 20 $C_1 = 500 \text{ pF}(\text{Note 1})$ t⊨ ns

ELECTRICAL CHARACTERISTICS (CONTINUED)

Note 1: Design guidance only; Not 100% tested in production.

2: Stray capacitance on OSC input pin must be \leq 5 pF.

TEMPERATURE SPECIFICATIONS

Parameters	Sym.	Min.	Тур.	Max.	Units	Conditions	
TEMPERATURE RANGES							
Operating Temperature	—	-55	_	125	°C		
Storage Temperature	—	-65	_	150	°C		
PACKAGE THERMAL RESISTANCES							
14-lead SOIC	θ _{ja}	_	83	_	°C/W		

1.1 Truth Table

TRUTH TABLE

SHUTDOWN	RESET	OUTPUT
Н	Н	Normal operation
Н	$H \rightarrow L$	Normal operation, no change
L	Н	Off, not latched
L	L	Off, latched
$L \rightarrow H$	L	Off, latched, no change

2.0 TYPICAL PERFORMANCE CURVES

Note: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g. outside specified power supply range) and therefore outside the warranted range.



FIGURE 2-1: Error Amplifier Output Impedance (Z0).



FIGURE 2-2: PSRR –Error Amplifier and Reference.





FIGURE 2-4: Output Switching Frequency vs. Oscillator Resistance.



FIGURE 2-5: Error Amplifier Open-Loop Gain/Phase.



FIGURE 2-6: R_{DIS0} (HV9113 only).

R_{DISCHARGE} vs. t_{OFF}

3.0 PIN DESCRIPTION

Table 3-1 shows the pin description for HV9110/HV9112/HV9113. The locations of the pins are listed in Features.

Pin Number	HV9110/HV9112/HV9113 Pin Name	Description
1	BIAS	Internal bias, current set
2	V _{IN}	High-voltage V _{DD} regulator input
3	CS	Current sense input
4	GATE	Gate drive output
5	GND	Ground
6	V _{DD}	High-voltage V _{DD} regulator output
7	OSCO	Oscillator output
8	OSCI	Oscillator input
9	DISC*	Oscillator discharge, current set **
10	V _{REF}	4V reference output Reference voltage level can be overridden by an externally applied voltage source.
11	NSD	Active low input to set shutdown latch
12	RST	Active high input to reset shutdown latch
13	COMP	Error amplifier output
14	FB	Feedback voltage input

TABLE 3-1: PIN DESCRIPTION

Remarks:

* NC for HV9110/HV9112

** No connection for HV9110/HV9112

4.0 TEST CIRCUITS

The test circuits for characterizing error amplifier output impedance, Z_{OUT}, and error amplifier, power supply rejection ratio, PSRR, are shown in Figure 4-1 and Figure 4-2.





5.0 DETAILED DESCRIPTION

5.1 High-Voltage Regulator

The high-voltage regulator included in HV9110/HV9112/HV9113 consists of a high-voltage N-Channel Depletion-mode DMOS transistor driven by an error amplifier, providing a current path between the VIN terminal and the $\ensuremath{\mathsf{V}_{\text{DD}}}$ terminal. The maximum current, about 20 mA, occurs when V_{DD} = 0, with current reducing as V_{DD} rises. This path shuts off when V_{DD} rises to somewhere between 8V and 9.4V. So, if V_{DD} is held at 10V or 12V by an external source, no current other than leakage is drawn through the high voltage transistor. This minimizes dissipation within the high-voltage regulator.

Use an external capacitor between V_{DD} and GND. This capacitor should have good high-frequency characteristics. Ceramic caps work well.

The device uses a compound resistor divider to monitor VDD for both the undervoltage lockout circuit and the shutoff circuit of the high-voltage FET. Setting the undervoltage sense point about 0.6V lower on the string than the FET shutoff point guarantees that the undervoltage lockout releases before the FET shuts off.

5.2 Bias Circuit

HV9110/HV9112/HV9113 require an external bias resistor, connected between the Bias pin and GND, to set currents in a series of current mirrors used by the analog sections of the chip. The nominal external bias current requirement is 15 μA to 20 μA, which can be set by a 390 kΩ to 510 kΩ resistor if V_{DD} = 10V, or a 510 kΩ to 680 kΩ resistor if V_{DD} = 12V. A precision resistor is not required, ±5% meets device requirements.

5.3 Clock Oscillator

The clock oscillator of the HV9110/HV9112/HV9113 consists of a ring of CMOS inverters, timing capacitors, and a capacitor-discharge FET. A single external resistor between the OSCI and OSCO sets the oscillator frequency. (See Figure 2-4.)

The HV9110 and HV9112 include a frequency-dividing flip-flop that allows the part to operate with a 50% duty limit. Accordingly, the effective switching frequency of the power converter is half the oscillator frequency. (See Figure 2-4.)

An internal discharge FET resets the oscillator ramp at the end of the oscillator cycle. The discharge FET is externally connected to GND, by way of a resistor. The resistor programs the oscillator dead time at the end of the oscillator period. The oscillator turns off during shutdown to reduce supply current by about 150 $\mu A.$

5.4 Reference

The reference of the HV9110/HV9112/HV9113 consists of a band-gap reference, followed by a buffer amplifier, which scales the voltage up to 4V. The scaling resistors of the buffer amplifier are trimmed during manufacture so that the output of the error amplifier, when connected in a gain of -1 configuration, is as close to 4V as possible. This nulls out the input offset of the error amplifier. As a consequence, even though the observed reference voltage of a specific part may not be exactly 4V, the feedback voltage required for proper regulation will be 4V.

An approximately 50 k Ω resistor is located internally between the output of the reference buffer amplifier and the circuitry it feeds - reference output pin and noninverting input to the error amplifier. This allows overriding the internal reference with a low impedance voltage source \leq 6V. Using an external reference reinstates the input offset voltage of the error amplifier. Overriding the reference should seldom be necessary.

The reference of the HV9110/HV9112/HV9113 is a high-impedance node, and usually there will be significant electrical noise nearby. Therefore, a bypass capacitor between the reference pin and GND is strongly recommended. The reference buffer amplifier is compensated to be stable with a capacitive load of 0.01 μ F to 0.1 μ F.

5.5 Error Amplifier

The error amplifier on HV9110/HV9112/HV9113 is a low-power, differential-input, operational amplifier. A PMOS input stage is used, so the common mode range includes ground and the input impedance is high.

5.6 Current Sense Comparators

The HV9110/HV9112/HV9113 use a dual-comparator system with independent comparators for modulation and current limiting. This provides the designer greater latitude in compensation design, as there are no clamps, except ESD protection, on the compensation pin.

5.7 Remote Shutdown

The NSD and RST pins control the shutdown latch. These pins have internal current-source pull-ups so they can be driven from open drain logic. When not used they should be left open or connected to V_{DD} .

5.8 Output Buffer

The output buffer of HV9110/HV9112/HV9113 is of standard CMOS construction P-channel pull-up and N-Channel pull-down. Thus, the body-drain diodes of the output stage can be used for spike clipping. External Schottky diode clamping of the output is not required.



FIGURE 5-1: Shutdown Timing Waveforms.

6.0 PACKAGING INFORMATION

6.1 Package Marking Information





Legend	: XXX Y YY WW NNN @3 *	Product Code or Customer-specific information Year code (last digit of calendar year) Year code (last 2 digits of calendar year) Week code (week of January 1 is week '01') Alphanumeric traceability code Pb-free JEDEC [®] designator for Matte Tin (Sn) This package is Pb-free. The Pb-free JEDEC designator ((e3)) can be found on the outer packaging for this package.
Note:	In the even be carried characters not include	nt the full Microchip part number cannot be marked on one line, it will d over to the next line, thus limiting the number of available s for product code or customer-specific information. Package may or e the corporate logo.

14-Lead Plastic Small Outline (D3X) - Narrow, 3.90 mm Body [SOIC]



Microchip Technology Drawing No. C04-065-D3X Rev E Sheet 1 of 2

14-Lead Plastic Small Outline (D3X) - Narrow, 3.90 mm Body [SOIC]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	MILLIMETERS				
Dimension	l Limits	MIN	NOM	MAX	
Number of Pins	N		14		
Pitch	е		1.27 BSC		
Overall Height	Α	-	-	1.75	
Molded Package Thickness	A2	1.25	-	-	
Standoff §	A1	0.10	-	0.25	
Overall Width	E	6.00 BSC			
Molded Package Width	E1	3.90 BSC			
Overall Length	D		8.65 BSC		
Chamfer (Optional)	h	0.25 - 0.50			
Foot Length	L	0.40	-	1.27	
Footprint	L1	1.04 REF			
Lead Angle	Θ	0°	-	-	
Foot Angle	φ	0° - 8°			
Lead Thickness	С	0.10 - 0.25			
Lead Width	b	0.31	-	0.51	
Mold Draft Angle Top	α	5°	-	15°	
Mold Draft Angle Bottom	β	5°	-	15°	

Notes:

- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. § Significant Characteristic
- 3. Dimension D does not include mold flash, protrusions or gate burrs, which shall not exceed 0.15 mm per end. Dimension E1 does not include interlead flash or protrusion, which shall not exceed 0.25 mm per side.
- 4. Dimensioning and tolerancing per ASME Y14.5M
 - BSC: Basic Dimension. Theoretically exact value shown without tolerances. REF: Reference Dimension, usually without tolerance, for information purposes only.
- 5. Datums A & B to be determined at Datum H.

Microchip Technology Drawing No. C04-065-D3X Rev E Sheet 2 of 2

14-Lead Plastic Small Outline (D3X) - Narrow, 3.90 mm Body [SOIC]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



RECOMMENDED LAND PATTERN

	MILLIMETERS			
Dimension	MIN	NOM	MAX	
Contact Pitch	E		1.27 BSC	
Contact Pad Spacing	С		5.40	
Contact Pad Width (X14)				0.60
Contact Pad Length (X14)	Y			1.55

Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2065-D3X Rev E

NOTES:

APPENDIX A: REVISION HISTORY

Revision B (October 2024)

- Updated section Features.
- Updated Functional Block Diagram for HV9110/HV9112 and Functional Block Diagram for HV9113.
- Updated package drawing in section **Packaging Information**.

Revision A (June 2016)

- Merged Supertex Doc #s DSFP-HV9110, DSFP-HV9112 and DSFP-DSFP-HV9113 to Microchip DS20005505A.
- Revised Electrical Characteristics to accommodate the merged products.
- Updated pin names to reflect new naming convention.
- Significant text changes to **Detailed Description**.
- · Minor text changes throughout.

PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

PART NO.	<u>xx</u>	- <u>x</u> - x	Ex	amples:	
Device	Package Options	Environmental Media Type	a)	HV9110NG-G:	High-Voltage Current-mode PWM Controller, 10V to 120V Input Voltage Range, 49% Duty Cycle, 14-lead SOIC package, 53/Tube
Device:	HV9110	 High-Voltage Current-mode PWM Controller, 10V to 120V Input Voltage Range, 49% Duty Cycle 	b)	HV9112NG-G	High-Voltage Current-mode PWM Controller, 9V to 80V Input Voltage Range, 49% Duty Cycle, 14-lead SOIC package, 53/Tube
	HV9112 HV9113	 High-Voltage Current-mode PWM Controller, 9V to 80V Input Voltage Range, 49% Duty Cycle High-Voltage Current-mode PWM Controller, 10V to 120V Input Voltage Range, 99% Duty Cycle 	c)	HV9113NG-G	High-Voltage Current-mode PWM Controller, 10V to 120V Input Voltage Range, 99% Duty Cycle,14-lead SOIC package, 53/Tube
Package:	NG	= 14-lead SOIC			
Environmental	G	= Lead (Pb)-free/ROHS-compliant Package			
Media Type:	(blank)	= 53/Tube for an NG package			

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